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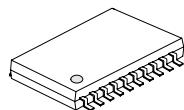
TEA2025D

LINEAR INTEGRATED CIRCUIT

STEREO AUDIO AMPLIFIER

■ DESCRIPTION

The UTC TEA2025D is a monolithic integrated circuit that intended for use as dual or bridge power audio amplifier portable radio cassette players.



SOP-20

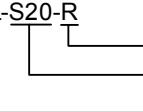
■ FEATURES

- * Dual or bridge connection modes.
- * Few external components.
- * Supply voltage down to 3V.
- * High channel separation.
- * Very low switch on\off noise.
- * Max gain of 45dB with adjust external resistor.
- * Soft clipping.
- * Thermal protection.
- * $3V < V_{CC} < 15V$
- * $P = 2 * 1W, V_{CC} = 6V, R_L = 4\Omega$
- * $P = 2 * 2.3W, V_{CC} = 9V, R_L = 4\Omega$
- * $P = 2 * 0.1W, V_{CC} = 3V, R_L = 4\Omega$

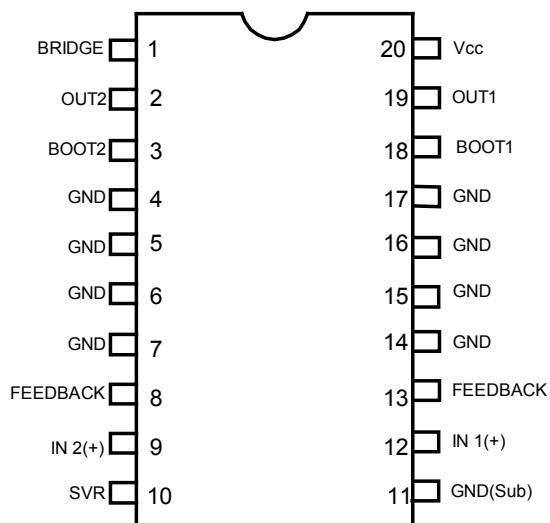
*Pb-free plating product number: TEA2025DL

■ ORDERING INFORMATION

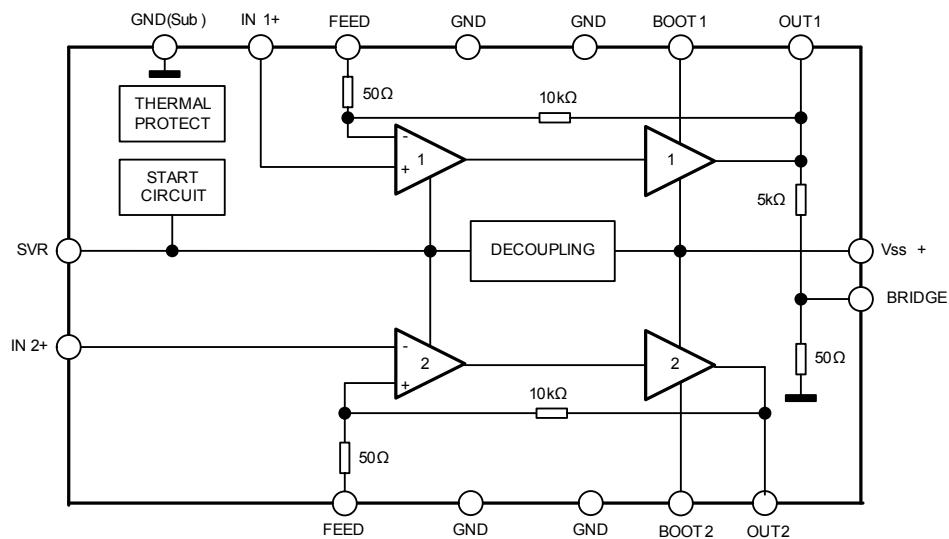
Order Number		Package	Packing
Normal	Lead Free Plating		
TEA2025D-S20-R	TEA2025DL-S20-R	SOP-20	Tape Reel
TEA2025D-S20-T	TEA2025DL-S20-T	SOP-20	Tube

TEA2025DL-S20-R 	(1)Packing Type (2)Package Type (3)Lead Plating	(1) R: Tape Reel, T: Tube (2) S20: SOP-20 (3) L: Lead Free Plating, Blank: Pb/Sn
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■ PIN CONFIGURATION



■ BLOCK DIAGRAM



■ ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	RATINGS	UNIT
Supply Voltage	V _{SS}	15	V
Output Peak Current	I _{O(Peak)}	1.5	A
Junction Temperature	T _J	150	°C
Storage Temperature	T _{STG}	150	°C

Note Absolute maximum ratings are those values beyond which the device could be permanently damaged.
Absolute maximum ratings are stress ratings only and functional device operation is not implied.

■ THERMAL DATA

PARAMETER	SYMBOL	RATINGS	UNIT
Thermal Resistance Junction-case	θ _{JC}	15	°C/W
Thermal Resistance Junction-ambient	θ _{JA}	65	°C/W

Note: The θ_{JA} is measured with 4 cm² copper area heatsink.

