

The TPS735xx family of low-dropout (LDO), low-power linear regulators offers excellent ac

performance with very low ground current. High

power-supply rejection ratio (PSRR), low noise, fast

start-up, and excellent line and load transient

response are provided while consuming a very low 46µA (typical) ground current. The TPS735xx is

stable with ceramic capacitors and uses an advanced

BiCMOS fabrication process to yield a typical dropout

voltage of 250mV at 500mA output. The TPS735xx

uses a precision voltage reference and feedback loop to achieve overall accuracy of 2% ($V_{OUT} > 2.2V$) over

all load, line, process, and temperature variations. It

is fully specified from $T_J = -40^{\circ}C$ to $+125^{\circ}C$ and is

offered in a low-profile, 2mm x 2mm SON and 3mm x

3mm SON packages that is ideal for wireless

handsets, printers, and WLAN cards.

500mA, Low Quiescent Current, Ultra-Low Noise, High PSRR Low Dropout Linear Regulator

DESCRIPTION

FEATURES

- 500mA Low Dropout Regulator with EN
- Low I_Q: 46μA
- Multiple Output Voltage Versions Available:
 - Fixed Outputs of 1.0V to 4.3V Using Innovative Factory EEPROM Programming
 Adjustable Outputs from 1.25V to 6.0V
- High PSRR: 60dB at 1kHz
- Ultra-low Noise: 28μV_{RMS}
- Fast Start-Up Time: 45µs
- Stable with a Low-ESR, 2.0μF Typical Output Capacitance
- Excellent Load/Line Transient Response
- 2% Overall Accuracy (Load/Line/Temp, V_{OUT} > 2.2V)
- Very Low Dropout: 250mV at 500mA
- 2mm × 2mm SON-6 and 3mm × 3mm SON-8 Packages

APPLICATIONS

- WiFi, WiMax
- Printers
- Cellular Phones, SmartPhones
- Handheld Organizers, PDAs

DRB PACKAGE 3mm x 3mm SON (TOP VIEW) OUT 8 IN 1 N/C 2 7 N/C GND NR/FB 3 6 N/C 4 GND 5 EN

DRV PACKAGE 2mm x 2mm SON (TOP VIEW)						
OUT NR/FB GND	2	GND	6 5 4	IN N/C EN		



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ORDERING INFORMATION⁽¹⁾

PRODUCT	V _{OUT} ⁽²⁾
	 XX is nominal output voltage (for example, 28 = 2.8V, 285 = 2.85V, 01 = Adjustable). YYY is package designator. Z is package quantity.

For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

(2) Output voltages from 1.0V to 3.6V in 50mV increments are available through the use of innovative factory EEPROM programming; minimum order quantities may apply. Contact factory for details and availability.

ABSOLUTE MAXIMUM RATINGS

Over operating temperature range (unless otherwise noted).⁽¹⁾

PARAMETER	TPS735xx	UNIT
V _{IN} range	-0.3 to +7.0	V
V _{EN} range	–0.3 to V _{IN} +0.3	V
V _{OUT} range	–0.3 to V _{IN} +0.3	V
V _{FB} range	–0.3 to V _{FB} (TYP) +0.3	V
Peak output current	Internally limited	
Continuous total power dissipation	See Dissipation Ratir	ngs Table
Junction temperature range, T _J	-55 to +150	°C
Storage temperature range , T _{STG}	-55 to +150	°C
ESD rating, HBM	2	kV
ESD rating, CDM	500	V

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

DISSIPATION RATINGS

BOARD	PACKAGE	$R_{\theta JC}$	$R_{ heta JA}$	DERATING FACTOR ABOVE T _A = +25°C	T _A < +25°C	T _A = +70°C	T _A = +85°C
Low-K ⁽¹⁾	DRV	20°C/W	140°C/W	7.1mW/°C	715mW	395mW	285mW
High-K ⁽²⁾	DRV	20°C/W	65°C/W	15.4mW/°C	1.54W	845mW	615mW
High-K ⁽²⁾⁽³⁾	DRB	1.2°C/W	40°C/W	25mW/°C	2.5W	1.38W	1.0W

(1) The JEDEC low-K (1s) board used to derive this data was a 3in x 3in (7,62cm x 7,62cm), two-layer board with 2-ounce (56,699g) copper traces on top of the board.

(2) The JEDEC high-K (2s2p) board used to derive this data was a 3in x 3in (7,62cm x 7,62cm), multilayer board with 1-ounce (28,35g) internal power and ground planes and 2-ounce (56,699g) copper traces on top and bottom of the board.

