



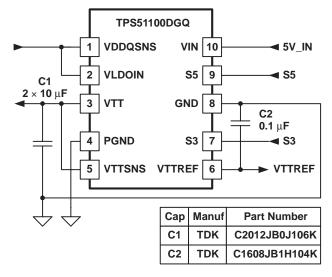
# 3-A SINK/SOURCE DDR TERMINATION REGULATOR

### **FEATURES**

- Input Voltage Range: 4.75 V to 5.25 V
- VLDOIN Voltage Range: 1.2 V to 3.6 V
- 3-A Sink/Source Termination Regulator Includes Droop Compensation
- Requires Only 20-μF Ceramic Output Capacitance
- Supports High-Z in S3 and Soft-Off in S5
- 1.2-V Input (VLDOIN) Helps Reduce Total Power Dissipation
- Integrated Divider Tracks 1/2VDDQSNS for VTT and VTTREF
- Remote Sensing (VTTSNS)
- ± 20-mV Accuracy for VTT and VTTREF
- 10-mA Buffered Reference (VTTREF)
- Built-In Soft-Start, UVLO and OCL
- Thermal Shutdown
- Supports JEDEC Specifications

### **APPLICATIONS**

- DDR, DDR2 Memory Termination
- SSTL-2, SSTL-18 and HSTL Termination



UDG-04015

### **DESCRIPTION**

The TPS51100 is a 3-A sink/source tracking termination regulator. It is specifically designed for low-cost/low-external component count systems, where space is a premium.

The TPS51100 maintains fast transient response only requiring 20- $\mu F$  (2  $\times$  10 $\mu F$ ) of ceramic output capacitance. The TPS51100 supports remote sensing functions and all features required to power the DDR and DDR2 VTT bus termination according to the JEDEC specification. In addition, the TPS51100 includes integrated sleep-state controls placing VTT in High-Z in S3 (suspend to RAM) and soft-off for VTT and VTTREF in S5 (suspend to disk). The TPS51100 is available in the thermally efficient 10-pin MSOP PowerPADTM and is specified from  $-40^{\circ}C$  to  $85^{\circ}C$ .

### ORDERING INFORMATION

TA	PLASTIC MSOP POWER PAD (DGQ) <sup>(1)</sup>
-40°C to 85°C	TPS51100DGQ

(1) The DGQ package is also available taped and reeled. Add an R suffix to the device type (i.e., TPS51100DGQR). See the application section of the data sheet for PowerPAD drawing and layout information.

### SLUS600B - APRIL 2004 - REVISED NOVEMBER 2005



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range unless otherwise noted<sup>(1)</sup>

		TPS51100	UNIT
(2)	VIN, VLDOIN, VTTSNS, VDDQSNS, S3, S5	-0.3 to 6	
Input voltage range(2)	PGND	-0.3 to 0.3	V
Output voltage range(2)	VTT, VTTREF	-0.3 to 6	1
Operating ambient temperature range, TA	-40 to 85	00	
Storage temperature, T <sub>Stg</sub>		-55 to 150	°C

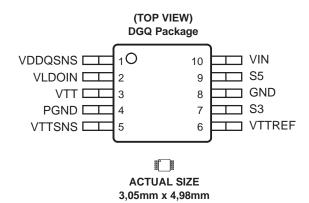
<sup>(1)</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### **DISSIPATION RATING TABLE**

PACKAGE	T <sub>A</sub> < 25°C	DERATING FACTOR	T <sub>A</sub> = 85°C
	POWER RATING	ABOVE T <sub>A</sub> = 25°C	POWER RATING
10-pin DGQ	1.73 W	17.3 mW/°C	0.694 W

### RECOMMENDED OPERATING CONDITIONS

		MIN	MAX	UNIT
Supply voltage, V <sub>IN</sub>		4.75	5.25	
	S3, S5	-0.10	5.25	
Malla na na na	VLDOIN, VDDQSNS, VTT, VTTSNS	-0.1	3.6	V
Voltage range	VTTREF	-0.1	1.8	
	PGND	-0.1	0.1	
Operating free-air temperature, TA		-40	85	°C



- (4) For more information on the DGQ package, refer to TI Technical Brief, Literature No. SLMA002.
- (5) PowerPAD™ heat slug must be connected to GND (pin 8) or electrically isolated from all other pins.



<sup>(2)</sup> All voltage values are with respect to the network ground terminal unless otherwise noted.

