SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

- Floating Bootstrap or Ground-Reference **High-Side Driver**
- **Adaptive Dead-Time Control**
- 50-ns Max Rise/Fall Times and 100-ns Max Propagation Delay – 3.3-nF Load
- **Ideal for High-Current Single or Multiphase Power Supplies**
- 2.4-A Typical Peak Output Current
- 4.5-V to 15-V Supply Voltage Range
- **Internal Schottky Bootstrap Diode**
- Low Supply Current....3-mA Typical
- -40°C to 125°C Operating Virtual Junction **Temperature**
- Available in SOIC Package

D PACKAGE (TOP VIEW) IN \Box 8 **□** BOOT PGND I 7 ☐ HIGHDR $DT \square$ 6 □ BOOTLO 5 LOWDR V_{CC} □

description

The TPS2832 and TPS2833 are MOSFET drivers for synchronous-buck power stages. These devices are ideal for designing a high-performance power supply using switching controllers that do not have MOSFET drivers. The drivers are designed to deliver 2.4-A peak currents into large capacitive loads. The high-side driver can be configured as a ground-reference driver or as a floating bootstrap driver. An adaptive dead-time control circuit eliminates shoot-through currents through the main power FETs during switching transitions and provides high efficiency for the buck regulator.

The TPS2832 has a noninverting input. The TPS2833 has an inverting input. The TPS2832/33 drivers, available in 8-terminal SOIC packages, operate over a junction temperature range of -40°C to 125°C.

AVAILABLE OPTIONS

	PACKAGED DEVICES					
TJ	SOIC (D)					
-40°C to 125°C	TPS2832D TPS2833D					

The D package is available taped and reeled. Add R suffix to device type (e.g., TPS2832DR)

Related Synchronous MOSFET Drivers

DEVICE NAME	ADDITIONAL FEATURES	INPUTS		
TPS2830	ENABLE OVALO LODOVADAD	01400	Noninverted	
TPS2831	ENABLE, SYNC and CROWBAR	CMOS	Inverted	
TPS2834	ENIARI E OVAIO I OROMARA		Noninverted	
TPS2835	ENABLE, SYNC and CROWBAR	TTL	Inverted	
TPS2836	W/O ENABLE CVAIC and CDOWDAD		Noninverted	
TPS2837	W/O ENABLE, SYNC and CROWBAR	TTL	Inverted	

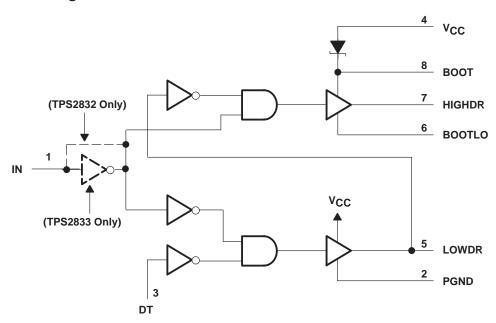


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SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

functional block diagram



Terminal Functions

TERMINAL		1/0	DECORIDETION
NAME	NO.	1/0	DESCRIPTION
воот	8	I	Bootstrap terminal. A ceramic capacitor is connected between BOOT and BOOTLO terminals to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1 μF and 1 μF . A 1-M Ω resistor should be connected across the bootstrap capacitor to provide a discharge path when the driver has been powered down.
BOOTLO	6	0	This terminal connects to the junction of the high-side and low-side MOSFETs.
DT	3	-1	Dead-time control terminal. Connect DT to the junction of the high-side and low-side MOSFETs
HIGHDR	7	0	Output drive for the high-side power MOSFET
IN	1	ı	Input signal to the MOSFET drivers (noninverting input for the TPS2832; inverting input for the TPS2833).
LOWDR	5	0	Output drive for the low-side power MOSFET
PGND	2		Power ground. Connect to the FET power ground.
VCC	4	I	Input supply. Recommended that a 1 μF capacitor be connected from $V_{\hbox{\footnotesize{CC}}}$ to PGND.



SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

detailed description

low-side driver

The low-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink.

high-side driver

The high-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink. The high-side driver can be configured as a ground-reference driver or a floating bootstrap driver. The internal bootstrap diode, is a Schottky for improved drive efficiency. The maximum voltage that can be applied between the BOOT terminal and ground is 30 V.

dead-time (DT) control[†]

Dead-time control prevents shoot through current from flowing through the main power FETs during switching transitions by controlling the turn-on times of the MOSFET drivers. The high-side driver is not allowed to turn on until the gate drive voltage to the low-side FET is low, and the low-side driver is not allowed to turn on until the voltage at the junction of the power FETs (Vdrain) is low; the DT terminal connects to the junction of the power FETs.

IN[†]

The IN terminal is a digital terminal that is the input control signal for the drivers. The TPS2832 has a noninverting input; the TPS2833 has an inverting input.

†High-level input voltages on IN and DT must be greater than or equal to 0.7V_{CC}.



SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage range, V _{CC} (see Note 1)	–0.3 V to 16 V
Input voltage range: BOOT to PGND (high-side driver ON)	–0.3 V to 30 V
BOOTLO to PGND	–0.3 V to 16 V
BOOT to BOOTLO	–0.3 V to 16 V
IN (see Note 2)	–0.3 V to 16 V
DT (see Note 2)	–0.3 V to 30 V
Continuous total power dissipation See	Dissipation Rating Table
Operating virtual junction temperature range, T _J	–40°C to 125°C
Storage temperature range, T _{stq}	–65°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds	260°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATING TABLE

PACKAGE	$T_{\mbox{A}} \le 25^{\circ}\mbox{C}$ POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING	
D	600 mW	6.0 mW/°C	330 mW	240 mW	

recommended operating conditions

	MIN	NOM MAX	UNIT
Supply voltage, V _{CC}	4.5	15	V
Input voltage BOOT to PGND	4.5	28	V

electrical characteristics over recommended operating virtual junction temperature range, V_{CC} = 6.5 V, C_L = 3.3 nF (unless otherwise noted)

supply current

	PARAMETER	TEST CO	MIN	TYP	MAX	UNIT	
	Supply voltage range			4.5		15	V
		V _{CC} =15 V				100	μΑ
Vcc	Quiescent current	V _{CC} =12 V, f _{SWX} = 200 kHz, CHIGHDR = 50 pF,	BOOTLO grounded, CLOWDR = 50 pF, See Note 3		3		mA

NOTE 3: Ensured by design, not production tested.



NOTES: 1. Unless otherwise specified, all voltages are with respect to PGND.

^{2.} High-level input voltages on the IN and DT terminals must be less than or equal to VCC.

SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

electrical characteristics over recommended operating virtual junction temperature range, $V_{CC} = 6.5 \text{ V}$, $C_L = 3.3 \text{ nF}$ (unless otherwise noted) (continued)

