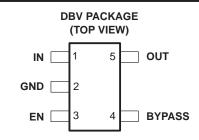
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- 150-mA Low Noise, Low-Dropout Regulator
- Output Voltage: 2.5 V, 2.7 V, 2.8 V, 3.0 V, 3.3 V
- Output Noise Typically 50 μV
- Quiescent Current Typically 85 µA
- Dropout Voltage, Typically 300 mV at 150 mA
- **Thermal Protection**
- **Over Current Limitation**
- Less Than 2-µA Quiescent Current in **Shutdown Mode**
- -40°C to 125°C Operating Junction **Temperature Range**
- 5-Pin SOT-23 (DBV) Package



## description

The TPS764xx family of low-dropout (LDO) voltage regulators offers the benefits of a low noise, low-dropout voltage, low-power operation, and miniaturized package. Additionally, they feature low quiescent current when compared to conventional LDO regulators. Offered in 5-terminal small outline integrated-circuit SOT-23 package, the TPS764xx series devices are ideal for low-noise applications, cost-sensitive designs and applications where board space is at a premium.

A combination of new circuit design and process innovation has enabled the usual pnp pass transistor to be replaced by a PMOS pass element. Because the PMOS pass element behaves as a low-value resistor, the dropout voltage is very low-typically 300 mV at 150 mA of load current (TPS76433)-and is directly proportional to the load current. Since the PMOS pass element is a voltage-driven device, the guiescent current is very low (140 μA maximum) and is stable over the entire range of output load current (0 mA to 150 mA). Intended for use in portable systems such as laptops and cellular phones, the low-dropout voltage feature and low-power operation result in a significant increase in system battery operating life.

The TPS764xx also features a logic-enabled sleep mode to shut down the regulator, reducing quiescent current to 1  $\mu$ A maximum at  $T_{.1}$  = 25°C.The TPS764xx is offered in 2.5-V, 2.7-V, 2.8-V, 3.0-V, and 3.3-V fixed-voltages.

#### **AVAILABLE OPTIONS**

TJ	VOLTAGE	PACKAGE	PART N	SYMBOL	
	2.5 V		TPS76425DBVT†	TPS76425DBVR‡	PBJI
	2.7 V	SOT-23 (DBV)	TPS76427DBVT <sup>†</sup>	TPS76427DBVR‡	PBKI
–40°C to 125°C	2.8 V		TPS76428DBVT <sup>†</sup>	TPS76428DBVR <sup>‡</sup>	PCEI
	3.0 V		TPS76430DBVT <sup>†</sup>	TPS76430DBVR <sup>‡</sup>	PBLI
	3.3 V		TPS76433DBVT <sup>†</sup>	TPS76433DBVR‡	PBMI

<sup>&</sup>lt;sup>†</sup> The DBVT passive indicates tape and reel of 250 parts.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

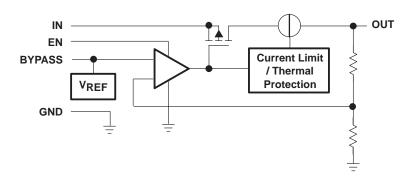


<sup>&</sup>lt;sup>‡</sup> The DBVR passive indicates tape and reel of 3000 parts.

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## functional block diagram

#### TPS76425/27/28/30/33



## **Terminal Functions**

TERMINAL NAME	1/0	DESCRIPTION
GND		Ground
EN	I	Enable input
BYPASS		Output bypass capacitor
IN	I	Input supply voltage
OUT	0	Regulated output voltage

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# absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Input voltage range (see Note 1)	0.3 V to 10 V
Voltage range at EN	0.3 V to V <sub>I</sub> + 0.3 V
Voltage on OUT,	
Peak output current	Internally limited
ESD rating, HBM	2 kV
Continuous total power dissipation	See dissipation rating tables
Operating virtual junction temperature range, T <sub>J</sub>	–40°C to 150°C
Storage temperature range, T <sub>stg</sub>	65°C to 150°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to network ground terminal.