### **ELECTRICAL CHARACTERISTICS**

 $T_A = -40^{\circ}C$  to 85°C,  $V_{VIN} = 5$  V, VLDOIN and VDDQSNS are connected to 2.5 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY CURF	RENT					
I <sub>VIN</sub>	Supply current, VIN	$T_A = 25^{\circ}C$ , $V_{VIN} = 5 V$ , no load $V_{S3} = V_{S5} = 5 V$	0.25	0.50	1.00	mA
IVINSTB	Standby currrent, VIN	$T_A = 25^{\circ}C$ , $V_{VIN} = 5 \text{ V}$ , no load $V_{S3} = 0 \text{ V}$ , $V_{S5} = 5 \text{ V}$	25	50	80	4
IVINSDN	Shutdown current, VIN	$ \begin{array}{llllllllllllllllllllllllllllllllllll$		0.3	1.0	μΑ
I <sub>VLDOIN</sub>	Supply current, VLDOIN	$T_A = 25^{\circ}C$ , $V_{VIN} = 5 V$ , no load $V_{S3} = V_{S5} = 5 V$	0.7	1.2	2.0	mA
IVLDOINSTB	Standby currrent, VLDOIN	$T_A = 25^{\circ}C$ , $V_{VIN} = 5 \text{ V}$ , no load $V_{S3} = 0 \text{ V}$ , $V_{S5} = 5 \text{ V}$		6	10	4
IVLDOINSDN	Shutdown current, VLDOIN	$T_A = 25^{\circ}C$ , $V_{VIN} = 5 V$ , no load $V_{S3} = V_{S5} = 0 V$		0.3	1.0	μА
INPUT CURRE	ENT					
IVDDQSNS	Input current, VDDQSNS	$V_{VIN} = 5 \text{ V}, \qquad V_{S3} = V_{S5} = 5 \text{ V}$	1	3	5	^
I <sub>VTTSNS</sub>	Input current, VTTSNS	$V_{VIN} = 5 \text{ V}, \qquad V_{S3} = V_{S5} = 5 \text{ V}$	-1.00	-0.25	1.00	μΑ
VTT OUTPUT						
.,	Output voltage, VTT	V <sub>VLDOIN</sub> = V <sub>VDDQSNS</sub> = 2.5 V		1.25		V
VVTTSNS		V <sub>V</sub> LDOIN = V <sub>V</sub> DDQSNS = 1.8 V		0.9		V
		V <sub>V</sub> LDOIN = V <sub>V</sub> DDQSNS = 2.5  W <sub>TT</sub>   = 0 A	-20		20	
VVTTTOL25	Output voltage tolerance to VTTREF, VTT	V <sub>V</sub> LDOIN = V <sub>V</sub> DDQSNS = 2.5  W <sub>T</sub> T  = 1.5 A	-30		30	
	VIINER, VII	V <sub>V</sub> LDOIN = V <sub>V</sub> DDQSNS = 2.5  W <sub>TT</sub>   = 3 A	-40		40	
		V <sub>V</sub> LDOIN = V <sub>V</sub> DDQSNS = 1.8  W <sub>T</sub> T  = 0 A	-20		20	mV
VVTTTOL18	Output voltage tolerance to	V <sub>VLDOIN</sub> = V <sub>VDDQSNS</sub> = 1.8  W <sub>VTT</sub>   = 1 A	-30		30	
**********	VTTREF, VTT	V <sub>VLDOIN</sub> = V <sub>VDDQSNS</sub> = 1.8  W <sub>TT</sub>   = 2 A	-40		40	
           	Source current limit, VTT	$V_{TT} = \left(\frac{V_{VDDQSNS}}{2}\right) \times 0.95,  PGOOD = High$	3.0	3.8	6.0	
		V <sub>VTT</sub> = 0 V	1.5	2.2	3.0	Α
IVTTOCLSNK	Sink current limit, VTT	$V_{TT} = \left(\frac{V_{VDDQSNS}}{2}\right) \times 1.05,  PGOOD = High$	3.0	3.6	6.0	A
		V <sub>VTT</sub> = V <sub>VDDQ</sub>	1.5	2.2	3.0	
I <sub>VTTLK</sub>	Leakage current, VTT	$V_{TT} = \left(\frac{V_{VDDQSNS}}{2}\right) = 1.25 \text{ V},  T_A = 25^{\circ}\text{C}$ $V_{S3} = 0 \text{ V},  V_{S5} = 5 \text{ V}$	-1.0	0.5	1.0	μА
IVTTSNSLK	Leakage current, VTTSNS	$V_{TT} = \left(\frac{V_{VDDQSNS}}{2}\right) = 1.25 \text{ V},  T_A = 25^{\circ}\text{C}$	-1.00	0.01	1.00	•
IDSCHRG	Discharge current, VTT	T <sub>A</sub> = 25°C, V <sub>S3</sub> = V <sub>S5</sub> = 0 V, V <sub>VDDQSNS</sub> = 0 V, V <sub>VTT</sub> = 0.5 V	10	17		mA

