

# **ICs for Communications**

Signal Processing Subscriber Line Interface Codec Filter SLICOFI<sup>®</sup>

PEB 3065 Version 3.2 PEF 3065 Version 3.2

Data Sheet 01.98

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#### **General Description**

#### 1 General Description

The Signal Processing Subscriber Line Interface Codec Filter SLICOFI (PEB 3065/PEF 3065) is a logic continuation of the well established family of the SIEMENS PCM-Codec-Filter-IC's with the vertical integration of all DC-feeding, Supervision and Meterpulse Injection features on chip as well. Fabricated in a standard 1  $\mu$ m BiCMOS technology the SLICOFI is tailored for very flexible solutions in digital communication systems.

For the first time the SLICOFI uses the benefits of a DSP not only for the voice channel but even for line feeding and supervision which leads to a very high flexibility without the need for external components.

Based on an advanced digital filter concept, the PEB 3065/PEF 3065 provides excellent transmission performance. The new filter concept (second generation in SIEMENS-Codec-family) leads to a maximum of independence between the different filter blocks. Each filter block can be seen as a one to one representative of the corresponding network element. Together with the software package SLICOS, filter optimizing to different applications can be done in a clear and straight forward procedure. The AC frequency behavior is mainly determined by the digital filters. Using the new oversampling 1 bit-AD/DA converter, linearity is only limited by second order parasitic effects.

The new - digital - solution of line feeding offers free programmability of feeding current and voltage as well as very fast settling of the dc-operating point after transitions. A 0.3 Hz lowpass filter in the DC-loop is mainly responsible for the system stability.

Additionally teletax generation and filtering is implemented as well as free programmable (balanced) ring generation with zero-crossing injection. Offhook detection with programmable thresholds is possible in all operating modes. To reduce overall power consumption of the line card, the SLICOFI provides a special mode called Power Denial where Offhook is done via 2 high voltage inputs ( $V_{\text{LINE1}}$  and  $V_{\text{LINE2}}$ ) directly connected to the line if the HV-SLIC is switched off.

## SIEMENS

### Signal Processing Subscriber Line Interface Codec Filter SLICOFI<sup>®</sup>

### Data Sheet for the Version 3.2

#### 1.1 Features

- Single chip CODEC and FILTER including all LOW VOLTAGE SLIC functions
- Only few external components required
- No trimming or adjustments required
- Specification according to relevant CCITT, LSSGR and DBP recommendations
- Digital signal processing technique
- Advanced low power 1 μm BiCMOS<sup>1</sup>) technology
- PCM encoded digital voice transmission (A-Law or μ-Law)
- Four pin serial IOM-2 Interface
- Standard P-LCC-44 package
- High performance AD and DA Conversion
- Programmable digital filters for
  - Impedance matching
  - Transhybrid balancing
  - Frequency response
  - Gain
- Advanced test capabilities
  - Integrated line and circuit tests
  - Two programmable tone generators
- Optimized HV-SLIC Interface
- Fully digital programmable DC-Characteristic
  - Programmable Constant Current from 0-70 mA
  - Programmable Resistive Values from 0-2  $\times$  500  $\Omega$
- Programmable Integrated Teletax Injection and Filtering during Conversation and Onhook
  - Programmable up to 125 mVrms (5 Vrms at ab-wire)
  - Programmable frequency 12/16 kHz

<sup>1)</sup> Abbreviations see **chapter 10.4**.

Туре	Package
PEB 3065N V3.2	P-LCC-44 / Tube
PEB 3065N V3.2	P-LCC-44 / Tape in Real
PEF 3065N V3.2	P-LCC-44 / Tape in Real



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PEB 3065 PEF 3065

**CMOS** 

### **General Description**

- Polarity reversal (programmable soft or hard)
- Integrated (balanced) Ringing Generation with zero crossing injection
  - Programmable frequency between 16.6 and 70 Hz (up to 300 Hz for test)
     Programmable amplitude up to 2.125 Vrms (85 Vrms at ab-wire)
- Four operating modes: Power Denial, Power Down, Active and Ringing
- Offhook detection with programmable thresholds for all operating modes
- Integrated Ring Trip Detection with zero crossing turn off function
- Ground Start and Loop Start possible
- Integrated checksum Calculation for CRAM
- Line Card Identification

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PEB 3065 PEF 3065

### **Pin Configuration**

### 2 Pin Configuration



### Figure 1

### **Pin Configuration**

### 2.1 Pin Definition and Functions

The following tables group the pins according to their functions. They include pin number, pin name, type, a brief description of the function, and cross-references referring to the sections in which the pin functions are discussed.

#### Table 1

Pin No.	Name	Туре	Function	Reference
27	GNDA	_	Analog Ground	chapter 9.1.1
1	GNDD	_	Digital Ground	chapter 9.1.1
34	$V_{DDA}$	_	+ 5 V Analog Supply Voltage	chapter 9.1.1
2	$V_{\rm DDD}$	_	+ 5 V Digital Supply Voltage	chapter 9.1.1
33	V <sub>SS</sub>	_	- 5 V Analog Supply Voltage	chapter 9.1.1