output drivers

	PARAMETER	२	TEST CONDIT	TONS	MIN	TYP	MAX	UNIT		
		Duty cycle < 2%,	VBOOT - VBOOTLO = 4.5 V,	VHIGHDR = 4 V	0.7	1.1				
	High-side sink (see Note 4)	$t_{DW} < 100 \ \mu s$	VBOOT - VBOOTLO = 6.5 V,	VHIGHDR = 5 V	1.1	1.5		Α		
	(666 11616 1)	(see Note 3)	$V_{BOOT} - V_{BOOTLO} = 12 V$	V _{HIGHDR} = 10.5 V	2	2.4				
	High-side	Duty cycle < 2%,	$V_{BOOT} - V_{BOOTLO} = 4.5 V$	V _{HIGHDR} = 0.5V	1.2	1.4				
	source	t _{pw} < 100 μs	$V_{BOOT} - V_{BOOTLO} = 6.5 V$	V _{HIGHDR} = 1.5 V	1.3	1.6		Α		
Peak output-	(see Note 4)	(see Note 3)	$V_{BOOT} - V_{BOOTLO} = 12 V$	VHIGHDR = 1.5 V	2.3	2.7				
current		Duty cycle < 2%,	$V_{CC} = 4.5 \text{ V},$	V _{LOWDR} = 4 V	1.3	1.8				
	Low-side sink (see Note 4)	t _{pw} < 100 μs	$V_{CC} = 6.5 \text{ V},$	V _{LOWDR} = 5 V	2	2.5		Α		
	(666 : 1616 :)	(see Note 3)	V _{CC} = 12 V,	V _{LOWDR} = 10.5 V	3	3.5				
	Low-side source (see Note 4)	Low-side	Low-side	Duty cycle < 2%,	V _{CC} = 4.5 V,	$V_{LOWDR} = 0.5V$	1.4	1.7		
		t _{pw} < 100 μs	$V_{CC} = 6.5 \text{ V},$	$V_{LOWDR} = 1.5 V$	2	2.4		А		
		(see Note 3)	V _{CC} = 12 V,	$V_{LOWDR} = 1.5 V$	2.5	3				
			$V_{BOOT} - V_{BOOTLO} = 4.5 V_{s}$	VHIGHDR = 0.5 V			5			
	High-side sink (s	ee Note 4)	$V_{BOOT} - V_{BOOTLO} = 6.5 \text{ V}, V_{HIGHDR} = 0.5 \text{ V}$				5	Ω		
			$V_{BOOT} - V_{BOOTLO} = 12 V$	VHIGHDR = 0.5 V			5			
			VBOOT - VBOOTLO = 4.5 V,	VHIGHDR = 4 V			75			
	High-side source	(see Note 4)	$V_{BOOT} - V_{BOOTLO} = 6.5 V_{s}$	V _{HIGHDR} = 6 V			75	Ω		
Output			$V_{BOOT} - V_{BOOTLO} = 12 V$	V _{HIGHDR} =11.5 V			75			
resistance			$V_{DRV} = 4.5 V,$	V _{LOWDR} = 0.5 V			9			
	Low-side sink (se	ee Note 4)	V _{DRV} = 6.5 V	V _{LOWDR} = 0.5 V			7.5	Ω		
			V _{DRV} = 12 V,	V _{LOWDR} = 0.5 V			6			
			V _{DRV} = 4.5 V,	V _{LOWDR} = 4 V			75			
	Low-side source	(see Note 4)	V _{DRV} = 6.5 V,	V _{LOWDR} = 6 V			75	Ω		
			V _{DRV} = 12 V,	V _{LOWDR} = 11.5 V			75			

NOTES: 3. Ensured by design, not production tested.

4. The pull-up/pull-down circuits of the drivers are bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the Rds(on) of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.

dead time

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IH}	High-level input voltage	LOWIDD	Occasilla IV	0.7V _{CC}			.,
V_{IL}	Low-level input voltage	LOWDR	Over the V _{CC} range (see Note 3)			1	V
VIH	High-level input voltage	DT	Over the V _{CC} range	0.7V _{CC}			V
V_{IL}	Low-level input voltage	וטו	Over the VCC range			1	V

NOTE 3: Ensured by design, not production tested.

digital control terminals

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
١	/IH High-level input voltage	Over the Vee range	0.7V _{CC}			V
١	/ _{IL} Low-level input voltage	Over the V _{CC} range			1	V



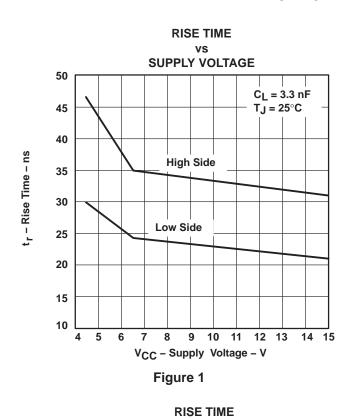
TPS2832, TPS2833 FAST SYNCHRONOUS-BUCK MOSFET DRIVERS WITH DEAD-TIME CONTROL SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

switching characteristics over recommended operating virtual junction temperature range, $C_L=3.3~\mathrm{nF}$ (unless otherwise noted)