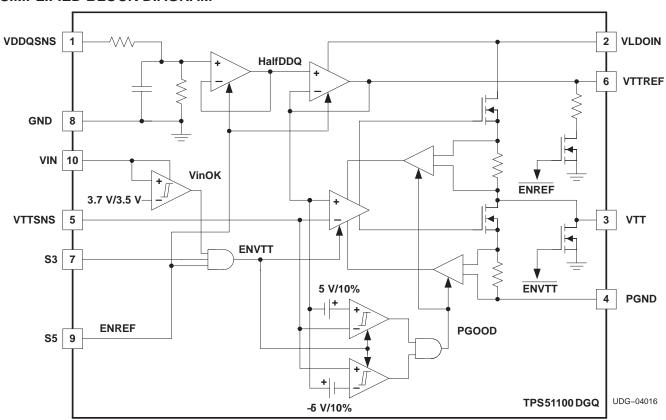
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VTTREF OUTP	UT					
VVTTREF	Output voltage, VTTREF		V <sub>TTREF</sub>	$V = \left(\frac{V_{VD}}{V_{D}}\right)$	DDQSNS 2	٧
VVTTREFTOL25	Output voltage tolerance to VDDQSNS/2, VTTREF	VVLDOIN = VVDDQSNS = 2.5 VTTREF < 10 mA	-20		20	
VVTTREFTOL18	Output voltage tolerance to VDDQSNS/2, VTTREF	VVLDOIN = VVDDQSNS = 1.8 VTTREF < 10 mA	-17		17	mV
IVTTREFOCL	Source current limit, VTTREF	V <sub>VVTTREF</sub> = 0 V	10	20	30	mA
UVLO/LOGIC T	THRESHOLD					
.,	LIVILO de manda del contra man AVIA	Wake up	3.4	3.7	4.0	
VVINUV	UVLO threshold voltage, VIN	Hysteresis	0.15	0.25	0.35	
VIH	High-level input voltage	S3, S5	1.6			V
VIL	Low-level input voltage	S3, S5			0.3	
VIHYST	Hysteresis voltage	S3, S5		0.2		
ILEAK	Logic input leakage current	S2, S5, T <sub>A</sub> = 25°C	-1		1	μΑ
THERMAL SHU	JTDOWN					
_	<del>-</del>	Shutdown temperature		160		
T <sub>SDN</sub>	Thermal shutdown threshold	Hysteresis		10		°C

### **TERMINAL FUNCTIONS**

TERMIN	TERMINAL		TERMINAL		TERMINAL		
NAME	NO.	1/0	DESCRIPTION				
GND	8	-	Signal ground. Connect to negative terminal of the output capacitor				
PGND	4	_	Power ground output for the VTT LDO				
S3	7	- 1	S3 signal input				
S5	9	I	S5 signal input				
VDDQSNS	1	I	VDDQ sense input				
VIN	10	I	5-V power supply				
VLDOIN	2	- 1	Power supply for the VTT LDO and VTTREF output stage				
VTT	3	0	Power output for the VTT LDO				
VTTREF	6	0	VTT reference output. Connect to GND through 0.1-μF ceramic capacitor.				
VTTSNS	5	I	Voltage sense input for the VTT LDO. Connect to plus terminal of the output capacitor.				