### Table 2IOM<sup>®</sup>-2 Pins

Pin No.	Name	Туре	Function	Reference
6	DU	0	IOM-2 Data Upstream	chapter 4
5	DD	Ι	IOM-2 Data Downstream	chapter 4
4	DCL	Ι	IOM-2 Data-Clock	chapter 4
3	FSC	Ι	IOM-2 Frame-Sync.	chapter 4
43	TS0	Ι	Time Slot selection Pin 0	chapter 4
42	TS1	Ι	Time Slot selection Pin 1	chapter 4
41	TS2	I	Time Slot selection Pin 2	chapter 4
40	SEL24	I	Select DCL = 2 or 4 MHz	chapter 4

Table 3 Interface to HV-SLIC

Pin No.	Name	Туре	Function	Reference
25	$V_{BIM}$	I	Battery Image Input	chapter 7
28	PDN	0	Set the HV-SLIC to Power Denial	chapter 7
19	IT	1	Transversal Current Input (AC + DC)	chapter 7
21	ITAC	1	Transversal Current Input (for AC)	chapter 7
22	GNDIT	1	Analog Ground	chapter 7
15	IL	1	Longitudinal Current Input	chapter 7

### **Pin Configuration**

Pin No.	Name	Туре	Function	Reference
26	$V_{2W}$	0	Two Wire Output Voltage	chapter 7
9	C1	0	Ternary Interface to HV-SLIC	chapter 7
10	C2	0	Ternary Interface to HV-SLIC	chapter 7
11	V <sub>LINE 1</sub>	I	Offhook-Detection in Power Denial Mode	chapter 7
12	V <sub>LINE 2</sub>	I	Offhook-Detection in Power Denial Mode	chapter 7

### Table 3 Interface to HV-SLIC (cont'd)

#### Table 4 IO Pins

Pin No.	Name	Туре	Function	Reference
7	IO1	I/O	User-Programmable I/O Pin	chapter 5.6
8	102	I/O	User-Programmable I/O Pin	chapter 5.6
38	11	Ι	Fixed Input Pin	chapter 5.6
39	01	0	Fixed Output Pin	chapter 5.6

#### Table 5 Miscellaneous Function Pins

Pin No.	Name	Туре	Function	Reference Values
36	RES	I	Reset	chapter 6.1
30	CAP	I	External Capacitor to GNDA	68 nF 5%
23	RREF	I	External Resistor to GNDA	30 k 1%
29	REXT	I	External Ring Sync. Input	chapter 6.6
31	ID-L	I	External Identification (Pin strapping)	chapter 10.2
32	ID-M	I	External Identification (Pin strapping)	chapter 10.2
35	ID-H	I	External Identification (Connect ASIC)	chapter 10.2
20	TE3	0	Test Pin, mustn't be connected	-
24	TE1	_	Test Pin (Not connected)	-
44	TE2	0	Test Pin, mustn't be connected	-

### **Pin Configuration**

Table 6		Pins not Us		
1	Din No	Namo	Type	Eunction

Pin No.	Name	Туре	Function	Reference
13	RESERVED	_	Reserved (not connected)	_
37	RESERVED	0	Reserved test pin, mustn't be connected	_
14	N.C.	_	Not connected (not used)	_
16	N.C.	_	Not connected (not used)	-
17	N.C.	_	Not connected (not used)	-
18	N.C.	_	Not connected (not used)	_

### SLICOFI<sup>®</sup> Principles

### 3 SLICOFI<sup>®</sup> Principles

Five Oversampling AD/DA converters are necessary for data conversion to gain the aspired programmability in the DSP. Generally the SLICOFI can be divided between the AC-Loop which is handling the voice and additionally teletax and the DC-Loop for line feeding, ringing injection and supervision.



### 3.1 SLICOFI<sup>®</sup> Signal Flow Graph: AC

### Figure 2

#### **Transmit Path**

The analog input signal has to be connected to pin 21 (ITAC) by an external capacitor (680 nF - 1  $\mu$ F) for AC/DC separation. After passing a simple initializing prefilter (PREFI) the voice signal is converted to a 1-bit digital data stream in the  $\Sigma\Delta$ -converter. The first down sampling steps are done in fast running digital hardware filters. The following steps are implemented in the micro code which has to be executed by the central Digital Signal Processor. This DSP-machine is able to handle the workload for the DC-loop as well. At the end the fully processed signal (flexibly programmed in many parameters) is transferred to the IOM-2 Interface in a PCM-compressed signal representation.

### **Receive Path**

The digital input signal is received via the IOM-2 Interface. Expansion, PCM-lowpass-filtering, gain correction and frequency response correction are the next

### SLICOFI<sup>®</sup> Principles

steps which are done by the DSP-machine. The up sampling interpolation steps are again processed by fast hardware structures to reduce the DSP-workload. The upsampled 1-bit data stream is then converted to an analog equivalent which is smoothed by a POST-Filter (POFI). At the summing point the values of the TTX-Generator and the DC-loop are added and then transferred to the output pin 26  $(V_{2W})$ .

### Loops

There are two different loops implemented: The Impedance Matching (IM) loop which is divided in 3 separate loops to guarantee very high flexibility to various impedances, and the Transhybrid Balancing (TH) loop.



### 3.2 SLICOFI<sup>®</sup> Signal Flow Graph: DC

### Figure 3

### DC Characteristic

The incoming information at pin IT (scaled transversal current (AC + DC) transferred to a voltage via a resistor) is first lowpass filtered (0.3 Hz) for stability and noise reasons and then fed into the DC-characteristic block. This consists of two branches which represents different kinds of feeding behavior. In typical applications it acts as a programmable constant current source ( $R_{in} > 30$  k). If the desired value cannot be held

### SLICOFI<sup>®</sup> Principles

feeding switches automatically and smooth to the resistive branch ( $R_{in} > 0-1$  k). For superposing voice as well as Teletax pulses the necessary drop at the line can be calculated and taken into account as well. The outgoing DC-feeding value - superposed with the AC-Loop result at the summing point is transferred to pin 26 ( $V_{2W}$ ).

