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TRS3122E SLLSET7C - MAY 2016 - REVISED MAY 2016

TRS3122E 1.8 V Low Power Dual RS-232 Transceiver

Technical

Documents

Features 1

- Extended V_{CC} operating nodes: 1.8 V, 3.3 V, or 5.0 V
 - Unique Tripler Charge Pump Architecture Enables Low V_{CC} of 1.8V While Maintaining Compatibility with 3.3 V and 5 V Supplies
- Integrated Level-Shifting Functionality Eliminates the Need for External Power or Additional Level Shifter While Interfacing with Low-Voltage MCUs
- Enhanced ESD Protection on RIN Inputs and **DOUT** Outputs
 - ±15 kV IEC 61000-4-2 Air-Gap Discharge
 - ±8 kV IEC 61000-4-2 Contact Discharge
 - ±15 kV Human-Body Model
- Specified 1000-kbps Data Rate
- Auto Powerdown Plus Feature
- Low 0.5uA Shutdown Supply Current
- Meets or Exceeds Compatibility Requirements of RS-232 Interface
- For 2.5 V Single Supply Applications, Consider TRS3318E as an Optimized Solution

Applications 2

- Remote Radio Unit (RRU)
- Base Band Unit (BBU)
- Electronic Point of Sale (EPOS)
- **Diagnostics & Data Transmission**
- **Battery-Powered Equipment**

3 Description

Tools &

Software

The TRS3122E is a two-driver and two-receiver RS-232 interface device, with split supply pins for mixedvoltage operation. All RS-232 inputs and outputs are protected to ±15 kV using the IEC 61000-4-2 Air-Gap Discharge method, ±8 kV using the IEC 61000-4-2 Contact Discharge method, and ±15 kV using the Human-Body Model.

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The charge pump requires five small 0.1-µF capacitors for operation from as low as a 1.8-V supply. The TRS3122E is capable of running at data rates up to 1000 kbps, while maintaining RS-232compatible output levels.

The TRS3122E has a unique V_{L} pin that allows operation in mixed-logic voltage systems. Both driver in (DIN) and receiver out (ROUT) logic levels are pin programmable through the V_L pin. This eliminates the need for additional voltage level shifter while interfacing with low-voltage microcontrollers or UARTs.

Auto Powerdown Plus automatically places the device in a low power mode when the device has not received or transmitted data for more than 30 seconds. This feature makes this device a very attractive option for battery powered or other powersensitive applications.

Device Information⁽¹⁾

		-	
PART NUMBER	PACKAGE (PINS)	BODY SIZE (NOM)	
TRS3122ERGER	RGE (24)	4.00 mm × 4.00 mm	

(1) For all available packages, see the orderable addendum at the end of the datasheet.

Feature Diagram



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4 Revision History

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C	hanges from Revision B (May 2016) to Revision C Page Reordered bullets in <i>Features</i> section by priority 1 Deleted Data Cables from <i>Applications</i> section of front page to make room for front page image 1 Changed Corrected swapped ESD levels in description text 1	
•	Reordered bullets in <i>Features</i> section by priority	1
•	Deleted Data Cables from Applications section of front page to make room for front page image	1
•	Changed Corrected swapped ESD levels in description text	1
C	hanges from Revision A (May 2016) to Revision B	Page

CI	hanges from Revision A (May 2016) to Revision B	Page
•	Updated ESD ratings values to reflect current device specifications	4
•	Added all Typical Characteristic graphs and schematics to the Typical Characteristics section	8
•	Added Application Curve image to Application Curves section	18

Changes from Original (June 2014) to Revision A			е
•	Added Pin Functions table.		3

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5 Pin Configuration and Functions