#### **DISSIPATION RATING TABLE**

	PACKAGE	$T_{\mbox{A}} \le 25^{\circ}\mbox{C}$ POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 85°C POWER RATING
Recommended	DBV	350 mW	3.5 mW/°C	192 mW	140 mW
Maximum	DBV	437 mW	3.5 mW/°C	280 mW	227 mW

## recommended operating conditions

	MIN	NOM M	AX	UNIT
Input voltage, V <sub>I</sub> †	2.7		10	V
Continuous output current, IO	0		50	mA
Operating junction temperature, T <sub>J</sub>	-40		25	°C

<sup>†</sup> To calculate the minimum input voltage for your maximum output current, use the following equation:  $V_{I(min)} = V_{O(max)} + V_{DO(max load)}$ 



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# electrical characteristics over recommended operating free-air temperature range, $V_I = V_{O(typ)} + V_{O(typ)}$ 1 V, $I_O$ = 1 mA, EN = IN, $C_O$ = 4.7 $\mu$ F (unless otherwise noted)

PARAMETER			TEST CONDITIONS		MIN	TYP	MAX	UNIT	
			I <sub>O</sub> = 1 mA to 100 mA,	T <sub>J</sub> = 25°C	2.45	2.5	2.55		
		TPS76425	I <sub>O</sub> = 1 mA to 100 mA		2.425	2.5	2.575	V	
		11570425	I <sub>O</sub> = 1 mA to 150 mA,	T <sub>J</sub> = 25°C	2.438	2.5	2.562	V	
			I <sub>O</sub> = 1 mA to 150 mA		2.407	2.5	2.593		
			$I_O = 1 \text{ mA to } 100 \text{ mA},$	T <sub>J</sub> = 25°C	2.646	2.7	2.754		
		TPS76427	I <sub>O</sub> = 1 mA to 100 mA	_	2.619	2.7	2.781	V	
		1175/1042/	I <sub>O</sub> = 1 mA to 150 mA,	T <sub>J</sub> = 25°C	2.632	2.7	2.768	V	
			I <sub>O</sub> = 1 mA to 150 mA		2.598	2.7	2.8013		
			I <sub>O</sub> = 1 mA to 100 mA	T <sub>J</sub> = 25°C	2.744	2.8	2.856		
V <sub>O</sub>	Output voltage	TPS76428	I <sub>O</sub> = 1 mA to 150 mA,		2.73	2.8	2.870	V	
'0	Output voltage	117370420	I <sub>O</sub> = 1 mA to 150 mA,	T <sub>J</sub> = 25°C	2.716	2.8	2.884		
			I <sub>O</sub> = 1 mA to 150 mA		2.695	2.8	2.905		
		TPS76430	$I_O = 1 \text{ mA to } 100 \text{ mA},$	$T_J = 25^{\circ}C$	2.94	3.0	3.06	V	
			I <sub>O</sub> = 1 mA to 100 mA		2.925	3.0	3.075		
			$I_O = 1 \text{ mA to } 150 \text{ mA},$	T <sub>J</sub> = 25°C	2.91	3.0	3.090		
			I <sub>O</sub> = 1 mA to 150 mA		2.887	3.0	3.112		
		TPS76433	$I_O = 1 \text{ mA to } 100 \text{ mA},$	T <sub>J</sub> = 25°C	3.234	3.3	3.366		
			I <sub>O</sub> = 1 mA to 100 mA		3.201	3.3	3.399	V	
			$I_O = 1 \text{ mA to } 150 \text{ mA},$	T <sub>J</sub> = 25°C	3.218	3.3	3.382	V	
			$I_O = 1 \text{ mA to } 150 \text{ mA}$		3.177	3.3	3.423		
I <sub>(Q)</sub>	Quiescent current	ant)	I <sub>O</sub> = 0 to 150 mA, See Note 2	$T_J = 25^{\circ}C$ ,		85	100		
	(GND terminal current)		I <sub>O</sub> = 0 to 150 mA,	See Note 2			140	μА	
	Standby ourrant		EN < 0.5 V,	T <sub>J</sub> = 25°C		0.5	1		
Standby current			EN < 0.5 V				2		
Vn	Output noise voltage		BW = 300 Hz to 50 kHz, $T_J = 25^{\circ}C$ ,	$C_0 = 10 \mu F$ , See Note 2		50		μV	
	Bypass voltage		T <sub>J</sub> = 25°C			1.192		V	
PSRR	Ripple rejection		$f = 1 \text{ kHz}, \ C_0 = 10 \ \mu\text{F},$	T <sub>J</sub> = 25°C, See Note 2		60		dB	
	Current limit	· ·	T <sub>J</sub> = 25°C	See Note 3		0.8	1.5	Α	