### Supervision

The HOOK-information is the most important one and the SLICOFI provides this information via CIDU (see **chapter 5.6**), in all operating modes:

For Power Denial via 2 high voltage input pins ( $V_{\text{LINE}}$ ) directly connected to the line.

For each other mode the line current information (from pin IT) is transferred via an ADC to the DSP where the Offhook information is extracted in the proper way:

Power Down: Offhook is detected if Constant current feeding is possible.

- Active: Offhook is detected if the incoming voltage at IT exceeds a programmed value. To avoid instable information, lowpass filtering and a hystereses is provided (2 independent programmable values for Offhook and Onhook detection).
- Ringing: Ring Trip occurs if the DC-value at IT exceeds the programmed Ring Trip threshold. The AC-value is filtered by the SLICOFI automatically. Ring Trip detection is reported within 2 cycles of the ring period and then the internal ring generator is switched off within 3 cycles at zero crossing of the ring voltage.

Ground key (CIDU-6: GNK) is reported if the absolute value of the voltage at pin IL exceeds 255 mV. With a programmable lowpass filter (see **chapter 5.6**) interfering frequencies (e.g. power lines with 50/60 Hz) can be filtered very effectively.

### 3.3 Test Features

The SLICOFI provides two different kinds of test features: Internal test loops for circuit testing and defined test loops to perform board and line tests. There are loops for testing AC and DC path. As a special feature it is possible to switch signals to and from the DC-path via the IOM-2 Interface. Additionally there is the possibility to cut off the AC-receive and transmit path.

(The different kinds of testmodes are described in **chapter 10.3**)

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PEB 3065 PEF 3065

### **SLICOFI<sup>®</sup> Principles**





Figure 4

### IOM<sup>®</sup>-2 Interface

### 4 IOM<sup>®</sup>-2 Interface

The IOM-2 interface consists of two data lines and two clock lines. DU (data upstream) carries data from the SLICOFI to a master device. DD (data downstream) carries data from the master device to the SLICOF. A frame synchronization clock signal (8 kHz, FSC) as well as a data clock signal (2048 kHz or 4096 kHz, DCL) has to be supplied to the SLICOFI. The SLICOFI handles data as described in the IOM-2 specification for analog devices.



Figure 5 IOM<sup>®</sup>-2 Interface Timing for 8 voice channels (per 8 kHz frame)

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### IOM<sup>®</sup>-2 Interface









Semiconductor Group

### IOM<sup>®</sup>-2 Interface

### IOM<sup>®</sup>-2 Time Slot Assignment

An assignment of 8 time slots is possible for each voice-channel. The IOM-2 operating mode and time slot selection is set completely by pin-strapping.

i apie i	Та	bl	е	7
----------	----	----	---	---

SEL24	TS2	TS1	TS0	IOM <sup>®</sup> -2 Operating Mode
0	0	0	0	Time slot 0; DCL = 2048 kHz
θ	θ	θ	4	Time slot 1; DCL = $2048 \text{ kHz}^{1}$
θ	θ	4	θ	Time slot 2; DCL = 2048 kHz <sup>1)</sup>
θ	θ	1	1	Time slot 3; DCL = 2048 kHz <sup>1)</sup>
0	1	0	0	Time slot 4; DCL = 2048 kHz
θ	4	θ	1	Time slot 5; DCL = 2048 kHz <sup>1)</sup>
0	1	1	0	Time slot 6; DCL = 2048 kHz
0	1	1	1	Time slot 7; DCL = 2048 kHz
1	0	0	0	Time slot 0; DCL = 4096 kHz
1	0	0	1	Time slot 1; DCL = 4096 kHz
1	0	1	0	Time slot 2; DCL = 4096 kHz
1	0	1	1	Time slot 3; DCL = 4096 kHz
1	1	0	0	Time slot 4; DCL = 4096 kHz
1	1	0	1	Time slot 5; DCL = 4096 kHz
1	1	1	0	Time slot 6; DCL = 4096 kHz
1	1	1	1	Time slot 7; DCL = 4096 kHz

<sup>1)</sup> Time slots 1, 2, 3 and 5 are not working with DCL = 2048 kHz.

For a workaround in the 2MHz mode please contact the SIEMENS HL Application group.

### 5 Programming the SLICOFI<sup>®</sup>

With the appropriate commands, the SLICOFI can be programmed and verified very flexible via the IOM-2 Interface monitor channel.

Data transfer to the SLICOFI starts with a SLICOFI-specific address byte  $(81_{H})$ .

With the second byte one of 3 different types of commands (SOP, TOP or COP) is selected. SOP and COP can be used as a write or read command, the TOP-Command is used for reading only. Due to the extended SLICOFI feature control facilities, SOP, COP and TOP commands contain additional information (e.g. number of subsequent bytes) for programming (write) and verifying (read) the SLICOFI status.

A write command is followed by up to 8 bytes of data. The SLICOFI responds to a read command with its IOM2 specific address and the requested information, that is up to 15 bytes of data (see **chapter 5.2**).