(3) The R_{BUC} value of the DRB package is junction-to-pad; note that this is not junction-to-case (top center of IC package).



ELECTRICAL CHARACTERISTICS

Over operating temperature range (T_J = -40°C to +125°C), V_{IN} = V_{OUT(TYP)} + 0.5V or 2.7V, whichever is greater; I_{OUT} = 1mA, V_{EN} = V_{IN}, C_{OUT} = 2.2 μ F, C_{NR} = 0.01 μ F, unless otherwise noted. For TPS73501, V_{OUT} = 3.0V. Typical values are at $T_{\perp} = +25^{\circ}C$.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IN}	Input voltage range ⁽¹⁾			2.7		6.5	V
V _{FB}	Internal reference (TPS735	501)		1.184	1.208	1.232	V
V _{OUT}	Output voltage range (TPS	73501)		V _{FB}		6.0	V
V _{OUT}	Output accuracy	Nominal	$T_J = +25^{\circ}C$	-1.0		+1.0	%
V	Output accuracy ⁽¹⁾	Over V _{IN} ,	$\label{eq:Vout} \begin{array}{l} V_{OUT} + 0.3V \leq V_{IN} \leq 6.5V \\ 1mA \leq I_{OUT} \leq 500mA, \ V_{OUT} > 2.2V \end{array}$	-2.0	±1.0	+2.0	%
V _{OUT}		I _{OUT} , Temp	$ \begin{array}{l} V_{OUT} + 0.3V \leq V_{IN} \leq 6.5V \\ 1mA \leq I_{OUT} \leq 500mA, \ V_{OUT} \leq 2.2V \end{array} $	-3.0	±1.0	+3.0	%
$\Delta V_{OUT}\%/\Delta V_{IN}$	Line regulation ⁽¹⁾		$V_{OUT(NOM)} + 0.3V \le V_{IN} \le 6.5V$		0.02		%/V
ΔV _{OUT} %/ ΔI _{OUT}	Load regulation		$500\mu A \le I_{OUT} \le 500 mA$		0.005		%/mA
V _{DO}	Dropout voltage ⁽²⁾ ($V_{IN} = V_{OUT(NOM)} - 0.1V$)		I _{OUT} = 500mA		250	500	mV
I _{CL}	Output current limit		$V_{OUT} = 0.9 \times V_{OUT(NOM)}$	800	1170	1720	mA
I _{GND}	Ground pin current		$500\mu A \le I_{OUT} \le 500mA$		45	65	μA
I _{SHDN}	Shutdown current (I _{GND})		$V_{\sf EN} \le 0.4 V$		0.15	1.0	μA
I _{FB}	Feedback pin current (TPS	73501)		-0.5		0.5	μΑ
PSRR V			f = 100Hz		60		dB
	Power-supply rejection ratio $V_{IN} = 3.85V, V_{OUT} = 2.85V,$		f = 1kHz		56		dB
	$C_{NR} = 0.01 \mu F$, $I_{OUT} = 100 m$, 1A	f = 10kHz		41		dB
			f = 100kHz		28		dB
V _N	Output noise voltage		$C_{NR} = 0.01 \mu F$	1	1 x V _{OUT}		μV _{RM}
۷N	BW = 10Hz to 100kHz, V_{OI}	_{JT} = 2.8V	C _{NR} = none	95 x V _{OUT}			μV_{RMS}
	Startup time, V _{OUT} = 0% to		C _{NR} = none		45		μs
т	90%		$C_{NR} = 0.001 \mu F$		45		μs
T _{STR}	V _{OUT} = 2.85V, R _L = 14Ω, C _{OUT} = 2.2μF		$C_{NR} = 0.01 \mu F$	50			μs
	$R_{L} = 14\Omega, C_{OUT} = 2.2\mu r$		$C_{NR} = 0.047 \mu F$		50		μs
V _{EN(HI)}	Enable high (enabled)			1.2		V _{IN}	V
V _{EN(LO)}	Enable low (shutdown)			0		0.4	V
I _{EN(HI)}	Enable pin current, enable	d	$V_{EN} = V_{IN} = 6.5V$		0.03	1.0	μA
Ŧ			Shutdown, temperature increasing		165		°C
T _{SD}	Thermal shutdown tempera	aure	Reset, temperature decreasing		145		°C
TJ	Operating junction tempera	ature		-40		+125	°C
	Under-voltage lock-out		V _{IN} rising	1.90	2.20	2.65	V
UVLO	Hysteresis		V _{IN} falling		70		mV