NOTES: 2. Minimum IN operating voltage is 2.7 V or  $V_{O(typ)}$  + 1 V, whichever is greater. 3. Test condition includes, output voltage  $V_{O}$ =0 V and pulse duration = 10 mS.

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#### electrical characteristics over recommended operating free-air temperature range, $V_I = V_{O(tvp)} + 1 V$ , $I_{O} = 1 \text{ mA}$ , EN = IN, $C_O = 4.7 \mu\text{F}$ (unless otherwise noted) (continued)

	PARAMETER		TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
	Output voltage line regulation (ΔV <sub>O</sub> /V <sub>O</sub> )		$V_{O} + 1 V < V_{I} \le 10 V$	V <sub>I</sub> ≥ 3.5 V, T <sub>J</sub> = 25°C		0.04	0.07	%/V
	(see Note 4)		$V_{O} + 1 V < V_{I} \le 10 V$ ,	V <sub>I</sub> ≥ 3.5 V			0.1	70/ V
٧ <sub>IH</sub>	EN high level input		See Note 2			1.4	2	V
$V_{IL}$	EN low level input		See Note 2		0.5	1.2		V
1.	EN input current		EN = 0 V			-0.01	-0.5	μΑ
l'I	Liv input current		EN = IN			-0.01	-0.5	μΑ
			$I_O = 0 \text{ mA},$	T <sub>J</sub> = 25°C		0.2		
			$I_O = 1 \text{ mA},$	T <sub>J</sub> = 25°C		3		
			$I_0 = 50 \text{ mA},$	T <sub>J</sub> = 25°C		120	150	
			I <sub>O</sub> = 50 mA				200	mV
		TPS76425	$I_{O} = 75 \text{ mA},$	T <sub>J</sub> = 25°C		180	225	
		11-370423	I <sub>O</sub> = 75 mA				300	
			I <sub>O</sub> = 100 mA,	T <sub>J</sub> = 25°C		240	300	
			I <sub>O</sub> = 100 mA				400	
			I <sub>O</sub> = 150 mA,	T <sub>J</sub> = 25°C		360	450	
\/ <b>&gt;</b>	Dropout voltage		I <sub>O</sub> = 150 mA				600	
VDO	(Note 5)		$I_O = 0 \text{ mA},$	T <sub>J</sub> = 25°C		0.2		
			$I_O = 1 \text{ mA},$	T <sub>J</sub> = 25°C		3		
			$I_O = 50 \text{ mA},$	T <sub>J</sub> = 25°C		100	125	
			I <sub>O</sub> = 50 mA				166	
		TPS76433	I <sub>O</sub> = 75 mA,	T <sub>J</sub> = 25°C		150	188	
		17370433	I <sub>O</sub> = 75 mA				250	mV
			I <sub>O</sub> = 100 mA,	T <sub>J</sub> = 25°C		200	250	
			I <sub>O</sub> = 100 mA				333	
			I <sub>O</sub> = 150 mA,	T <sub>J</sub> = 25°C		300	375	
			I <sub>O</sub> = 150 mA				500	