**Attention:** Each byte on the monitor channel has to be sent twice at least according to the IOM2 Monitor handshake procedure. (For more information on IOM-2 specific Monitor Channel Data Structure see **chapter 10**).

### 5.1 Types of Monitor Bytes

(x... don't care)

The 8-bit Monitor bytes have to be interpreted as either commands or status information stored in Configuration Registers or the Coefficient Ram. There are three different types of SLICOFI commands which are selected by bit 2 and 3 as shown below.

SOP	Status	Operatio	n:	SLICO	OFI status	setting/mo	onitoring	
Bit	7	6	5	4	3	2	1	0
					0	1		
ТОР	Transf	er Operat	ion:	Read	Certain S	tatus Optic	ons only	
Bit	7	6	5	4	3	2	1	0
					1	1		
СОР	Coeffic	cient Ope	ration:	filter c	oefficient	setting/mc	onitoring	
Bit	7	6	5	4	3	2	1	0
					Х	0		

### Storage of programming information:

8 (9) status configuration registers:

(SCR0), SCR1, ... SCR8 accessed by SOP commands

8 test configuration registers:	STCR1STCR8 accessed by SOP commands
18 Transfer configuration registers:	TCR1, TCR2TCR18 accessed by TOP commands
1 Coefficient RAM:	CRAM accessed by COP commands

### 5.2 SLICOFI<sup>®</sup> Programming Procedure

(DD... Data Downstream, DU... Data Upstream, only the Monitor Bytes are considered)

### **SOP– Write Commands**

DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0		DU
Address	1	0	0	0	0	0	0	1					ld	le					
SOP-Write 0 Byte		0			0	1	0	0					ld	le					
	•	•		•	•				,									-	
DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0	-	DU
Address	1	0	0	0	0	0	0	1					ld	le					
SOP-Write 2 Bytes		0			0	1	0	1					ld	le					
SCR1				Da	ata								ld	le					
SCR2				Da	ata								ld	le					
																		_	
DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0	-	DU
Address	1	0	0	0	0	0	0	1					ld	le					
SOP-Write 8 Bytes		0			0	1	1	0					ld	le					
SCR1				Da	ata								ld	le					
:					:									1				-	
SCR8				Da	ata								ld	le					
									J									-	
DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0	_	DU
Address	1	0	0	0	0	0	0	1					ld	le					
SOP-Write 8 Bytes		0			0	1	1	1					ld	le					
STCR1				Da	ata								ld	le				-	
:					:														
STCR8				Da	ata								ld	le					

#### **TOP – Write Commands**

no write command possible; reading only.

#### **COP – Write Commands**

7 6 5 4 3 2 1 0 Bit 7 6 5 4 3 2 1 0 DD DU 0 0 0 0 0 1 1 Address Idle **COP-Write 8 Bytes** Idle 0 0 0 Coeff. 1 Data Idle : : : Coeff. 8 Data Idle

#### **SOP** – Read Commands

DD	7	6	5	4	3	2	1	0	Bit	7 6 5 4 3 2 1 0 DU
Address	1	0	0	0	0	0	0	1		Idle
SOP-Read 1 Byte		1			0	1	0	0		Idle
			•	Id	lle	•				1 0 0 0 0 0 0 1 Address
				ld	lle					Data SCR0
DD	7	6	5	4	3	2	1	0	Bit	7 6 5 4 3 2 1 0 DU
Address	1	0	0	0	0	0	0	1		Idle
SOP-Read 3 Bytes		1			0	1	0	1		Idle
				ld	lle					1 0 0 0 0 0 0 1 Address
				ld	lle					Data SCR0
				ld	lle					Data SCR1
				ld	lle					Data SCR2
DD	7	6	5	4	3	2	1	0	Bit	7 6 5 4 3 2 1 0 DU
Address	1	0	0	0	0	0	0	1		Idle
SOP-Read 9 Bytes		1				1	1	0		Idle
				ld	lle					1 0 0 0 0 0 0 1 Address
				ld	lle					Data SCR0
										: :
				ld	lle					Data SCR8

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## Programming the SLICOFI®

DD	7	6	5	4	3	2	1	0	Bit	7 6 5 4 3 2 1 0 DU
Address	1	0	0	0	0	0	0	1		Idle
SOP-Read 8 Bytes		1				1	1	1		Idle
				ld	lle					1 0 0 0 0 0 0 1 Address
				ld	lle					Data STCR1
		:								: :
	ldle									Data STCR8

### **TOP – Read Commands**

DD	7	6	5	4	3	2	1	0	Bit	7 6 5 4 3 2 1 0 DU
Address	1	0	0	0	0	0	0	1		Idle
TOP-Read 1 Byte		1			1	1	0	0		Idle
				ld	le					1 0 0 0 0 0 0 1 Address
		Idle								Data TCR1

DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0		DU
Address	1	0	0	0	0	0	0	1					lc	lle					
TOP-Read 3 Bytes		1			1	1	0	1					lc	lle					
				ld	le					1	0	0	0	0	0	0	1	Address	
				ld	le								Da	ata				TCR1	
				ld	le								Da	ata		TCR2			
				ld	le					Data								TCR3	
									-										
DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0		DU

88	•	Ŭ	Ŭ	•	Ŭ	-	•	0	Dir	•
Address	1	0	0	0	0	0	0	1		
TOP-Read 15 Bytes		1			1	1	1	0		
		•		ld	le					1
				ld	le					
				ld	le					