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT		
		$V_{BOOT} = 4.5 V$	V _{BOOTLO} = 0 V			60			
	HIGHDR output (see Note 3)	$V_{BOOT} = 6.5 V$	VBOOTLO = 0 V			50	ns		
Diag time		V _{BOOT} = 12 V,	V _{BOOTLO} = 0 V			50			
Rise time		V _{CC} = 4.5 V				40			
	LOWDR output (see Note 3)	V _{CC} = 6.5 V				30	ns		
		V _{CC} = 12 V				30			
		$V_{BOOT} = 4.5 V$	V _{BOOTLO} = 0 V			60			
	HIGHDR output (see Note 3)	$V_{BOOT} = 6.5 V$	V _{BOOTLO} = 0 V			50	ns		
Fall 4:		V _{BOOT} = 12 V,	VBOOTLO = 0 V			50			
Fall time		V _{CC} = 4.5 V				40			
	LOWDR output (see Note 3)	DR output (see Note 3) V _{CC} = 6.5 V					ns		
		V _{CC} = 12 V				30			
	LIIOUDD and and land	$V_{BOOT} = 4.5 V$	VBOOTLO = 0 V			130			
	HIGHDR going low (excluding dead time) (see Note 3)	$V_{BOOT} = 6.5 V$	V _{BOOTLO} = 0 V			100	ns		
Duama matical dalay times	(exclusing used inner (essentials)	$V_{BOOT} = 12 V$,	V _{BOOTLO} = 0 V			75			
Propagation delay time	1004/22	$V_{BOOT} = 4.5 V$	VBOOTLO = 0 V			80			
	LOWDR going high (excluding dead time) (see Note 3)	$V_{BOOT} = 6.5 V$	VBOOTLO = 0 V			70	ns		
	(excluding dead lime) (eee Nete e)	V _{BOOT} = 12 V,	VBOOTLO = 0 V			60			
	1.004/55	V _{CC} = 4.5 V				80			
Propagation delay time	LOWDR going low (excluding dead time) (see Note 3)	V _{CC} = 6.5 V				70	ns		
	(exclusing usua inne) (ess note s)	V _{CC} = 12 V				60			
	DT (a LOW/DD and	V _{CC} = 4.5 V		40		170			
Driver nonoverlap time	DT to LOWDR and LOWDR to HIGHDR (see Note 3)	V _{CC} = 6.5 V		25		135	ns		
	(233.1010.0)	V _{CC} = 12 V		15		85			

NOTE 3: Ensured by design, not production tested.

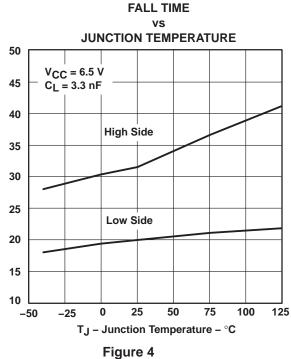




FALL TIME vs **SUPPLY VOLTAGE** 50 $C_{L} = 3.3 \text{ nF}$ T_J = 25°C 45 40 tf - Fall Time - ns 35 **High Side** 30 25 Low Side 20 15 10 5 6 9 10 V_{CC} - Supply Voltage - V

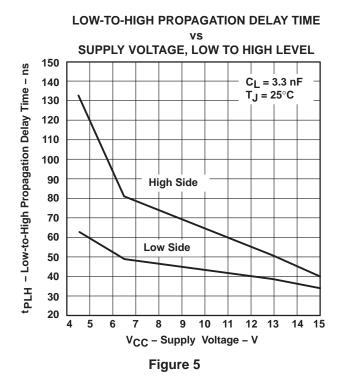


Figure 2



vs JUNCTION TEMPERATURE 50 V_CC = 6.5 V 45 $C_L = 3.3 \text{ nF}$ 40 **High Side** t_r - Rise Time - ns 35 30 Low Side 25 20 15 10 -50 -25 100 125 T_J - Junction Temperature - °C

Figure 3



LOW-TO-HIGH PROPAGATION DELAY TIME

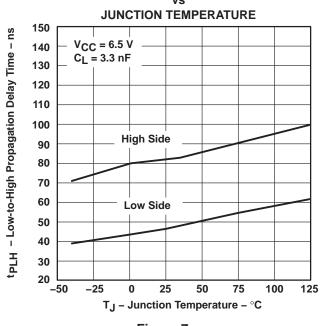
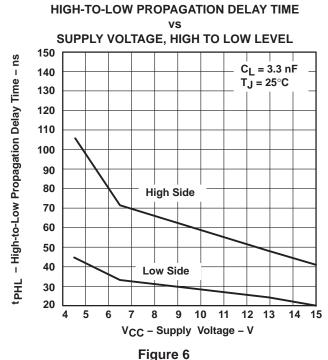