Pin Functions

Pin		1/0	DESCRIPTION		
NAME	NO.	1/0	DESCRIPTION		
C1+, C2+	21, 22	-	Positive terminals of voltage-doubler charge-pump capacitors (required)		
C3+	1	-	Positive terminal of voltage-tripler charge-pump capacitor (Not needed for VCC 3V to 5.5V)		
C1–, C2-	16, 23	-	pative terminals of voltage-doubler charge-pump capacitors (required)		
C3-	2	-	Negative terminal of voltage-tripler charge-pump capacitor (Not needed for VCC 3V to 5.5V)		
V+	20	-	Positive charge pump storage capacitor (required)		
V–	24	-	Negative charge pump storage capacitor (required)		
GND	17	-	Ground		
V _{CC}	19	-	1.8-V or 3-V to 5-V supply voltage		
VL	18	-	Logic-level supply. All CMOS inputs (DIN) and outputs (ROUT) are referenced to this supply.		
FORCEOFF	14	I	Auto Powerdown Control input (Refer to Truth Table)		
FORCEON	15	I	Auto Powerdown Control input (Refer to Truth Table)		
INVALID	13	0	Invalid Output Pin		
DIN1, DIN2	10,9	I	Driver inputs		
DOUT1, DOUT2	5, 6	0	RS-232 driver outputs		
RIN1, RIN2	3, 4	I	RS-232 receiver inputs		
ROUT1, ROUT2	12, 11	0	Receiver outputs; swing between 0 and V_L		
NC	7, 8	I	Factory pins, can be unconnected or connected to GND		



6 Specifications

6.1 Absolute Maximum Ratings⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

				MIN	MAX	UNIT
V_{CC}	Charge pump power sup	oply		-0.3	6	V
V_{L}	Logic power supply			-0.3	6	V
V+	Positive storage capacito	or voltage		-0.3	7	V
V–	Negative storage capacitor voltage			0.3	-7	V
	$V + + V - ^{(2)}$				13	V
	Input voltage	FORCEOFF, FORCEON		-0.3	6	
V		DIN		-0.3	$V_{L} + 0.3$	v
VI		RIN (0Ω series resistance)			±20	v
		RIN (≥250Ω series resistance)			±25	Ī
V	Output valtage	DOUT			±13.2	V
Vo	Output voltage	ROUT		-0.3	V _L + 0.3	v
TJ	Junction temperature	lunction temperature			150	°C
T _{stg}	Storage temperature ran	nge		-65	150	°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

(2) V+ and V- can have maximum magnitudes of 7 V, but their absolute difference cannot exceed 13 V.

6.2 ESD Ratings

				VALUE	UNIT
V _(ESD) Electrostatic discharge		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001,	M), per ANSI/ESDA/JEDEC JS-001, All pins except RS-232 bus		
	all pins ⁽¹⁾	RS-232 bus pins	±15000		
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	All pins	±500	V
		PS 222 hug ning	±15000		
		IEC 61000-4-2 Contact Discharge	RS-232 bus pins	±8000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



6.3 Recommended Operating Conditions

				MIN	TYP	MAX	UNIT
	Charge pump power supply		Tripler Mode	1.65	1.8	2	
V _{CC}			Doubler Mode	3	3.3	3.6	V
			Doublet Mode	4.5	5	5.5	
VL	Logic power supply			1.65		V_{CC}	V
RIN	RS-232 Receiver int	erface		-15		15	V
DOUT	RS-232 Transmitter	interface		-12		12	V
			$V_{L} = 5.0 V$	0		1.7	
V _{IL} GPIO Input logic	GPIO Input logic DIN, FORCEOFF, FORCEON	$V_L = 3.3V$	0		1.1	V	
			V _L = 1.8 V	0		0.6	
			$V_{L} = 5.0V$	3.3		V_{L}	
VIH	GPIO Input logic threshold high	DIN, FORCEOFF, FORCEON	$V_{L} = 3.3V$	2.2		V_{L}	V
		$V_{L} = 1.8V$	1.2		V_{L}		
V _{OZ}	ROUT disabled	·	FORCEOFF = 0V	0		V_{L}	V
	Operating temperatu	ıre		-40		85	°C

6.4 Thermal Characteristics

	THERMAL METRIC	TRS3122E	
		RGE	UNIT
R_{\thetaJA}	Junction-to-ambient thermal resistance	34.2	
R _{0JCtop}	Junction-to-case (top) thermal resistance	27.2	
$R_{\theta JB}$	Junction-to-board thermal resistance	11.4	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	0.4	0.00
Ψ_{JB}	Junction-to-board characterization parameter	11.4	
R _{0JC(bot)}	Junction-to-case (bottom) thermal resistance	3.6	

6.5 Power and Status Electrical Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{ V to } 5.5 \text{ V}), T_A = -40^{\circ}\text{C}$ to 85°C (unless otherwise noted). Typical data is $T_A = 25^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$ unless otherwise noted.