### SIMPLIFIED BLOCK DIAGRAM



### **DETAILED DESCRIPTION**

### VTT SINK/SOURCE REGULATOR

The TPS51100 is a 3-A sink/source tracking termination regulator designed specially for low-cost, low external components system where space is at premium such as notebook PC applications. TPS51100 integrates high-performance low-dropout linear regulator that is capable of sourcing and sinking current up to 3 A. This VTT linear regulator employs ultimate fast response feedback loop so that small ceramic capacitors are enough to keep tracking to the VTTREF within  $\pm 40$  mV at all conditions including fast load transient. To achieve tight regulation with minimum effect of trace resistance, a remote sensing terminal, VTTSNS, should be connected to the positive node of VTT output capacitor(s) as a separate trace from the high current line from VTT.

### VTTREF REGULATOR

The VTTREF block consists of an on-chip 1/2 divider, LPF and buffer. This regulator can source current up to 10 mA. Bypass VTTREF to GND using a 0.1-μF ceramic capacitor to ensure stable operation.

### Soft-Start

The soft-start function of the VTT is achieved via a current clamp, allowing the output capacitors to be charged with low and constant current that gives linear ramp up of the output voltage. The current limit threshold is changed in two stages using an internal powergood signal. When VTT is outside the powergood threshold, the current limit level is 2.2 A. When VTT rises above (VTTREF – 5%) or falls below (VTTREF + 5%) the current limit level switches to 3.8 A. The thresholds are typically VTTREF  $\pm 5\%$  (from outside regulation to inside) and  $\pm 10\%$  (when it falls outside). The soft-start function is completely symmetrical and it works not only from GND to VTTREF voltage, but also from VDDQ to VTTREF voltage. Note that the VTT output is in a high impedance state during the S3 state (S3 = low, S5 = high) and its voltage can be up to VDDQ voltage depending on the external condition. Note that VTT does not start under a full load condition.

### S3. S5 Control and Soft-Off

The S3 and S5 terminals should be connected to SLP\_S3 and SLP\_S5 signals respectively. Both VTTREF and VTT are turned on at S0 state (S3 = high), S5 = high). VTTREF is kept alive while VTT is turned off and left high impedance in S3 state (S3 = low, S5 = high). Both VTT and VTTREF outputs are turned off and discharged to the ground through internal MOSFETs during S4/S5 state (both S3 and S5 are low).

**VTTREF STATE S**3 **S5** VTT Н Н S0 S3 L Н 0 (high-Z) 1 S4/S5 L L 0 (discharge) 0 (discharge)

Table 1, S3 and S5 Control Table

(In case S3 is forced H and S5 to L, VTTREF is discharged and VTT is at High–Z state. This condition is NOT recommended.)

### **VTT Current Protection**

The LDO has a constant overcurrent limit (OCL) at 3.8 A. This trip point is reduced to 2.2 A before the output voltage comes within  $\pm 5\%$  of the target voltage or goes outside of  $\pm 10\%$  of the target voltage.



### **DETAILED DESCRIPTION**

### **VIN UVLO Protection**

For VIN undervoltage lockout (UVLO) protection, the TPS51100 monitors VIN voltage. When the VIN voltage is lower than UVLO threshold voltage, the VTT regulator is shut off. This is a non-latch protection.

### **Thermal Shutdown**

TPS51100 monitors its temperature. If the temperature exceeds threshold value, typically 160°C, the VTT and VTTREF regulators are shut off. This is also a non-latch protection.

### **Output Capacitor**

For stable operation, total capacitance of the VTT output terminal can be equal or greater than  $20 - \mu F$ . Attach two  $10 - \mu F$  ceramic capacitors in parallel to minimize the effect of ESR and ESL. If the ESR is greater than  $2 m\Omega$ , insert an R-C filter between the output and the VTTSNS input to achieve loop stability. The R-C filter time constant should be almost the same or slightly lower than the time constant of the output capacitor and its ESR.

Soft-start duration,  $T_{SS}$ , is also a function of this output capacitance. Where  $I_{TTOCL} = 2.2 \text{ A (typ)}$ ,  $T_{SS}$  can be calculated as,

$$T_{SS} = \left(\frac{C_{OUT} \times V_{VTT}}{I_{VTTOCL}}\right) \tag{1}$$

### **Input Capacitor**

Depending on the trace impedance between the VLDOIN bulk power supply to the part, transient increase of source current is supplied mostly by the charge from the VLDOIN input capacitor. Use a 10- $\mu$ F (or more) ceramic capacitor to supply this transient charge. Provide more input capacitance as more output capacitance is used at VTT. In general, use 1/2  $C_{OLT}$  for input.