NOTES: 2. Minimum IN operating voltage is 2.7 V or  $V_{O(typ)}$  + 1 V, whichever is greater. 4. If  $V_{O}$  < 2.5 V and  $V_{imax}$  = 10 V,  $V_{imin}$  = 3.5 V:

Line Reg. (mV) = 
$$(\%/V) \times \frac{V_O(V_{imax} - 3.5 \text{ V})}{100} \times 1000$$

If  $V_O > 2.5 \text{ V}$  and  $V_{imax} = 10 \text{ V}$ ,  $V_{imin} = V_O + 1 \text{ V}$ :

$$v_{max} = 10 \text{ V, V}_{imin} = V_O + 1 \text{ V:}$$
Line Reg. (mV) =  $(\%/\text{V}) \times \frac{V_O(V_{imax} - (V_O + 1))}{100} \times 1000$ 

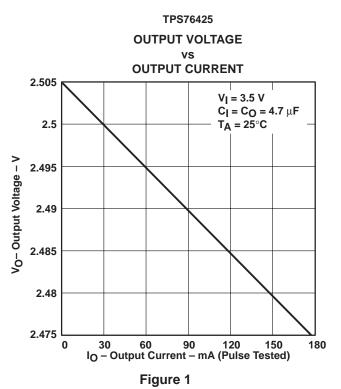
5. Dropout voltage is defined as the differential voltage between VO and Vi when VO drops 100 mV below the value measured with  $V_i = V_O + 1.0 V .:$ 

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#### TYPICAL CHARACTERISTICS

## **Table of Graphs**

			FIGURE
V	Custout voltage	vs Output current	1
Vo	Output voltage	vs Free-air temperature	2, 3, 4
V <sub>n</sub>	Output noise	vs Frequency	5
Vn	Output paige valtage	vs Bypass capacitance	6
	Output noise voltage	vs Load current	7
Zo	Output impedance	vs Frequency	8
$V_{DO}$	Dropout voltage	vs Free-air temperature	9
	Ripple rejection	vs Frequency	10
	Line transient response		11, 13
	Load transient response		12, 14
	Compensation series resistance (CSR)	vs Output current	15, 17
	Compensation series resistance (CSR)	vs Added ceramic capacitance	16, 18



### **TPS76425 OUTPUT VOLTAGE** vs FREE-AIR TEMPERATURE 2.53 $V_{I} = 3.5 V$ $C_{I} = C_{O} = 4.7 \mu F$ 2.52 Vo- Output Voltage - V 2.51 $I_0 = 1 \text{ mA}$ 2.5 I<sub>O</sub> = 150 mA 2.49 2.48 2.47 **-55 -35** -15 25 45 65 85 105 125 $T_A$ – Free-Air Temperature – $^{\circ}$ C Figure 2

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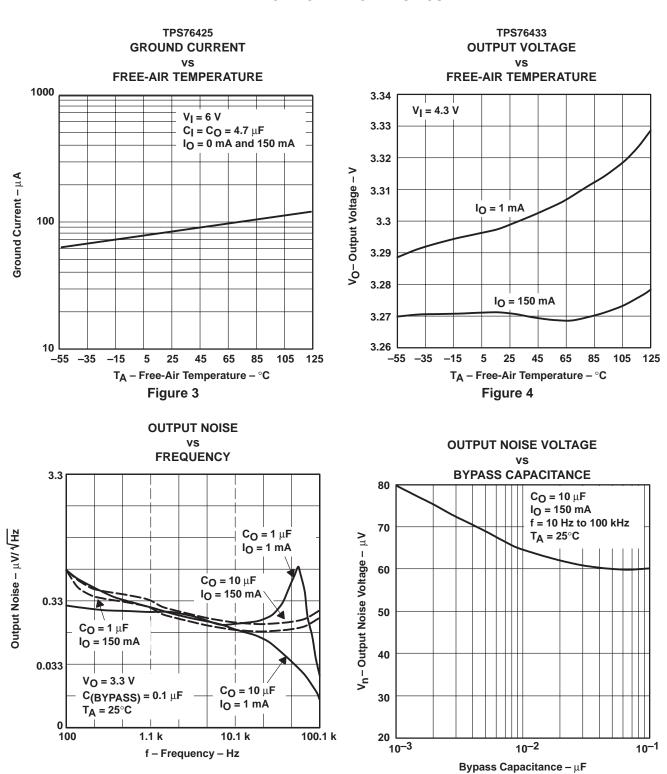