	PARAMETER		TEST CONDITIONS	6	MIN	TYP	MAX	UNIT
		DIN1 = GND or V_L ; V_{CC} = 1.65V to 2.0V			1.0	1.9		
Icc (Static))	$\frac{\text{DIN2} = \text{GND}}{\text{FORCEOFF}} = V_1$	No load	V_{CC} = 3.0V to 3.6V		0.7	1.4	mA
		FORCEON = V_L		V_{CC} = 4.5V to 5.5V		0.8	1.9	
Icc (off)		FORCEOFF = GND				0.4	10	μA
V _{IT+}	RIN postive voltage threshold for INVALID output change						2.4	V
V _{IT-}	RIN negative <u>voltage</u> threshold for INVALID output change		RIN1 = RIN2				-0.3	v
V _{OH}	INVALID high-level output voltage	$I_{OH} = -1 \text{ mA}$, FORCEON = GND, FORCEOFF = VL			V _L -0.4	V _L -0.08	V_{L}	V
V _{OL}	INVALID low-level output voltage	I _{OL} = 1.6 mA, FORCE	I_{OL} = 1.6 mA, FORCEON = GND, FORCEOFF = VL			0.06	0.4	V

RUMENTS

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6.6 Driver Electrical Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{ V to } 5.5 \text{ V}), T_A = -40^{\circ}\text{C}$ to 85°C (unless otherwise noted). Typical data is $T_A = 25^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$ unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V	Output voltage swing	All driver outputs loaded with 3 k Ω to ground C3 = 100 nF, V_{CC}= 1.8 V	±4.25	±4.7		V
V _{OUT}		All driver outputs loaded with 3 k Ω to ground C3 = 0 F, V_{CC}= 3.3 V or 5 V	±5	±5.4		v
r _O	Output resistance	$(V_{CC} = V + = V - = 0)$; Driver output = ±2 V	300	10M		Ω
I _{OS}	Output short-circuit current	V _{DOUT} = 0			±60	mA
I _{OZ}	Output leakage current	V _{DOUT} = ±12 V, FORCEOFF = GND	0		±25	μA
	Driver input hysteresis			0.5	1	V
	Input leakage current	DIN = GND to V_L ; FORCEOFF = GND to V_L ; FORCEON = GND to V_L		0	±5	μA

6.7 Receiver Electrical Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{ V to } 5.5 \text{ V}), T_A = -40^{\circ}\text{C}$ to 85°C (unless otherwise noted). Typical data is $T_A = 25^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$ unless otherwise noted.

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
I _{off}	Output leakage current	ROUT, receivers disable	d		±0.01	±10	μA
V _{OL}	Output voltage low	I _{OUT} = 2.0 mA			0.04	0.3	V
V _{OH}	Output voltage high	I _{OUT} = -2.0mA		V _L -0.3	V _L -0.04		V
			$V_L = 5 V$	0.8	1.5		
V _{IT-}	Input threshold low T _A =25°C	T _A =25°C	V _L = 3.3 V	0.7	1.1		V
			V _L = 1.8 V	0.6	0.7		
			$V_L = 5 V$		2.0	2.4	
V _{IT+}	Input threshold high	T _A =25°C	$V_{L} = 3.3 V$		1.5	2.4	V
			V _L = 1.8 V		0.9	1.4	
			V _L = 5 V		0.45		
V _{hys}	Input hysteresis	T _A =25°C	$V_{L} = 3.3 V$		0.35		V
-			V _L = 1.8 V		0.26		
	Input resistance	T _A =-40 to 85°C		3	5	7	kΩ

6.8 Driver Switching Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{ V to } 5.5 \text{ V}), T_A = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}$ (unless otherwise noted). Typical data is $T_A = 25^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$ unless otherwise noted.