Figure 7



HIGH-TO-LOW PROPAGATION DELAY TIME

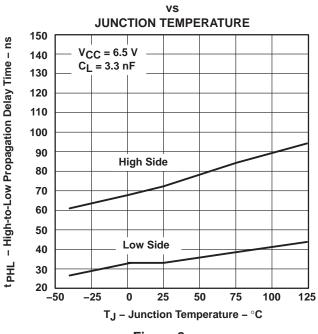
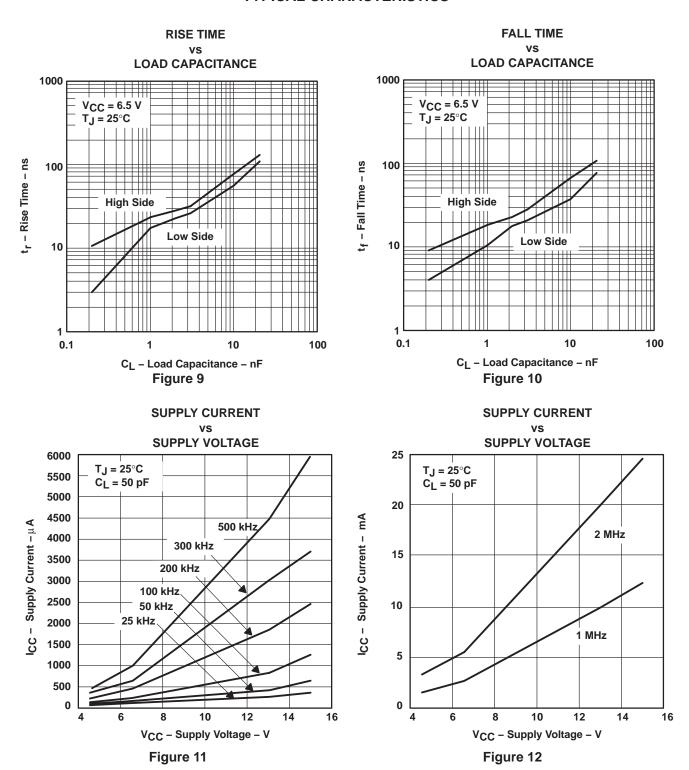
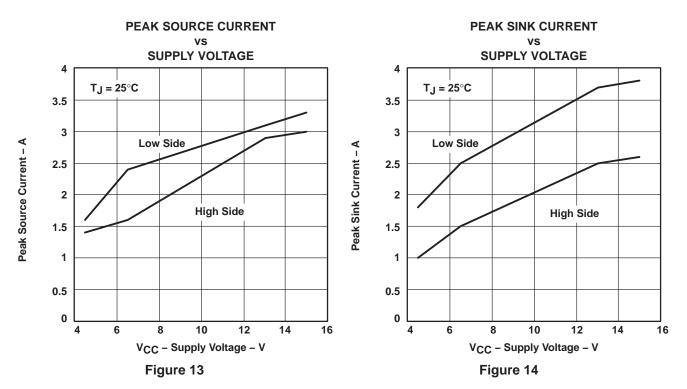


Figure 8







INPUT THRESHOLD VOLTAGE

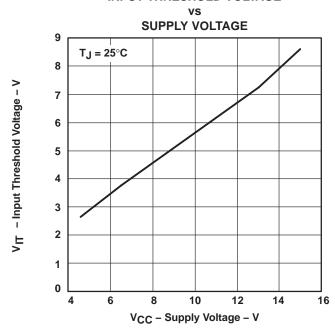


Figure 15

SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

APPLICATION INFORMATION

Figure 16 shows the circuit schematic of a 100-kHz synchronous-buck converter implemented with a TL5001A pulse-width-modulation (PWM) controller and a TPS2833 driver. The converter operates over an input range from 4.5 V to 12 V and has a 3.3 V output. The circuit can supply 3 A continuous load and the transient load is 5 A. The converter achieves an efficiency of 94% for $V_{IN} = 5$ V, $I_{load} = 1$ A, and 93% for $V_{in} = 5$ V, $I_{load} = 3$ A.