Figure 5

Figure 6

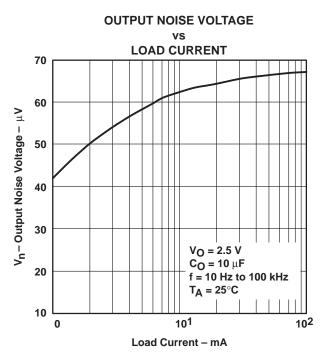
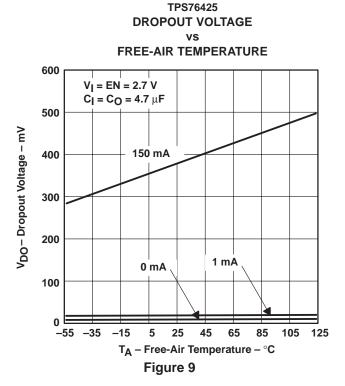
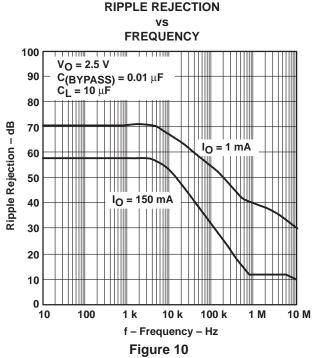


Figure 7



**OUTPUT IMPEDANCE FREQUENCY** 10  $Z_{o}-$  Output Impedance  $-\Omega$  $I_0 = 1 \text{ mA}$ I<sub>O</sub> = 150 mA  $C_{I} = C_{O} = 4.7 \mu F$  $ESR = 1 \Omega$ T<sub>A</sub> = 25°C 0.1 0.01 0.1 10 100 1000 f - Frequency - kHz Figure 8



**TPS76425** 

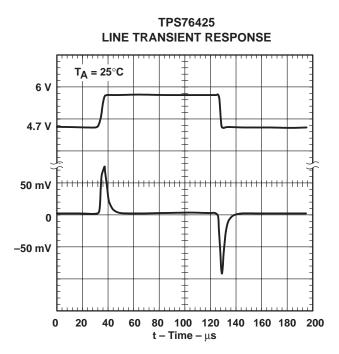


Figure 11

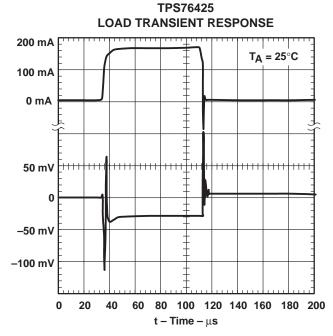


Figure 12

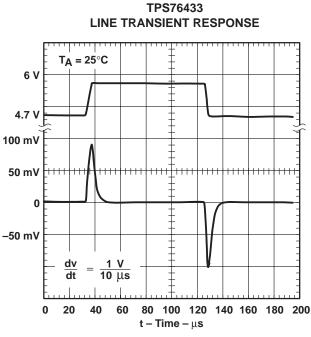
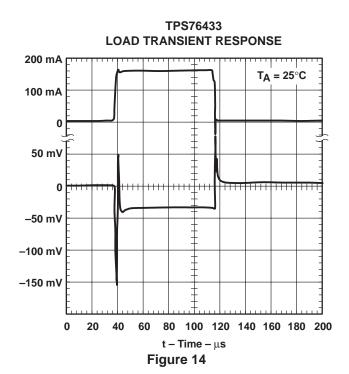
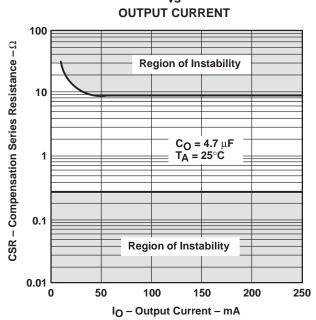