	20
Idle	
Idle	
1 0 0 0 0 0 1	Address
Data	TCR4
:	:
Data	TCR18

### **COP** – Read Commands

DD	7	6	5	4	3	2	1	0	Bit	7	6	5	4	3	2	1	0		DU
Address	1	0	0	0	0	0	0	1					ld	le					
COP-Read 8 Bytes		1	0			0							ld	le					
				ld	le					1	0	0	0	0	0	0	1	Address	
				ld	le						Data							Coeff.	1
																		:	
				ld	le								Da	ata				Coeff.	8

### Example for a Mixed Command

DD	765432	1 0	Bit	7 6 5 4 3 2 1 0	DU
Address	1 0 0 0 0 0	0 1		ldle	
SOP-Write 2 Bytes	0 0 1	0 1		ldle	
SCR1	Data	<u> </u>		ldle	
SCR2	Data			ldle	
COP-Write 8 Bytes	0 0 0			ldle	
Coeff. 1	Data			ldle	
Coeff. 2	Data			ldle	
Coeff. 3	Data			ldle	
Coeff. 4	Data			ldle	
Coeff. 5	Data			ldle	
Coeff. 6	Data			ldle	
Coeff. 7	Data			ldle	
Coeff. 8	Data			ldle	
SOP-Read 3 Bytes	1 0 1	0 1		ldle	
	Idle			1 0 0 0 0 0 0 1	Address
	Idle			Data	SCR0
	Idle			Data	SCR1
	Idle			Data	SCR2
Address	1 0 0 0 0 0	0 1		ldle	
COP-Read 8 Bytes	1 0 0			ldle	
	Idle			1 0 0 0 0 0 0 1	Address

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DD	7	6	5	4	3	2	1	0	Bit	7 6 5 4 3 2 1 0	DU
		ldle								Data	Coeff. 1
				ld	le					Data	Coeff. 2
				ld	lle					Data	Coeff. 3
		Idle								Data	Coeff. 4
		ldle								Data	Coeff. 5
		ldle								Data	Coeff. 6
				ld	le					Data	Coeff. 7
				ld	lle					Data	Coeff. 8
Address	1	0	0	0	0	0	0	1		Idle	
TOP-Read 1 Byte		1			1	1	0	0		Idle	
	Idle							1 0 0 0 0 0 0 1	Address		
				ld	le				]	Data	TCR1

### 5.3 SOP Command

To modify or evaluate the SLICOFI status, the contents of up to 8 configuration registers SCR1, ... SCR8 may be transferred to, or up to 9 (incl. SCR0) from the SLICOFI. This is done by a SOP-Command (status operation command).

With LSEL = 11 some test registers can be set/read (for internal use only!).

The two commands POLNR and RST are only valid if RW = 0 (write); they are ignored for RW = 1 (read)

Bit	7	6	5	4	3	2	1	0					
	0	RW	POLNR	RST	0	1	LSEL1	LSEL0					
RW													
POL	NR	General E POLNR = POLNR =	0 sets th	ne SLICOF	I to Norma	or Revers al Polarity se Polarity	feeding						
RST		Software RST = 0 RST = 1	Norma Reset	eset Normal Operation Reset SLICOFI (same as Reset pin 36 (RES)): sets the SLICOFI to the basic setting mode (see <b>chapter 6.1</b> ).									
LSEI													

### SCR0 Configuration Register 0

Configuration Register SCR0 can be read only. It gives a mirror of the SOP-Command itself to control its contents and represents the reset value as defined below.

Bit	7	6	5	4	3	2	1	0
	0	1	POLNR	RSTST	0	1	LSEL1	LSEL0

Reset value: 54<sub>H</sub> (if only SCR0 is read. It depends on LSEL1 and LSEL0.)

POLNR	General DC feeding Information: Normal or Reverse Polarity POLNR = 0 indicates, that the SLICOFI was set to Normal Polarity feeding						
	POLNR = 1	indicates, that the SLICOFI was set to Reverse Polarity feeding <sup>1)</sup>					
RSTST	Status of Res	set					
	Indicates the	occurrence of a reset:					
	RSTST = 1	if there has been a Reset by any of the following three					
		reasons:					
		<ul> <li>via the Reset-pin (RES)</li> </ul>					
		<ul> <li>via the Power on Reset</li> </ul>					
		<ul> <li>via the Software Reset (SOP–Command)</li> </ul>					
		the RSTST-bit is set to '1'.					
	RSTST = 0	no Reset has occurred since the last SOP-Read (with LSEL = 00b).					
	This bit is cle	ared only by a SOP-read with LSEL = 00b at the end of the					
	data transmis	ssion.					
LSEL	is the mirror of	of the SOP-Read LSEL contents.					

<sup>&</sup>lt;sup>1)</sup> The internal manipulation with "Reverse meterpulses" is not indicated by that bit.