### **VIN Capacitor**

Add a ceramic capacitor with a value between  $1.0 - \mu F$  and  $4.7 - \mu F$  placed close to the VIN pin, to stabilize 5-V from any parasitic impedance from the supply.

### Thermal design

As the TPS51100 is a linear regulator, the VTT current flow in both source and sink directions generate power dissipation from the device. In the source phase, the potential difference between V<sub>VLDOIN</sub> and V<sub>VTT</sub> times VTT current becomes the power dissipation, W<sub>DSRC</sub>.

$$W_{DSRC} = (V_{VLDOIN} - V_{VTT}) \times I_{VTT}$$
(2)

In this case, if VLDOIN is connected to an alternative power supply lower than V<sub>DDQ</sub> voltage, power loss can be decreased.

For the sink phase, VTT voltage is applied across the internal LDO regulator, and the power dissipation, and  $W_{DSNK}$ , is calculated by:

$$W_{DSNK} = V_{VTT} \times I_{VTT}$$
(3)

Since the device does not sink and source the current at the same time and  $I_{VTT}$  varies rapidly with time, actual power dissipation need to be considered for thermal design is an average of above value over thermal relaxation duration of the system. Another power consumption is the current used for internal control circuitry from VIN supply and VLDOIN supply. This can be estimated as 20 mW or less at normal operational conditions. This power needs to be effectively dissipated from the package. Maximum power dissipation allowed to the package is calculated by,



### **DETAILED DESCRIPTION**

$$W_{PKG} = \frac{\left(T_{J(max)} - T_{A(max)}\right)}{\theta_{JA}}$$
(4)

where

- T<sub>J(max)</sub> is 125°C
- T<sub>A(max)</sub> is the maximum ambient temperature in the system
- θ<sub>JA</sub> is the thermal resistance from the silicon junction to the ambient

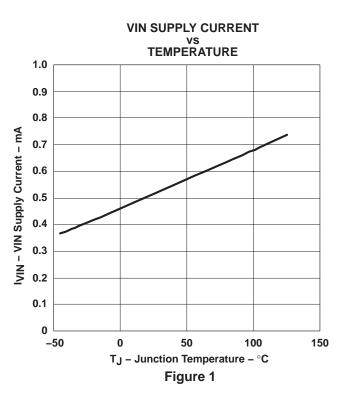
This thermal resistance strongly depends on the board layout. TPS51100 is assembled in a thermally enhanced PowerPAD<sup>TM</sup> package that has exposed die pad underneath the body. For improved thermal performance, this die pad needs to be attached to ground trace via thermal land on the PCB. This ground trace acts as a heat sink/spread. The typical thermal resistance,  $57.7^{\circ}$ C/W, is achieved based on a 3 mm  $\times$  2 mm thermal land with 2 vias without air flow. It can be improved by using larger thermal land and/or increasing vias number. For example, assuming 3 mm  $\times$  3 mm thermal land with 4 vias without air flow, it is  $45.4^{\circ}$ C/W. Further information about PowerPAD<sup>TM</sup> and its recommended board layout is described in the application note (SLMA002). This document is available at www.ti.com.

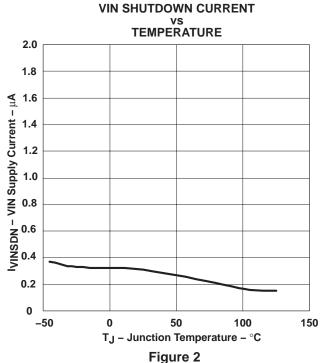
### LAYOUT CONSIDERATIONS

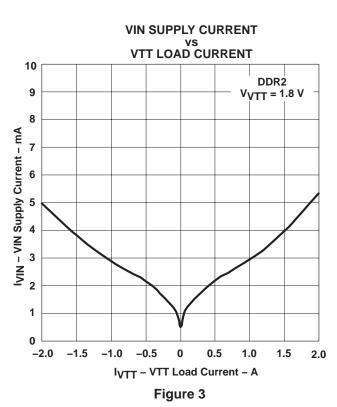
Consider the following points before the layout of TPS51100 design begins.