Figure 13

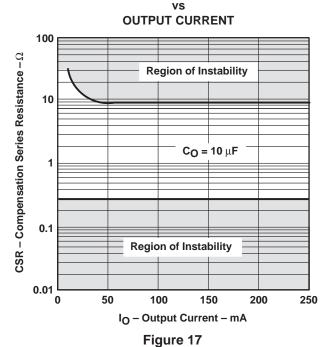


# TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR)†

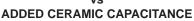


TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR)†

Figure 15



# TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR)†



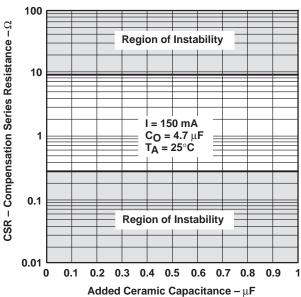


Figure 16

## TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE (CSR)†

#### ADDED CERAMIC CAPACITANCE

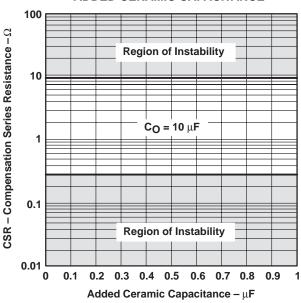


Figure 18

<sup>†</sup> CSR refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.



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#### **APPLICATION INFORMATION**

The TPS764xx family of low-noise and low-dropout (LDO) regulators are optimized for use in battery-operated equipment. They feature extremely low noise (50  $\mu$ V), low dropout voltages, low quiescent current (140  $\mu$ A), and an enable input to reduce supply current to less than 2  $\mu$ A when the regulator is turned off.

#### device operation

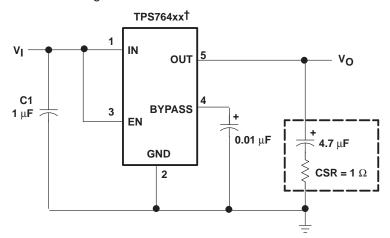
The TPS764xx uses a PMOS pass element to dramatically reduce both dropout voltage and supply current over more conventional PNP-pass-element LDO designs. The PMOS pass element is a voltage-controlled device which, unlike a PNP transistor, does not require increased drive current as output current increases. Supply current in the TPS764xx is essentially constant from no-load to maximum load.

Current limiting and thermal protection prevent damage by excessive output current and/or power dissipation. The device switches into a constant-current mode at approximately 1 A; further load reduces the output voltage instead of increasing the output current. The thermal protection shuts the regulator off if the junction temperature rises above 165°C. Recovery is automatic when the junction temperature drops approximately 25°C below the high temperature trip point. The PMOS pass element includes a back diode that safely conducts reverse current when the input voltage level drops below the output voltage level.

An internal resistor, in conjunction with external 0.01- $\mu$ F bypass capacitor, creates a low-pass filter to further reduce the noise. The TPS764xx exhibits only 50  $\mu$ V of output voltage noise using 0.01  $\mu$ F bypass and 4.7- $\mu$ F output capacitors.

A logic low on the enable input, EN, shuts off the output and reduces the supply current to less than 2  $\mu$ A. EN should be tied high in applications where the shutdown feature is not used.

A typical application circuit is shown in Figure 22.



† TPS76425, TPS76427 TPS76430, TPS76433.