### **SCR1 Configuration Register 1**

Configuration register SCR1 defines the basic feeding modes of the SLICOFI and enables/disables test features:

Bit	7	6	5	4	3	2	1	0				
	PD	N/BB	LB	ETG1	HI-b	HI-a	DHP-X	COR				
Deee		0										
PD	t value: 0	••	a aat aith	or in Dowe	r Down o	r Dowor D	onial mad	a togothor				
FU				6,7 (see <b>c</b>			enial mode	e logelhei				
		PD = 0			• /		line superv	vision via				
		PD = 1	$V_{LINE1}$	, <sup>2</sup> DFI set to I	Power Dov	vn mode						
N/BB	6				ed Battery	mode (see	e chapter (	6 <b>.5</b> ).				
		N/BB = 0 Normal feeding										
		N/BB = 1	<ul> <li>B = 1 Changes ternary interface to HV-SLIC which sets the HV-SLIC to Boosted Battery mode</li> </ul>									
LB		Handling of	g of Loop Back functions for on chip test loops									
		LB = 0		al function	5			N .				
		LB = 1		ed (selecte		•	og or digita with the	al) is				
				t (SCR2-3)	•	o, logothol	with the					
ETG	1	Enables p	rogramma	ble Test T	one Gene	rator 1						
		ETG1 = 0		Tone Gene								
ᆈᇥ		ETG1 = 1		Fone Gene	rator 1 is e	enabled						
HI-b		For HV-SL HI-b = 0		al operation	n							
		HI-b = 1		•		to HV-SLI	C which se	ts the				
			•	of the line	into high ir	npedance	state					
HI-a		For HV-SL			-							
		HI-a = 0 HI-a = 1		al operation les ternarv		to HV-SLI	C which se	ts the				
				of the line								
DHP	-X					``	hapter 10.3	<b>B</b> )				
		DHP-X = ( DHP-X =		mit Highpa								
COR				Transmit Highpass Filter is disabled eive Path for test reasons (see <b>chapter 10.3</b> )								
		COR = 0		ve Path tra	•	-	,					
		COR = 1	Recei	ve Path is	disabled							

### SCR2 Configuration Register 2

Configuration register SCR2 defines some testmode output results, some special SLMA-mode requirements and the possibility to program 2 I/O-ports.

Bit	7	6	5	4	3	2	1	0					
	MVA	OKTON	OKTTX	OKRNG	ТМ	NOSL	IO1	IO2					
Deee													
	Reset value: 00 <sub>H</sub> (then as measured) <b>MVA</b> Internal measurement results shown in the following 3 bits are valid or												
IVIVA		Internal measurement results shown in the following 3 bits are valid or not valid (read only) (see <b>chapter 10.3</b> )											
		MVA = 0	• •	llowing 3 o		ts are not v	/alid						
		MVA = 1	the fo	llowing 3 o	k-bit resul	ts are valic	1						
ΟΚΤ	ON	Test Tone measurement information (read only) - programmed via COP-command (Testloop: DLB_4M and TG1 enabled, see <b>chapter 10.3</b> ) OKTON = 0 Test tone value out of defined range OKTON = 1 Test tone value in defined range											
окт	гх					•	- program	nmed via					
				e chapter	-	<b>,</b> ,,	1						
		OKTTX = OKTTX =		eletax met eletax met	•								
OKR	NG		mand (see 0 Ring t	informatio e <b>chapter</b> f cone value cone value	10.3) smaller th		value	nmed via					
ТМ		TM = 0	resets	the SLICO the assign	ned tests (	normal mo	ode)						
		TM = 1		he assigne 3-bit (SCR <sup>2</sup>	•		SCRO, log	einer with					
NOS	L	NOSL = 0	Slope	of TTX-Sig	gnal is sm		signal is sv	witched off					
104		NOSL = 1		switch of T	-								
IO1		Selection $IO1 = 0$		mmable IC he pin IO1		ıt							
		IO2 = 1		he pin IO1	-								
102		Selection	for progra	mmable IC	- Pin IO2								
		IO1 = 0		he pin IO2									
		IO2 = 1	sets t	he pin IO2	as an outp	DUT							

### **SCR3 Configuration Register 3**

Configuration register SCR3 defines the meterpulse settings and the Data Upstream Persistency Counter.

Bit	7	6	5	4	3	2	1	0
	TTXNO	TTX12	SOREV	PDADIS	DUP3	DUP2	DUP1	DUP0

#### Reset value: 8A<sub>H</sub>

ΤΤΧΝΟ	Meterpulses are represented by teletax (TTX) with 12 or 16 kHz or with Reverse Polarity TTXNO = 0 Meterpulses are represented with 12 kHz or 16 kHz TTXNO = 1 Meterpulses are represented with Reverse Polarity
TTX12	Teletax-signal with 12 kHz or 16 kHz TTX12 = 0   16 kHz teletax-signal TTX12 = 1   12 kHz teletax-signal
SOREV	The reversal pulse is either soft or hard SOREV = 0 hard reversal SOREV = 1 soft reversal

Note: For proper function special coefficients generated by SLICOS should be used.

To realize this function following settings must be done:

- 1. Enable the testregisters (Configuration Register 5: SCR5-1 (ENTR)=1), (page 32)
- 2. The testregisterblock must be load with STCR3-0 (SOFTVER) = 1, (see chapter 10.3)

### STCR3 Test Configuration Register 3

Bit	7	6	5	4	3	2	1	0
	0	0	0	0	0	0	0	1

### 3. SCR3-5 (SOREV) = 1

PDADIS The automatic HV-SLIC Power Down - Active switching (see chapter 6.4) can be switched off

PDADIS = 0 use automatic Power Down-Active switching

PDADIS = 1 disables automatic Power Down-Active switching

**DUP** To restrict the rate of upstream C/I-bit changes, deglitching (persistence checking) of the status information from the SLICOFI may be applied. New status information will be transmitted upstream, after it has been stable for N milliseconds. N is binary programmable in the range of 1 to 15 ms in steps of 1 ms; with DUP =  $0_H$  the deglitching is disabled. Reset value is 10 ms. The HOOK, SLCX and the I(O)-bits are influenced (different counters but

same programming).