PARAMETER				MIN	TYP	MAX	UNIT
Maximum data rate		$R_L = 3 k\Omega$, $C_L = 500 pF$ (one	e driver)	1000			khoo
Maximum dai	la rale	$R_{L} = 3 k\Omega, C_{L} = 1000 pF$ (or	e driver)	500			kbps
Time-to-exit p	owerdown	V _{DOUT} > 3.7 V			30	150	μs
t _{PHL} – t _{PLH} Driver skew ⁽¹)	$R_L = 3 k\Omega$	$R_L = 3 k\Omega$		50	100	ns
	Transition-region slew rate	$\begin{array}{l} R_{L}=3\ k\Omega\ to\ 7\ k\Omega,\\ T_{A}=25^{\circ}C\\ \text{Measured from 3 V to -3 V}\\ or\ -3\ V\ to\ 3\ V \end{array}$	$VCC = 1.8V, C_{L} = 200 \text{ pF}$		33		
			VCC = 1.8V, C _L = 1000 pF		25		
Transition roo			VCC = $3.3 \text{ V}, \text{ C}_{\text{L}} = 200 \text{ pF}$		38		Mue
Transmon-reç			$VCC = 3.3 V, C_L = 1000 pF$		28		V/µs
			$VCC = 5 V, C_{L} = 200 pF$		41		
			$VCC = 5 V, C_L = 1000 pF$		30		

(1) Driver skew is measured at the driver zero crosspoint.



6.9 Receiver Switching Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{ V to } 5.5 \text{ V}), T_A = -40^{\circ}\text{C}$ to 85°C (unless otherwise noted). Typical data is $T_A = 25^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$ unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PHL}	Receiver propagation delay, high to low	Receiver input to receiver output		0.15	0.4	
t _{PLH}	Receiver propagation delay, low to high	C _L = 150 pF		0.15	0.4	μs
t _{PHL} – t _{PLH}	Receiver skew			50	300	ns
t _{en}	Receiver output enable time	From FORCEOFF to ROUT= $V_L/2$		200	400	ns
t _{dis}	Receiver output disable time	CL = 150 pF, RL = 3 kΩ		200	400	ns

6.10 Power and Status Switching Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{ V to } 5.5 \text{ V}), T_A = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}$ (unless otherwise noted). Typical data is $T_A = 25^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$ unless otherwise noted.

	PARAMETER	MIN	TYP	MAX	UNIT
t _{valid}	Propagation delay time, low- to high-level output		1		μs
t _{invalid}	Propagation delay time, high- to low-level output		30		μs
t _{dis}	Receiver or driver edge to auto-powerdown plus	15	30	60	S

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6.11 Typical Characteristics





7 Parameter Measurement Information



A. C_L includes probe and jig capacitance.

B. The pulse generator has the following characteristics: PRR = 1000 kbit/s, Z_0 = 50 Ω , 50% duty cycle, $t_r \le 10$ ns, $t_f \le 10$ ns.





TEST CIRCUIT

VOLTAGE WAVEFORMS

- A. C_L includes probe and jig capacitance.
- B. The pulse generator has the following characteristics: PRR = 1000 kbit/s, Z_0 = 50 Ω , 50% duty cycle, $t_r \le 10$ ns, $t_f \le 10$ ns.

Figure 8. Driver Pulse Skew



B. The pulse generator has the following characteristics: $Z_0 = 50 \ \Omega$, 50% duty cycle, $t_r \le 10 \ ns$, $t_f \le 10 \ ns$.

Figure 9. Receiver Propagation Delay Times



Parameter Measurement Information (continued)

- C_L includes probe and jig capacitance.
- B. The pulse generator has the following characteristics: $Z_0 = 50 \Omega$, 50% duty cycle, $t_r \le 10$ ns, $t_f \le 10$ ns.
- C. t_{PLZ} and t_{PHZ} are the same as t_{dis} .

Α.