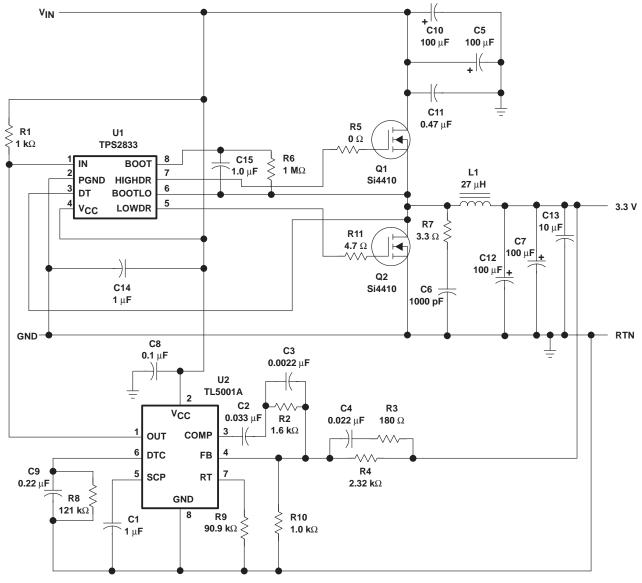


Figure 16. 3.3 V 3 A Synchronous-Buck Converter Circuit

SLVS195C - FEBRUARY 1999 - REVISED JANUARY 2001

APPLICATION INFORMATION

Great care should be taken when laying out the pc board. The power-processing section is the most critical and will generate large amounts of EMI if not properly configured. The junction of Q1, Q2, and L1 should be very tight. The connection from Q1 drain to the positive sides of C5, C10, and C11 and the connection from Q2 source to the negative sides of C5, C10, and C11 should be as short as possible. The negative terminals of C7 and C12 should also be connected to Q2 source.

Next, the traces from the MOSFET driver to the power switches should be considered. The BOOTLO signal from the junction of Q1 and Q2 carries the large gate drive current pulses and should be as heavy as the gate drive traces. The bypass capacitor (C14) should be tied directly across V_{CC} and PGND.

The next most sensitive node is the FB node on the controller (terminal 4 on the TL5001A) This node is very sensitive to noise pickup and should be isolated from the high-current power stage and be as short as possible. The ground around the controller and low-level circuitry should be tied to the power ground as the output. If these three areas are properly laid out, the rest of the circuit should not have any other EMI problems and the power supply will be relatively free of noise.







19-Feb-2015

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TPS2832D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		2832	Samples
TPS2832DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		2832	Samples
TPS2832DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		2832	Samples
TPS2832DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		2832	Samples
TPS2833D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		2833	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



PACKAGE OPTION ADDENDUM

19-Feb-2015

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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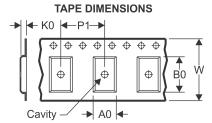
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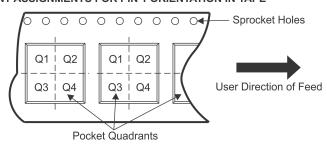
TAPE AND REEL INFORMATION





Α0	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

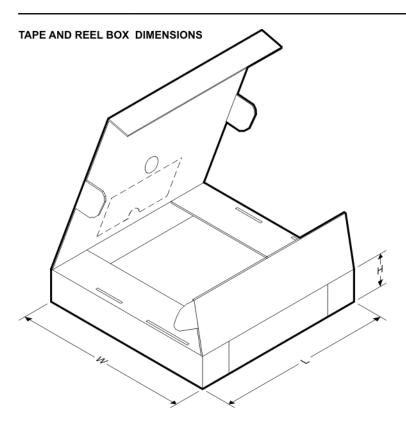
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

7 ili dimendene die nemina												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2832DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS2832DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
TPS2832DR	SOIC	D	8	2500	340.5	338.1	20.6	
TPS2832DR	SOIC	D	8	2500	367.0	367.0	38.0	

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