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DEVICE INFORMATION

FUNCTIONAL BLOCK DIAGRAMS



Figure 1. Fixed Voltage Versions



PIN CONFIGURATIONS





PIN DESCRIPTIONS

	TPS735xx		
NAME	DRV	DRB	DESCRIPTION
IN	6	8	Input supply.
GND	3, Pad	4	Ground. The pad must be tied to GND.
EN	4	5	Driving the enable pin (EN) high turns on the regulator. Driving this pin low puts the regulator into shutdown mode. EN can be connected to IN if not used.
NR	2	3	Fixed voltage versions only; connecting an external capacitor to this pin bypasses noise generated by the internal bandgap. This allows output noise to be reduced to very low levels.
FB	2	3	Adjustable version only; this is the input to the control loop error amplifier, and is used to set the output voltage of the device.
OUT	1	1	Output of the regulator. A small capacitor (total typical capacitance $\ge 2.0\mu$ F ceramic) is needed from this pin to ground to assure stability.
N/C	5	2, 6, 7	Not internally connected. This pin must either be left open, or tied to GND.

TPS735xx



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TYPICAL CHARACTERISTICS

Over operating temperature range (T_J = -40°C to +125°C), $V_{IN} = V_{OUT(TYP)}$ + 0.5V or 2.7V, whichever is greater; I_{OUT} = 1mA, $V_{EN} = V_{IN}$, C_{OUT} = 2.2 μ F, C_{NR} = 0.01 μ F, unless otherwise noted. For TPS73501, V_{OUT} = 2.8V. Typical values are at T_J = +25°C.



TYPICAL CHARACTERISTICS (continued)

Over operating temperature range (T_J = -40°C to +125°C), $V_{IN} = V_{OUT(TYP)} + 0.5V$ or 2.7V, whichever is greater; $I_{OUT} = 1$ mA, $V_{EN} = V_{IN}$, $C_{OUT} = 2.2\mu$ F, $C_{NR} = 0.01\mu$ F, unless otherwise noted. For TPS73501, $V_{OUT} = 2.8V$. Typical values are at $T_J = +25$ °C.



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TYPICAL CHARACTERISTICS (continued)

Over operating temperature range (T_J = -40°C to +125°C), $V_{IN} = V_{OUT(TYP)}$ + 0.5V or 2.7V, whichever is greater; I_{OUT} = 1mA, $V_{EN} = V_{IN}$, C_{OUT} = 2.2 μ F, C_{NR} = 0.01 μ F, unless otherwise noted. For TPS73501, V_{OUT} = 2.8V. Typical values are at T_J = +25°C.





APPLICATION INFORMATION

The TPS735xx family of LDO regulators combines the high performance required of many RF and precision analog applications with ultra-low current consumption. High PSRR is provided by a high gain, high bandwidth error loop with good supply rejection at very low headroom (V_{IN} - V_{OUT}). Fixed voltage versions provide a noise reduction pin to bypass noise generated by the bandgap reference and to improve PSRR while a quick-start circuit fast-charges this capacitor at startup. The combination of high performance and low ground current also make the TPS735xx an excellent choice for portable applications. All versions have thermal and over-current protection and are fully specified from -40°C to +125°C.

Figure 20 shows the basic circuit connections for fixed voltage models. Figure 21 gives the connections for the adjustable output version (TPS73501). R_1 and R_2 can be calculated for any output voltage using the formula in Figure 21.



Figure 20. Typical Application Circuit for Fixed Voltage Versions



Figure 21. Typical Application Circuit for Adjustable Voltage Versions

Input and Output Capacitor Requirements

Although an input capacitor is not required for stability, it is good analog design practice to connect a 0.1µF to 1µF low equivalent series resistance (ESR) capacitor across the input supply near the regulator. The ground of this capacitor should be connected as close as the ground of output capacitor; a capacitor value of 0.1μ F is enough in this condition. When it is difficult to place these two ground points close together, a 1µF capacitor is recommended. This capacitor counteracts reactive input sources and improves transient response, noise rejection, and ripple rejection. A higher-value capacitor may be necessary if large, fast rise-time load transients are anticipated, or if the device is located several inches from the power source. If source impedance is not sufficiently low, a 0.1µF input capacitor may be necessary to ensure stability.

The TPS735xx is designed to be stable with standard ceramic output capacitors of values 2.2μ F or larger. X5R and X7R type capacitors are best because they have minimal variation in value and ESR over temperature. Maximum ESR of the output capacitor should be < 1.0Ω , so output capacitor type should be either ceramic or conductive polymer electrolytic.