■ ELECTRICAL CHARACTERISTICS (Ta=25°C, V_{CC}=9V, Stereo unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
Supply Voltage	V _{SS}			3		12	V	
Quiescent Current	I _Q				35	50	mA	
Quiescent Output Voltage	V _{OUT}				4.5		V	
Voltage Gain	G _V	Stereo		43	45	47	dB	
		Bridge		49	51	53	dB	
Voltage Gain Difference	ΔG _{V(DIFF)}					±1	dB	
Input Impedance	R _I				30		kΩ	
Output Power (d=10%)	R _L =4Ω	P _{OUT}	Stereo 8 (per channel)	V _{SS} =3V	0.1			
					0.02			
				V _{SS} =6V	0.7	1		
					0.6			
					0.25			
					0.13			
				V _{SS} =9V	1.7	2.3		
					1.3			
Output Power (d=10%)	R _L =8Ω	P _{OUT}	Bridge	V _{SS} =12V	2.4			
					0.18			
				V _{SS} =3V	0.06			
					2.8			
					1.5			
					4.7			
				V _{SS} =6V	0.3	1.5	%	
					0.5		%	
Distortion	Stereo	d	V _{SS} =9V; R _L =4Ω					
	Bridge							
Supply Voltage Rejection	SVR	f=100Hz, V _R =0.5V, R _g =0		40	46		dB	
Input Noise Voltage	eN _(IN)	R _g =0			1.5	3	mV	
		R _g =104Ω			3	6	mV	
Cross-Talk	CT	f=1KHz, R _g =10KΩ		40	52		dB	

■ APPLICATION INFORMATION

Input Capacitor

Input capacitor is PNP type allowing source to be referenced to ground.

In this way no input coupling capacitor is required. However, a series capacitor(0.22μF) to the input side can be useful in case of noise due to variable resistor contact.

Bootstrap

The bootstrap connection allows to increase the output swing.

The suggested value for the bootstrap capacitors (100μF) avoids a reduction of the output signal also at low frequencies and low supply voltage.

Voltage Gain Adjust

STEREO MODE

The voltage gain is determined by on-chip resistors R1 and R2 together with the external RfC1 series connected between pin 8(13) and ground. The frequency response is given approximated by:

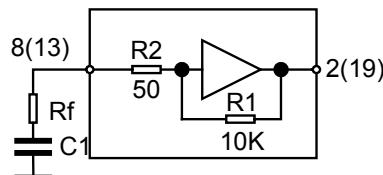
$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf + R2 + \frac{1}{JWC1}}$$

With Rf=0,C1=100μF, the gain results 46dB with pole at f=32Hz

The purpose of Rf is to reduce the gain. It is recommended to not reduce it under 36dB.

BRIDGE MODE

Figure 7



The bridge configuration is realized very easily thanks to an internal voltage divider which provides (at pin 1)the CH1 output signal after reduction. It is enough to connect pin8(inverting input of CH 2)with a capacitor to pin 1 and to connect to ground the pin 9.

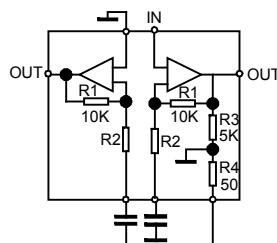
The total gain of the bridge is given by:

$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf + R2 + \frac{1}{JWC1}} \left(1 + \frac{R3}{R4} \frac{R1}{R2 + R4 + \frac{1}{JWC1}} \right)$$

and with the suggested values (C1=C2=100μF,Rf=0)means:

Gv=52dB

Figure 8



With first pole at f=32Hz

■ APPLICATION INFORMATION(Cont.)

Output Capacitors.

The low cut off frequency due to output capacitor depending on the load is given by:

$$F_L = \frac{1}{2\pi C_{OUT} * R_L}$$

With C_{OUT} 470μF and $R_L=4$ ohm it means $F_L=80$ Hz.

Stability

A good layout is recommended in order to avoid oscillations. In general, the designer must pay attention on the following points:

- Short wires of components and short connections.
- No ground loops.
- Bypass of supply voltage with capacitors as nearest as possible to the supply I.C. pin. The low value (polyester) capacitors must have good temperature and frequency characteristics.
- No sockets.

The heatsink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of excessive junction temperature: all that happens is that P_{OUT} (and therefore P_D) and I_d are reduced.

APPLICATION SUGGESTION

The recommended values of the components are those shown on stereo application circuit of Fig.2 different values can be used, the following table can help the designer.