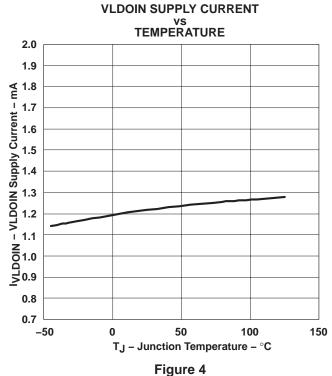
- The input bypass capacitor for VLDOIN should be placed to the pin as close as possible with short and wide connection.
- The output capacitor for VTT should be placed close to the pin with short and wide connection in order to avoid additional ESR and/or ESL of the trace.
- VTTSNS should be connected to the positive node of VTT output capacitor(s) as a separate trace from the
  high current power line and is strongly recommended to avoid additional ESR and/or ESL. If it is needed
  to sense the voltage of the point of the load, it is recommended to attach the output capacitor(s) at that point.
  Also, it is recommended to minimize any additional ESR and/or ESL of ground trace between GND pin and
  the output capacitor(s).
- Consider adding LPF at VTTSNS in case ESR of the VTT output capacitor(s) is larger than 2 mΩ.
- VDDQSNS can be connected separately from VLDOIN. Remember that this sensing potential is the reference voltage of VTTREF. Avoid any noise generative lines.
- Negative node of VTT output capacitor(s) and VTTREF capacitor should be tied together by avoiding common impedance to the high current path of the VTT source/sink current.
- GND (Signal GND) pin node represents the reference potential for VTTREF and VTT outputs. Connect GND to negative nodes of VTT capacitor(s), VTTREF capacitor and VDDQ capacitor(s) with care to avoid additional ESR and/or ESL. GND and PGND (Power GND) should be isolated, with a single point connection between them.
- In order to effectively remove heat from the package, prepare thermal land and solder to the package's thermal pad. Wide trace of the component–side copper, connected to this thermal land, will help heat spreading. Numerous vias 0.33 mm in diameter connected from the thermal land to the internal/solder–side ground plane(s) should be used to help dissipation.

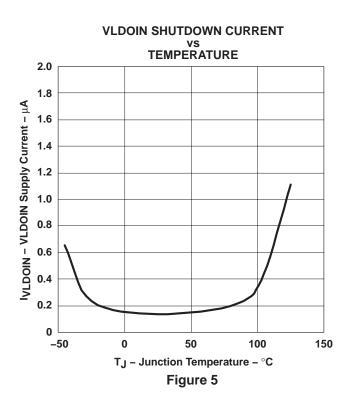


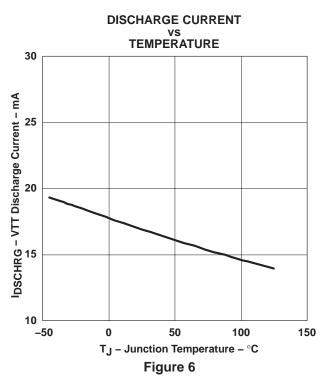




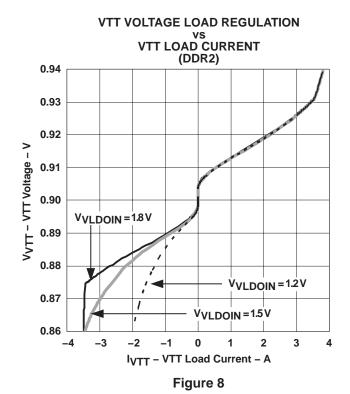




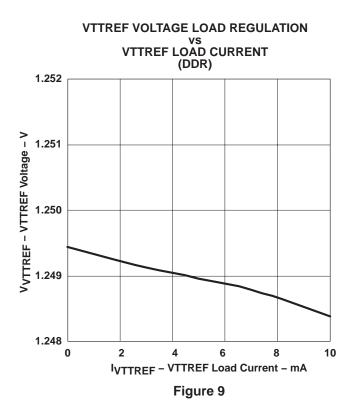


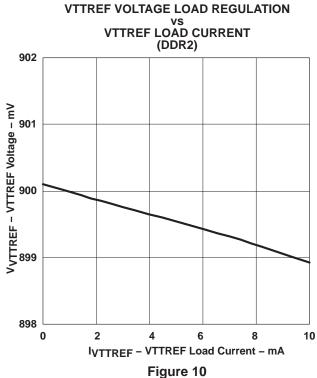


### **VTT VOLTAGE LOAD REGULATION** vs VTT LOAD CURRENT (DDR) 1.29 1.28 1.27 VVTT - VTT Voltage - V 1.26 1.25 $V_{VLDOIN} = 2.5 V$ 1.24 1.23 V<sub>VLDOIN</sub> = 1.8 V 1.22 1.21 -3 -2 0 2 3 I<sub>VTT</sub> - VTT Load Current - A Figure 7