Feedback Capacitor Requirements (TPS73501 only)

The feedback capacitor, C_{FB} , shown in Figure 21 is required for stability. For a parallel combination of R_1 and R_2 equal to 250k Ω , any value from 3pF to 1nF can be used. Fixed voltage versions have an internal 30pF feedback capacitor that is quick-charged at start-up. The adjustable version does not have this quick-charge circuit, so values below 5pF should be used to ensure fast startup; values above 47pF can be used to implement an output voltage soft-start. Larger value capacitors also improve noise slightly. The TPS73501 is stable in unity-gain configuration (OUT tied to FB) without C_{FB} .

Output Noise

In most LDOs, the bandgap is the dominant noise source. If a noise reduction capacitor (C_{NR}) is used with the TPS735xx, the bandgap does not contribute significantly to noise. Instead, noise is dominated by the output resistor divider and the error amplifier input. To minimize noise in a given application, use a 0.01µF noise reduction capacitor; for the adjustable version, smaller value resistors in the output resistor divider reduce noise. A parallel combination that gives 2µA of divider current has the same noise performance as a fixed voltage version. To further



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As with any linear regulator, PSRR and transient response are degraded as $(V_{IN} - V_{OUT})$ approaches dropout. This effect is shown in the Typical Characteristics section.

Startup and Noise Reduction Capacitor

Fixed voltage versions of the TPS735xx use a quick-start circuit to fast-charge the noise reduction capacitor, C_{NR} , if present (see the Functional Block Diagrams). This architecture allows the combination of very low output noise and fast start-up times. The NR pin is high impedance so a low leakage C_{NR} capacitor must be used; most ceramic capacitors are appropriate in this configuration.

Note that for fastest startup, V_{IN} should be applied first, then the enable pin (EN) driven high. If EN is tied to IN, startup is somewhat slower. Refer to the Typical Characteristics section. The quick-start switch is closed for approximately 135 μ s. To ensure that C_{NR} is fully charged during the quick-start time, a 0.01 μ F or smaller capacitor should be used.

Transient Response

As with any regulator, increasing the size of the output capacitor reduces over/undershoot magnitude but increases duration of the transient response. In the adjustable version, adding C_{FB} between OUT and FB improves stability and transient response. The transient response of the TPS735xx is enhanced by an active pull-down that engages when the output overshoots by approximately 5% or more when the device is enabled. When enabled, the pull-down device behaves like a 400 Ω resistor to ground.

Undervoltage Lock-Out (UVLO)

The TPS735xx utilizes an undervoltage lock-out circuit to keep the output shut off until internal circuitry is operating properly. The UVLO circuit has a de-glitch feature so that it typically ignores undershoot transients on the input if they are less than 50μ s duration.

Minimum Load

The TPS735xx is stable and well-behaved with no output load. To meet the specified accuracy, a minimum load of 500 μ A is required. Below 500 μ A at junction temperatures near +125°C, the output can drift up enough to cause the output pull-down to turn on. The output pull-down limits voltage drift to 5% typically but ground current could increase by approximately 50 μ A. In typical applications, the junction cannot reach high temperatures at light loads because there is no appreciable dissipated power. The specified ground current would then be valid at no load in most applications.

optimize noise, equivalent series resistance of the output capacitor can be set to approximately 0.2Ω . This configuration maximizes phase margin in the control loop, reducing total output noise by up to 10%.

Noise can be referred to the feedback point (FB pin) such that with $C_{NR} = 0.01 \mu F$, total noise is given approximately by Equation 1:

$$V_{\rm N} = \frac{11\mu V_{\rm RMS}}{V} \times V_{\rm OUT}$$
(1)

The TPS73501 adjustable version does not have the noise-reduction pin available, so ultra-low noise operation is not possible. Noise can be minimized according to the above recommendations.

Board Layout Recommendations to Improve PSRR and Noise Performance

To improve ac performance such as PSRR, output noise, and transient response, it is recommended that the board be designed with separate ground planes for $V_{\rm IN}$ and $V_{\rm OUT}$, with each ground plane connected only at the GND pin of the device. In addition, the ground connection for the bypass capacitor should connect directly to the GND pin of the device.

Internal Current Limit

The TPS735xx internal current limit helps protect the regulator during fault conditions. During current limit, the output sources a fixed amount of current that is largely independent of output voltage. For reliable operation, the device should not be operated in current limit for extended periods of time.

The PMOS pass element in the TPS735xx has a built-in body diode that conducts current when the voltage at OUT exceeds the voltage at IN. This current is not limited, so if extended reverse voltage operation is anticipated, external limiting may be appropriate.

Shutdown

The enable pin (EN) is active high and is compatible with standard and low voltage TTL-CMOS levels. When shutdown capability is not required, EN can be connected to IN.