D. t_{PZL} and t_{PZH} are the same as t_{en} .

Figure 10. Receiver Enable and Disable Times

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Parameter Measurement Information (continued)



Figure 11. INVALID Propagation-Delay Times and Supply-Enabling Time

8 Detailed Description

8.1 Overview

The TRS3122E is an upgrade to standard RS232 transceivers, offering compatibility with modern system needs like 1.8-V GPIO capability, enhanced ESD & ultra low stand-by current. The majority of RS-232 transceivers with 1.8-V GPIO compatibility require a logic supply pin for the I/O translation, in addition to a minimum 3.3 V V_{CC} for all of the other active circuitry on the chip. Unlike these transceivers, TRS3122E can operate with both V_L and V_{CC} equal to 1.8 V. When V_{CC}= 3.0 V to 5.5 V, the charge pump will sense V_{CC} and switch to doubler mode. C1 & C2 are the necessary flying capacitors, C3 is not needed, and the charge pump outputs V+ & V- will regulate to ~+/-5.4 V. When V_{CC}= 1.65 V to 2.0 V, the charge pump will sense V_{CC} and switch to tripler mode. C1, C2 & C3 are all necessary, and the charge pump outputs V+ & V- will regulate to ~+/-2.65*V_{CC} from V_{CC}= 1.65 V to 2.0 V.

With many modern applications expanding into products that use RS232 as a backup communication protocol, it is important for the transceiver to have efficient standby operation. In order to accommodate this, Auto Powerdown Plus has been integrated to shut-off all active circuitry, allowing TRS3122E to achieve an I_{off} of 1 uA.

In order to comply with common interface system needs and environments, the RS-232 receive and transmit I/O pins comply with IEC 61000-4-2 ratings.

8.2 Functional Block Diagram



Figure 12. Schematic



8.3 Feature Description

8.3.1 Charge Pump

The internal power supply consists of a regulated auto-sensing charge pump that provides RS-232 compatible output voltages, over the 1.65 V to 2.0 V and 3.0 V to 5.5 V V_{CC} ranges. The charge pump operates in two modes to efficiently accommodate low voltage (1.8 V) and higher voltage (3.3 V & 5.0 V) supplies.

8.3.1.1 Doubler Mode

The charge pump requires two flying capacitors (C1, C2) and reservoir capacitors (C4, C5) to generate the V+ and V- supplies of approximately ± 5.4 V when V_{CC} is greater than 3 V. When V_{CC} is >2.9V, TRS3122E will sense the supply voltage level and switch the charge pump to a doubler. Hence, no need for a third flying capacitor. C3+ & C3- pins can be left open for proper operation. If a capacitor is placed between C3+ & C3-, the charge pump will ignore this capacitor and still behave as a doubler.

For capacitor choice recommendations, please refer to Table 1.

8.3.1.2 Tripler Mode

The charge pump requires three flying capacitors (C1, C2 & C3) and reservoir capacitors (C4, C5) to generate the V+ and V- supplies of approximately $\pm 2.65 * V_{CC}$ when V_{CC} is greater than 1.65 V. When V_{CC} is <2.1 V, TRS3122E will sense the supply voltage level and switch the charge pump to a tripler.

For capacitor choice recommendations, please refer to Table 1.

8.3.2 Drivers

The drivers are inverting level transmitters that convert TTL or CMOS logic levels to RS-232 levels. For V_{CC} =3.0 V to 5.0 V, the RS-232 output voltage swing is typically ±5.4 V fully loaded and ±5 V minimum fully loaded. For Vcc = 1.8 V, the RS-232 output voltage swing is typically ±4.7 V fully loaded and ±4.25 V minimum fully loaded.

The driver outputs are protected against indefinite short-circuits to ground without degradation in reliability. These drivers are compatible with RS-232 logic levels and all previous RS-232 versions. Unused driver inputs should be connected to GND or VCC.