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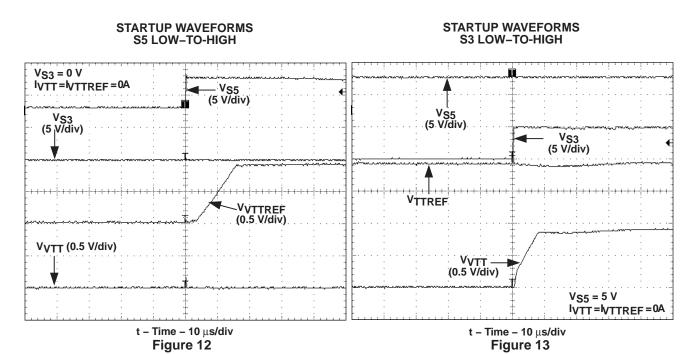




# VVLDOIN (50 mV/div) Offset: 1.8 V VVTTREF (20 mV/div) Offset 0.9 V t – Time – 20 µs/div

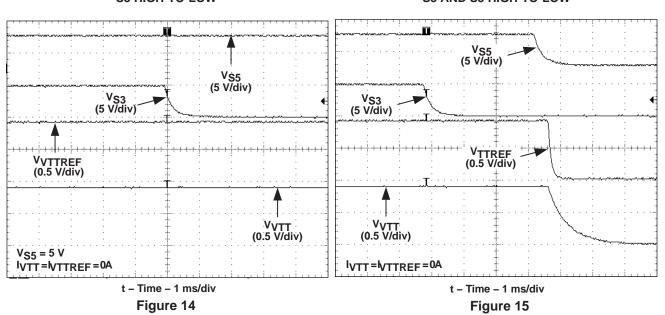
Figure 11

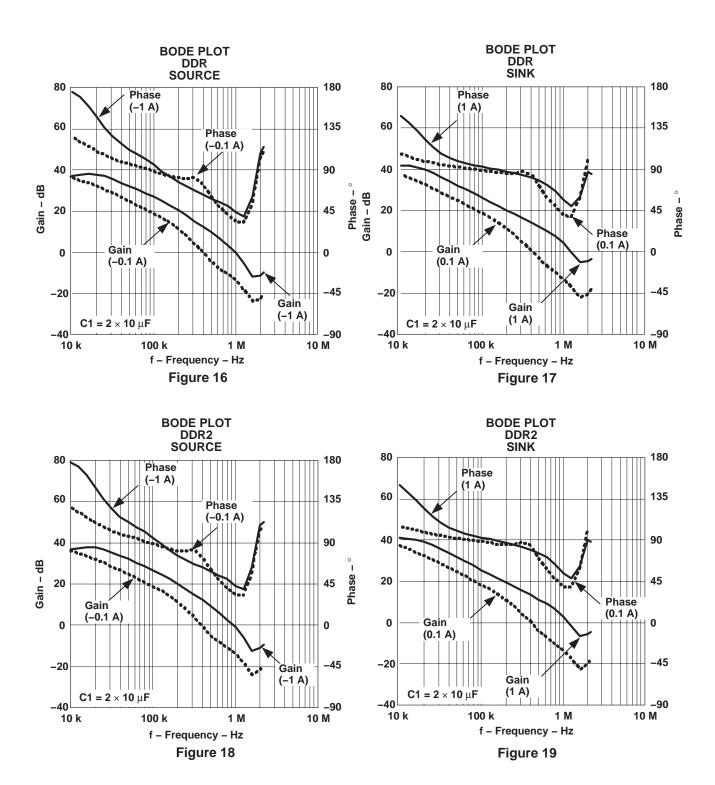




### SHUTDOWN WAVEFORMS S3 HIGH-TO-LOW

### SHUTDOWN WAVEFORMS S3 AND S5 HIGH-TO-LOW











i.com 27-Feb-2006

### PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TPS51100DGQ	ACTIVE	MSOP- Power PAD	DGQ	10	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS51100DGQG4	ACTIVE	MSOP- Power PAD	DGQ	10	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS51100DGQR	ACTIVE	MSOP- Power PAD	DGQ	10	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS51100DGQRG4	ACTIVE	MSOP- Power PAD	DGQ	10	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