Figure 19. Typical Application Circuit



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#### **APPLICATION INFORMATION**

#### external capacitor requirements

Although not required, a 0.047- $\mu$ F or larger ceramic bypass input capacitor, connected between IN and GND and located close to the TPS764xx, is recommended to improve transient response and noise rejection. A higher-value electrolytic input capacitor may be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

Like all low dropout regulators, the TPS764xx requires an output capacitor connected between OUT and GND to stabilize the internal loop control. The minimum recommended capacitance value is 4.7  $\mu F$  and the ESR (equivalent series resistance) must be between 0.2  $\Omega$  and 10  $\Omega$ . Capacitor values 4.7  $\mu F$  or larger are acceptable, provided the ESR is less than 10  $\Omega$ . Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described above. Most of the commercially available 4.7  $\mu F$  surface—mount solid tantalum capacitors, including devices from Sprague, Kemet, and Nichico, meet the ESR requirements previously stated. Multilayer ceramic capacitors should have minimum values of 1  $\mu F$  over the full operating temperature range of the equipment.

#### **CAPACITOR SELECTION**

PART NO.	MFR.	VALUE	MAX ESR <sup>†</sup>	SIZE $(H \times L \times W)^{\dagger}$
T494B475K016AS	KEMET	4.7 μF	1.5 Ω	$1.9 \times 3.5 \times 2.8$
195D106x0016x2T	SPRAGUE	10 μF	1.5 Ω	$1.3\times7.0\times2.7$
695D106x003562T	SPRAGUE	10 μF	1.3 Ω	$2.5\times7.6\times2.5$
TPSC475K035R0600	AVX	4.7 μF	$0.6\Omega$	$2.6\times6.0\times3.2$

<sup>†</sup> Size is in mm. ESR is maximum resistance in ohms at 100 kHz and T<sub>A</sub> = 25°C. Listings are sorted by height.



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#### **APPLICATION INFORMATION**

## power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of  $125^{\circ}$ C; the maximum junction temperature allowable without damaging the device is  $150^{\circ}$ C. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation,  $P_{D(max)}$ , and the actual dissipation,  $P_{D}$ , which must be less than or equal to  $P_{D(max)}$ .

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(max)} = \frac{T_J max - T_A}{R_{\theta,JA}}$$

Where

T<sub>J</sub>max is the maximum allowable junction temperature

R<sub>0</sub>JA is the thermal resistance junction-to-ambient for the package, i.e., 285°C/W for the 5-terminal SOT23.

T<sub>A</sub> is the ambient temperature.

The regulator dissipation is calculated using:

$$P_{D} = (V_{I} - V_{O}) \times I_{O}$$

Power dissipation resulting from quiescent current is negligible.

## regulator protection

The TPS764xx pass element has a built-in back diode that safely conducts reverse current when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. If extended reverse voltage is anticipated, external limiting might be appropriate.

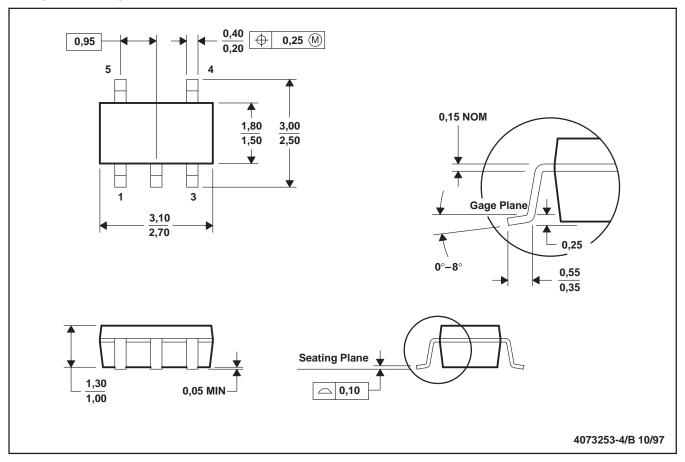
The TPS764xx also features internal current limiting and thermal protection. During normal operation, the TPS764xx limits output current to approximately 800 mA. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 165°C, thermal-protection circuitry shuts it down. Once the device has cooled down to below 140°C, regulator operation resumes.

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#### **MECHANICAL DATA**

## DBV (R-PDSO-G5)

#### PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions include mold flash or protrusion.

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