8.3.3 Receivers

The receivers convert EIA/TIA-232 levels to TTL or CMOS logic output levels. Receivers have an inverting output that can be disabled by using the FORCEOFF pin. Receivers remain active when the Auto Powerdown Plus circuitry autonomously enters a low power state. See Auto Powerdown Plus for more information on the Auto Powerdown mode. If the FORCEOFF pin is manually set low, the receivers will be disabled and put into 3-state mode. In either of these powerdown modes, the device will typically consume about 0.5 uA. The truth table logic of the TRS3122E driver and receiver outputs can be found in Device Functional Modes. Since receiver input is usually from a transmission line where long cable lengths and system interference can degrade the signal, the inputs have a typical hysteresis margin of 300 mV. This ensures that the receiver is virtually immune to noisy transmission lines. Should an input be left unconnected, an internal $5k\Omega$ pull-down resistor to ground will commit the output of the receiver to a HIGH state.

8.3.4 ESD Protection

ESD protection structures are incorporated on all pins to protect against electrostatic discharges encountered during handling and assembly. The bus pins (driver outputs and receiver inputs) have extra protection structures, which have been tested up to ± 15 kV.

ESD protection is tested in various ways. TI uses the following standards to qualify the ESD structures designed into TRS3122E:

- ±8 kV using IEC 61000-4-2 Contact Discharge (on RINx and DOUTx pins)
- ±15 kV using IEC 61000-4-2 Airgap Discharge (on RINx and DOUTx pins)
- ±15 kV using the Human Body Model (HBM) (on RINx and DOUTx pins)
- ±2 kV using the Human Body Model (HBM) (on all pins except RINx and DOUTx pins)
- ±0.5 kV using the Charged Device Model (CDM) (on all pins)



Feature Description (continued)

The IEC 61000-4-2 standard is more rigorous than HBM, resulting in lower voltage levels compared with HBM for the same level of ESD protection. Because IEC 61000-4-2 specifies a lower series resistance, the peak current is higher than HBM. The TRS3122E has passed both HBM and IEC 61000-4-2 testing.

8.3.5 Auto Powerdown Plus

Powerdown is engaged in two separate cases: automatically, when no activity has occurred for a period of time, and manually, using the FORCEOFF device pin.

8.3.5.1 Automatic Powerdown

Auto Powerdown Plus is enabled when FORCEON is set LOW and FORCEOFF is set HIGH. Using TRS3122E's integrated edge detection circuitry and timer, the device can sense when there is no activity on the driver or receiver inputs for 30 seconds. When this condition is sensed by the device, it automatically shuts the charge pump off, reducing supply current to 0.5 uA. When a valid transition is sensed on one of the driver or receiver inputs, the charge pump turns back on and TRS3122E exits powerdown. The typical time to exit powerdown is typically in 30 us, but can be as long as 150 us. As a result, the system saves power without requiring any software control. Device Functional Modes summarizes the operating modes in truth table form.

While in the low power mode with Automatic Powerdown enabled (FORCEOFF = HIGH and FORCEON = LOW), the receiver inputs are still enabled.

8.3.5.2 Manual Powerdown

The device can be manually powered down by externally setting FORCEOFF pin to low logic level. Both the drivers and receivers will be powered off. Device Functional Modes summarizes the operating modes in truth table form.

8.3.5.3 Forced On

If the FORCEOFF and FORCEON pins are both set HIGH, the device will power on with Auto Powerdown Plus disabled. Both the drivers and receiver will be active regardless of inactivity. Because powerdown is autonomous, FORCEON can be used ensure drivers are ready for new data transmission if the time since last transmission (or receive data) was more than 15 seconds. Device Functional Modes summarizes the operating modes in truth table form.



8.4 Device Functional Modes

8.4.1 Each Driver⁽¹⁾

		INPUTS		OUTPUT	
DIN	FORCEON	FORCEOFF	TIME ELAPSED SINCE LAST RIN OR DIN TRANSITION	DOUT	DRIVER STATUS
Х	Х	L	Х	Z	Powered off
L	Н	Н	Х	Н	Normal operation with
Н	Н	Н	Х	L	auto-powerdown plus disabled
L	L	Н	<30 s	Н	Normal operation with
Н	L	Н	<30 s	L	auto-powerdown plus enabled
L	L	Н	>30 s	Z	Powered off by
Н	L	Н	>30 s	Z	auto-powerdown plus feature