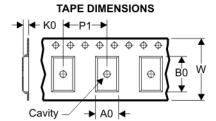
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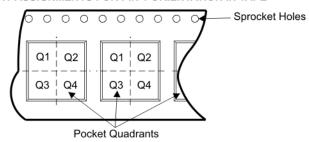
### TAPE AND REEL BOX INFORMATION

# REEL DIMENSIONS Reel Diameter Reel Widtle



	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



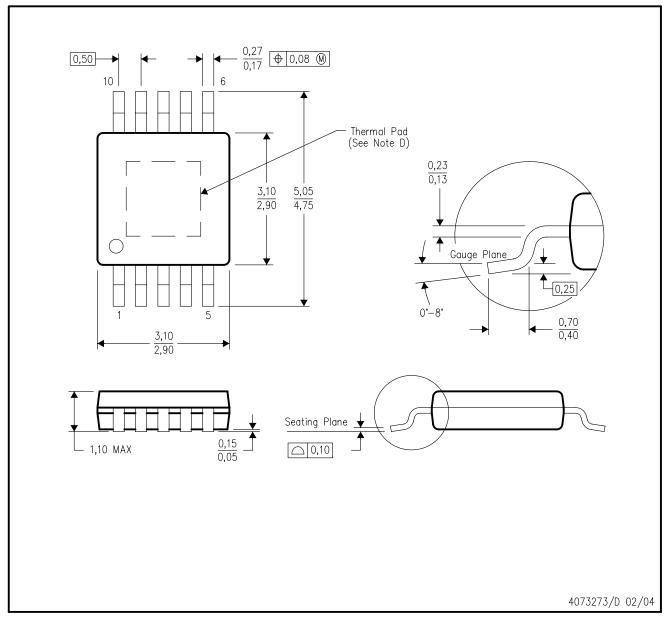
Device	Package	Pins	Site	Reel Diameter (mm)	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS51100DGQR	DGQ	10	SITE 60	330	12	5.3	3.4	1.4	8	12	Q1





Device	Package	Pins	Site	Length (mm)	Width (mm)	Height (mm)
TPS51100DGQR	DGQ	10	SITE 60	346.0	346.0	29.0

## DGQ (S-PDSO-G10) PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <a href="https://www.ti.com">http://www.ti.com</a>>.
- E. Falls within JEDEC MO-187 variation BA-T.

PowerPAD is a trademark of Texas Instruments.



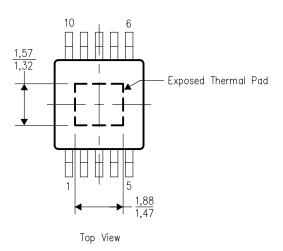


### THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. When the thermal pad is soldered directly to the printed circuit board (PCB), the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to a ground or power plane (whichever is applicable), or alternatively, a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

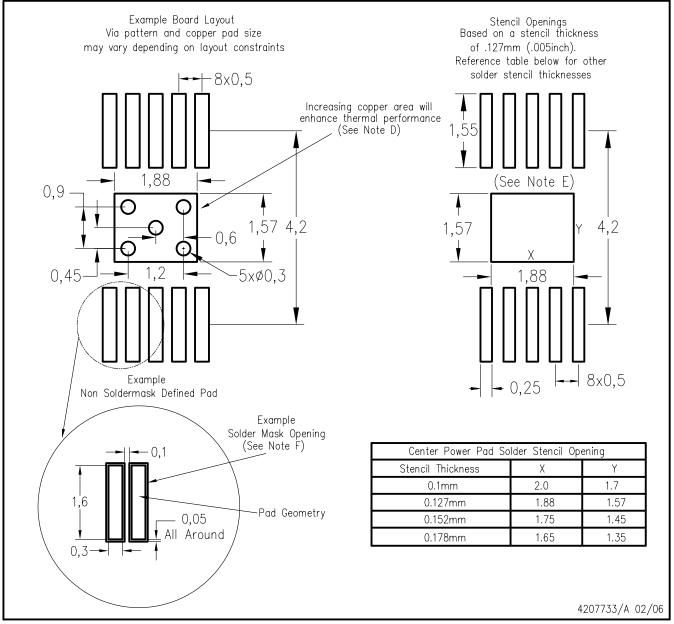
The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

# DGQ (R-PDSO-G10) PowerPAD™



### NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
- F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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