Detailed info see chapter 5.4.

### **SCR4 Configuration Register 4**

Configuration register SCR4 defines the basic SLICOFI settings which enable / disable the programmable digital filters and the second tone generator.

Bit	7	6	5	4	3	2	1	0					
	TH	IM	FRX	FRR	AX	AR	ETG2	PTG					
Poso	t value: 0	n											
TH	t value. U	•••	whrid Bala	ancina Filte	er - toaeth	er with the	hit FIXC	(SCR5-5)					
		Set transhybrid Balancing Filter – together with the bit FIXC (SCR5-5). For FIXC = 1 the TH-Filter is set to $H_{TH}$ = for $Z_{BRD}$ ; for FIXC = 0: TH = 0 TH-filter is disabled TH = 1 TH-filter is enabled (use programmed values)											
		TH = 1				•	-						
ΙΜ		Set DSP-implemented Impedance Matching Filter - together with the bit FIXC (SCR5-5). For FIXC = 1 the IM-Filter is set to $H_{IM}$ = for 900; for FIXC = 0: IM = 0 IM-filter is disabled IM = 1 IM-filter is enabled (use programmed values)											
FRX		Enable FF FRX = 0 FRX = 1	FRX-f	ency Resp ilter is disa ilter is ena	abled (H <sub>FR</sub> )	<sub>(</sub> = 1)							
FRR		Enable FF FRR = 0 FRR = 1	FRR-	iency Resp filter is disa filter is ena	abled (H <sub>FRI</sub>	<sub>R</sub> = 1)							
AX		Set AX- (A AX = 0 AX = 1	AX-fil	on/Attenua ter is set to ter is enab	default va	alue (H <sub>AX</sub> =							
AR		Set AR- (A AR = 0 AR = 1	AR-fil	on/Attenua ter is set to ter is enab	o default va	alue (H <sub>AR</sub> =		3)					
ETG	2	Enable programmable Test Tone Generator 2 ETG2 = 0 Test Tone Generator 2 is disabled ETG2 = 1 Test Tone generator 2 is enabled											
PTG		User programmable frequency or fixed frequency is selected PTG = 0 fixed frequency for both Test Tone Generators TG1 = 1008 Hz, $TG2 = 2$ kHz											
		PTG = 1	progra	ammed fre	quency for	both Test	Tone Ger	erators					

### SCR5 Configuration Register 5

Configuration register SCR5 defines various different features.

Bit	7	6	5	4	3	2	1	0						
	DHP-R	LAW	FIXC	LIN	IDR	REXTEN	ENTR	0						
<b>_</b>		0												
Reset value: 20 <sub>H</sub>														
DHP-	-R	DHP-R =	Disable Receive Highpass for test reasons (see <b>chapter 10.3</b> ) DHP-R = 0 Receive Highpass Filter is enabled DHP-R = 1 Receive Highpass Filter is disabled											
LAW		PCM - law	selection											
		LAW = 0	A-Lav	v is selecte	ed									
		LAW = 1	μ-Law	/ (μ255 PC	CM) is sele	ected								
FIXC						nts or the pr	ogramme	d ones.						
		FIXC = 0 FIXC = 1	1 5	ammed co coefficient	efficients u	used								
		fixed coef												
LIN			•	•	,	nation in vo	ice chann	al A (unner						
		byte) and		•										
		LIN = 0		mode is s	elected									
		LIN = 1	linear	mode is s	elected									
IDR		Initialize D	ata RAM											
		IDR = 0		•	n is select									
		IDR = 1	conte	nts of Data	a RAM is s	set to 0 (for	test purpo	ses)						
REX	TEN	Ringing E												
		REXTEN												
	_		REXTEN = 1 used for external (unbalanced) ringing											
ENT	र	Enable Te		•		to a to a fith a								
		ENTR = 0		•		tents of the fault values	•	sters are						
		ENTR = 1	-	-		egisters ca		ged						

### **SCR6 Configuration Register 6**

Configuration register SCR6 defines various test features and test loops.

Bit	7	6	5	4	3	2	1	0
	COT8	COT16	OPIMAN	OPIM4M		TEST I	OOPS	

### Reset value: 00<sub>H</sub>

COT8	Cut Off Transmit Path at 8 kHz for test reasons (Input of Compression) COT8 = 0 transmit path transmission is enabled COT8 = 1 transmit path is disabled (output is zero for $\mu$ -law and linear mode, + 1 (= LSB) for A-law)					
COT16	Cut Off Transmit Path at 16 kHz for test reasons (Input of TH-Filter) COT16 = 0 transmit path transmission is enabled COT16 = 1 transmit path is disabled					
OPIMAN	Open analog Impedance Matching Loop (IMAN) OPIMAN = 0 normal operation OPIMAN = 1 opens analog IM-Loop ( $H_{IMAN} = 0$ )					
OPIM4M	Open fast digital Impedance Matching Loop (IM4M) OPIM4M = 0 normal operation OPIM4M = 1 opens fast digital IM-Loop ( $H_{IM4M} = 0$ )					
TEST LOOPS	4 bit field for various analog and digital test loops can be set together with LB and TM (see <b>chapter 10.3</b> , for detailed information).					