(1) H = high level, L = low level, X = irrelevant, Z = high impedance (off), 30s is typical inactivity time

8.4.2 Each Receiver⁽¹⁾

	INPL	JTS	OUTPUTS			
RIN	FORCEOFF	TIME ELAPSED SINCE LAST RIN OR DIN TRANSITION	ROUT	RECEIVER STATUS		
Х	L	Х	Z	Powered off		
L	Н	Х	Н	Normal operation with		
н	Н	Х	L	auto-powerdown plus		
Open	Н	Х	Н	disabled/enabled		

(1) H = high level, L = low level, X = irrelevant, Z = high impedance (off), Open = input disconnected or connected driver off

8.4.3 INVALID Status Truth Table⁽¹⁾

	INPUTS					
RIN1, RIN2	FORCEON	FORCEOFF	TIME ELAPSED SINCE LAST RIN OR DIN TRANSITION	INVALID		
Any L or H	Х	Х	Х	Н		
All Open	Х	Х	Х	L		

(1) H = high level, L = low level, X = irrelevant, Z = high impedance (off), Open = input disconnected or connected driver off

8.4.4 Capacitor Selection Table

Table 1. Capacitor Selection

$V_{CC} = V_{L}$	C1 Capacitor Value	C2 Capacitor Value	C3 Capacitor Value	C4 Capacitor Value	C5 Capacitor Value	
1.65 V to 2 V ⁽¹⁾		100 nF				
3.0 V to 3.6 V ⁽¹⁾	100) nF	100 nF or open	100 nF		
4.5 V to 5.5 V ⁽¹⁾	47 nF	330 nF	100 nF or open	330	nF	
3 V to 5.5 V ⁽²⁾	47 nF	470 nF	100 nF or open	470	nF	

(1) For optimized performance, we recommend using these configurations.

(2) For applications where the Vcc variation is larger, this configuration is acceptable.



9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

RS232 is used to communicate between two electrical units on separate PCBs across cables <40 ft. Common RS232 cables are RJ45, DB9 & DB25.

9.2 Typical 1.8-V Application



Figure 13. TRS3122E Typical Application

9.2.1 Design Requirements

For this design example, use the parameters listed in Table 2 as the input parameters.

Table 2. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE	
MCU GPIO Supple Voltage	1.8 V	
Transmission Voltage	+/-4.7 V	
Data-rate	1 Mbps	
Number of Transmitters / Receivers	2	
Charge Pump Capacitor Values	100nF (see Table 3)	

9.2.2 Detailed Design Procedure

When using TRS3122E, determine the following:

- All DIN, FORCEOFF, and FORCEON inputs must be connected to valid low or high logic levels.
- Select capacitor values based on V_{CC} level for best performance. (see Table 3)



9.2.2.1 Data-Rate and Cable Length

RS-232 intended is for short range data transmission. The rise time for RS-232 driver edges is slow enough that the data cable appears as a capacitor instead of a transmission line impedance. The elapsed time for one bit of data far exceeds the transit time of any practical RS-232 cable length. The capacitance of the cable is the limiting factor. Therefore the capacitance per foot (or meter) of the cable is important if long data cables are used. Capacitance slows the rise and fall time of the signal. For low data rates, the delay is insignificant. However, high data rates will have reduced percentage of time that the output is at V_{OL} or V_{OH} and more time in the transitions. The timing of the UART (universal asynchronous receiver/transmitter) must sample the signal at the right time to coincide with V_{OL} and V_{OH} plateaus. At some point data reliability will be impacted. There are no hard limits for cable capacitance and data rate.



Figure 14. Typical Waveform with Capacitive Load V_{CC} = 3.3 V, R_{LOAD} = 3 k Ω , Date Rate = 500kbps

The maximum cable length depends on the cable used (pf/ft), data rate, timing of receiving UART, system tolerance to data errors.