Component	Recommended	Purpose	Larger Than	Smaller Than
C1,C2	0.22μF	Input DC Decoupling in Case of Slider Contact Noise of Variable Resistor		
C3	100μF	Ripple Rejection		Degradation of SVR, Increase of THD at Low Frequency and Low Voltage
C4,C5	100μF	Bootstrap		
C6,C7	470μF	Output DC Decoupling		Increase of Low Frequency Cut-off
C8,C9	0.15μF	Frequency Stability		Danger of Oscillations
C10,C11	100μF	Inverting Input DC Decoupling		Increase of Low Frequency Cut-off

■ TYPICAL APPLICATION CIRCUIT

Figure 1:Bridge Application(Powerdip)

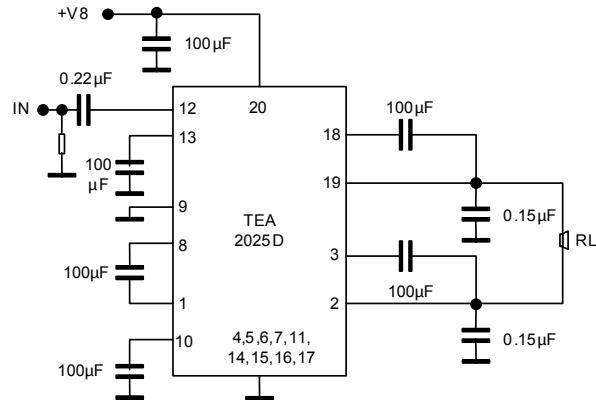
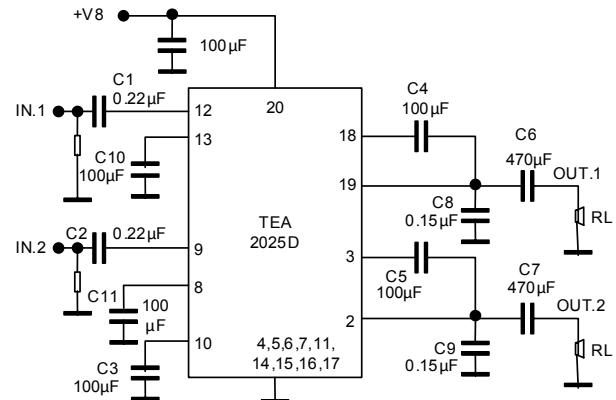
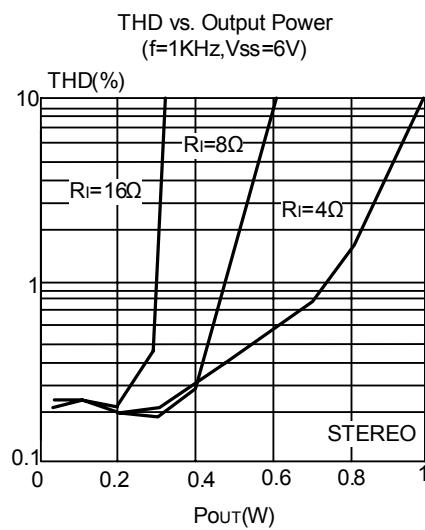
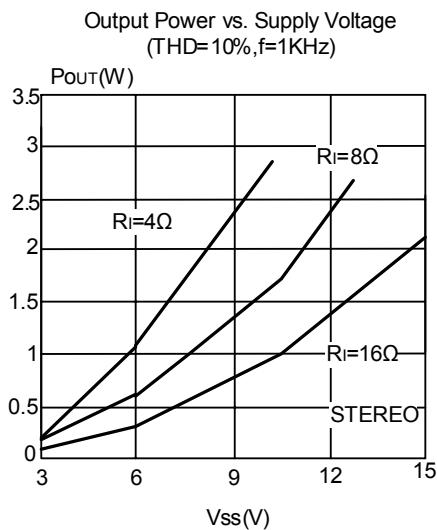
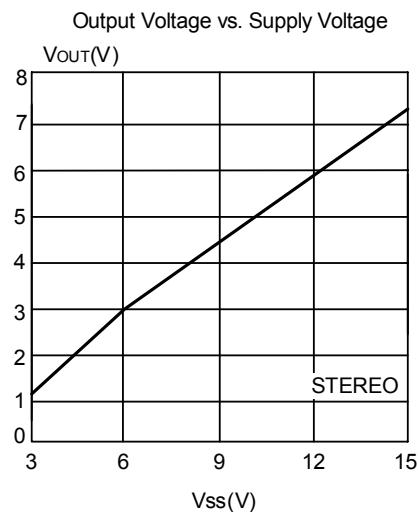
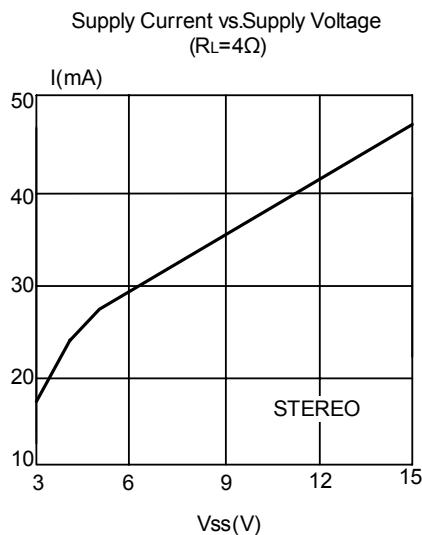


Figure 2: Stereo Application (Powerdip)



■ TYPICAL CHARACTERISTICS



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