Dropout Voltage

The TPS735xx uses a PMOS pass transistor to achieve low dropout. When $(V_{IN} - V_{OUT})$ is less than the dropout voltage (V_{DO}) , the PMOS pass device is in its linear region of operation and the input-to-output resistance is the R_{DS, ON} of the PMOS pass element. Because the PMOS device behaves like a resistor in dropout, V_{DO} approximately scales with output current.

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Thermal Information

Thermal Protection

Thermal protection disables the output when the junction temperature rises to approximately +165°C, allowing the device to cool. When the junction temperature cools to approximately +145°C the output circuitry is again enabled. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits the dissipation of the regulator, protecting it from damage as a result of overheating.

Any tendency to activate the thermal protection circuit indicates excessive power dissipation or an inadequate heatsink. For reliable operation, junction temperature should be limited to +125°C maximum. To estimate the margin of safety in a complete design (including heatsink), increase the ambient temperature until the thermal protection is triggered; use worst-case loads and signal conditions. For good reliability, thermal protection should trigger at least +35°C above the maximum expected ambient condition of your particular application. This configuration produces a worst-case junction temperature of +125°C at the highest expected ambient temperature and worst-case load.

The internal protection circuitry of the TPS735xx has been designed to protect against overload conditions. It was not intended to replace proper heatsinking. Continuously running the TPS735xx into thermal shutdown degrades device reliability.

Power Dissipation

The ability to remove heat from the die is different for each package type, presenting different considerations in the PCB lavout. The PCB area around the device that is free of other components moves the heat from the device to the ambient air. Performance data for JEDEC low- and high-K boards are given in the Dissipation Ratings table. Using heavier copper increases the effectiveness in removing heat from the device. The addition of plated through-holes to heat-dissipating lavers also improves the heatsink effectiveness.

Power dissipation depends on input voltage and load conditions. Power dissipation is equal to the product of the output current time the voltage drop across the output pass element, as shown in Equation 2:

$$\mathsf{P}_{\mathsf{D}} = \left(\mathsf{V}_{\mathsf{IN}} - \mathsf{V}_{\mathsf{OUT}}\right) \cdot \mathsf{I}_{\mathsf{OUT}} \tag{2}$$

Note: When the device is used in a condition of higher input and lower output voltages with the DRV and DRB packages, P_D exceeds the package rating at room temperature. This equation shows an example of the DRB package:

 $P_D = (6.5V - 1.0V) \times 500mA = 2.75W$, which is greater than 2.5W at +25°C.

Package Mounting

Solder pad footprint recommendations for the TPS735xx are available from the Texas Instruments web site at www.ti.com.

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TPS73501DRBR	PREVIEW	SON	DRB	8	3000	TBD	Call TI	Call TI
TPS73501DRBT	PREVIEW	SON	DRB	8	250	TBD	Call TI	Call TI
TPS73501DRVR	PREVIEW	SON	DRV	6		TBD	Call TI	Call TI
TPS73501DRVT	PREVIEW	SON	DRV	6		TBD	Call TI	Call TI
TPS73525DRBR	ACTIVE	SON	DRB	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS73525DRBRG4	ACTIVE	SON	DRB	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS73525DRBT	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS73525DRBTG4	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS73525DRVR	PREVIEW	SON	DRV	6		TBD	Call TI	Call TI
TPS73525DRVT	PREVIEW	SON	DRV	6		TBD	Call TI	Call TI
TPS73533DRBR	PREVIEW	SON	DRB	8	3000	TBD	Call TI	Call TI
TPS73533DRBT	PREVIEW	SON	DRB	8	250	TBD	Call TI	Call TI
TPS73533DRVR	PREVIEW	SON	DRV	6	3000	TBD	Call TI	Call TI
TPS73533DRVT	PREVIEW	SON	DRV	6	250	TBD	Call TI	Call TI

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device		Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS73525DRBR	SON	DRB	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPS73525DRBT	SON	DRB	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2



PACKAGE MATERIALS INFORMATION

8-Dec-2008



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS73525DRBR	SON	DRB	8	3000	346.0	346.0	29.0
TPS73525DRBT	SON	DRB	8	250	190.5	212.7	31.8

PLASTIC SMALL OUTLINE



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

B. This drawing is subject to change without notice.

DRB (S-PDSO-N8)

C. Small Outline No-Lead (SON) package configuration.

The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.

A Metalized features are supplier options and may not be on the package.





THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No-Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

DRB (S-VSON-N8)



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for solder mask tolerances.



MECHANICAL DATA





C. Small Outline No-Lead (SON) package configuration.

DRV (S-PDSO-N6)

The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.



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