9.2.2.2 Capacitor Selection

The capacitor type used for C1–C5 is not critical for proper operation; polarized or non-polarized capacitors can be used, though lower ESR capacitors are preferred. The charge pump requires 0.1 μ F capacitors for V_{CC} = 1.8-V or V_{CC} = 3.3-V operation. For other supply voltages, see Table 1 for required capacitor values. Do not use values smaller than those listed in Table 1. Increasing the capacitor values(e.g., by a factor of 2), except for C1, reduces ripple on the transmitter outputs and slightly reduces power consumption. C2, C3, C4 and C5 can be increased without changing C1's value. However, do not increase C1 without also increasing the values of C2, C3, C4, C5, C_{BYPASS1}, and C_{BYPASS2} to maintain the proper ratios (C1 to the other capacitors). When using the minimum required capacitor values, make sure the capacitor value does not degrade excessively with temperature. If in doubt, use capacitors with a larger nominal value. The capacitor's equivalent series resistance (ESR) usually increases at low temperatures.

For best charge pump efficiency locate the charge pump and bypass capacitors as close as possible to the IC. Surface mount capacitors are best for this purpose. Using capacitors with lower equivalent series resistance (ESR) and self-inductance, along with minimizing parasitic PCB trace inductance will optimize charge pump operation. Designers are also advised to consider that capacitor values may shift over time and operating temperature.

$V_{CC} = V_{L}$	C1 Capacitor Value	C2 Capacitor Value	C3 Capacitor Value	C4 Capacitor Value	C5 Capacitor Value	
1.65 V to 2 V ⁽¹⁾			100 nF			
3.0 V to 3.6 V ⁽¹⁾	100) nF	100 nF or open	100 nF		
4.5 V to 5.5 V ⁽¹⁾	47 nF	330 nF	100 nF or open	330) nF	
3 V to 5.5 V ⁽²⁾	47 nF	470 nF	100 nF or open	470) nF	

Table	3.	Capacitor Selection
1 4 4 1 4	•••	Capacito: Colocitori

(1) For optimized performance, we recommend using these configurations.

(2) For applications where the Vcc variation is larger, this configuration is acceptable.

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9.2.3 Application Curves



Figure 15. 1 Mbps Eye Diagram, 2 V/div, 200 ns/ div V_{CC} = 1.8 V, C_{LOAD} = 500 pF, R_{LOAD} = 3 k Ω

10 Power Supply Recommendations

In most circumstances, a 0.1- μ F V_{CC} bypass capacitor and a 1- μ F V_L bypass capacitor are adequate. In applications that are sensitive to power-supply noise, use larger value V_{CC} bypass capacitor. There is no maximum limit for bypass capacitor. Place bypass capacitors as close to the IC as possible.

It is not recommended to use this device when V_{CC} is powered and $V_{L}= 0$ V or floating for an extended period of time because operation is undefined. V_{CC} and V_{L} must be powered to guarantee charge pump operation.

Also, to achieve full functionality as described in Specifications, it is recommended to not use a higher voltage on V_L than V_{CC} . Full functionality can be achieved when V_{CC} is greater than or equal to V_L .



11 Layout

11.1 Layout Guidelines

Minimize the length of all capacitor traces to ensure the device can maintain quick rising and falling transitions. Vias are recommended to accommodate layouts for the capacitors.

11.2 Layout Example



Figure 16. TRS3122E Typical Layout



12 Device and Documentation Support

12.1 Device Support

12.1.1 Third-Party Products Disclaimer

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12.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

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Design Support TI's Design Support Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.3 Trademarks

E2E is a trademark of Texas Instruments.

12.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



19-May-2016

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
TRS3122ERGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TRS3122	Samples
TRS3122ERGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TRS3122	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

19-May-2016

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TRS3122ERGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
TRS3122ERGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

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PACKAGE MATERIALS INFORMATION

19-May-2016



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TRS3122ERGER	VQFN	RGE	24	3000	367.0	367.0	35.0
TRS3122ERGET	VQFN	RGE	24	250	210.0	185.0	35.0

MECHANICAL DATA



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-Leads (QFN) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions. F. Falls within JEDEC MO-220.
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RGE (S-PVQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTES: A. All linear dimensions are in millimeters



RGE (S-PVQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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