### SCR7 Configuration Register 7

Configuration register SCR7 is the Mask register. With it each bit of TCR1 (Signalling register) can be masked; that means changes of such a "masked bit" are not causing a change of the SLCX - bit (Data Upstream C/I-channel byte).

Bit	7	6	5	4	3	2	1	0
	HOOKM	GNKM	VB/2M	ICONM	TEMPM	CFAILM	1	1
Reset value: FF								

Reset value. I	ЧН	
НООКМ	Mask bit for C	Offhook information
	HOOKM = 0	each change of the HOOK bit leads to an interrupt (SLCX-bit)
	HOOKM = 1	changes of HOOK bit are neglected
GNKM	Mask bit for g GNKM = 0	round key information each change of the GNK bit leads to an interrupt (SLCX-bit)
	GNKM = 1	changes of GNK bit are neglected
VB/2M	Mask bit for h VB/2M = 0	alf battery information each change of the VB/2 bit leads to an interrupt (SLCX-bit)
	VB/2M = 1	
ICONM	Mask bit for c ICONM = 0	onstant current information each change of the ICON bit leads to an interrupt (SLCX-bit)
	ICONM = 1	changes of ICON bit are neglected
ТЕМРМ		ver temperature information each change of the TEMPA bit leads to an interrupt (SLCX-bit)
	TEMPM = 1	changes of TEMPA bit are neglected
CFAILM		lock fail information each change of the CFAIL bit leads to an interrupt (SLCX-bit)
	CFAILM = 1	changes of CFAIL bit are neglected
Information of	out chonging	half bettery and constant surrent information will be

Information about changing half battery- and constant current- information will be neglected on both of the Power Denial and the Ringing state, and information about changing ground key information will be neglected in the Power Denial state.

### **SCR8 Configuration Register 8**

Configuration register SCR8 defines some Test Mode Settings and the Ground Key/External Indication Data Upstream Persistency Counter.

Bit	7	6	5	4	3	2	1	0
	DCANAL	CHOPACT	DCHOLD	EXT_MCLK 1	DUPGNK3	DUPGNK2	DUPGNK1	DUPGNK0

### Reset value: 05<sub>H</sub>

DCANAL	Test bit to shorten internally the IT with the $V_{2W}$ pin DCANAL = 0 normal operation DCANAL = 1 the DC Analog Loop is closed
CHOPACT	Transforms DC-Test values to 500 Hz rectangular values at the PCM interface CHOPACT = 0 normal operation CHOPACT = 1 chopping function is activated
DCHOLD	Holds the actual DC-value at the $V_{2W}$ output DCHOLD = 0 normal operation DCHOLD = 1 hold DC-value at V2W
EXT_MCLK1	<ul> <li>External Masterclock (16 MHz)</li> <li>EXT_MCLK1 = 0 internal masterclock is used</li> <li>EXT_MCLK1 = 1 external masterclock is used</li> <li>To use an external masterclock of 16 MHz following steps must be done:</li> <li>1. IO1 must be set to input and becomes the input-pin of the masterclock (page 42)</li> <li>2. Connect the internal clockline to IO1 and disable the PLL by setting the bit EXT_MCLK1 = 1</li> </ul>
DUPGNK	To restrict the rate of upstream C/I-bit changes, deglitching (persistence checking) of the status information from the SLICOFI may be applied. New status information will be transmitted upstream, after it has been stable for N milliseconds. N is binary programmable in the range of 4 to 60 ms in steps of 4 ms, with DUPGNK = 0h the deglitching is disabled. Reset value is 20 ms. The HOOK bit (for external Indication) and the GNK bit are influenced. Detailed info see <b>chapter 5.6</b> .

### 5.4 TOP Command

If no status modification of the SLICOFI is required (there is no TOP-write operation) a transfer operation byte TOP may be transferred.

Bit	7	6	5	4	3	2	1	0
	0	RW	0	0	1	1	LSEL1	LSEL0

RW

Read Information: Enables reading from the SLICOFI RW = 0 No operation

RW = 1 Read from SLICOFI

LSEL Length select information (also see programming procedure, chapter 5.2). This field identifies the number of subsequent data bytes. LSEL = 00 Read TCR1 (Signalling Register) only LSEL = 01 Read 3 bytes of data (TCR1, TCR2, TCR3) LSEL = 10 Read extended line card design and configuration information only (TCR4, ... TCR18). Details see chapter 10.2 LSEL = 11 future reserved

### TCR1 Configuration Register 1

TCR1 is the Signalling register. It indicates status information. If there is any change of one or more bit, it is indicated via the SLCX bit in the C/I-channel. Each bit can be masked by SCR7 Register.

Bit	7	6	5	4	3	2	1	0
	HOOK	GNK	VB/2	ICON	TEMPA	CFAIL	Х	Х

Reset value: 00<sub>H</sub>

НООК	Loop information HOOK = $0$ HOOK = $1$	tion On/Offhook (same as in C/I-channel) Onhook Offhook		
GNK	Ground key or Ground start information via IL-pin (same C/I-channel) interrupt masked in Power Denial State GNK = 0 no longitudinal current detected			
	GNK = 1	longitudinal current detected		
Programming the SLICOFI <sup>®</sup>				
--------------------------------------				
--------------------------------------				

VB/2	Half battery voltage across the HV-SLIC is detected ( $V_{2W}$ compared to $V_{BIM}/2$ ) interrupt masked in Power Denial and Ringing State VB/2 = 0 line voltage smaller than half battery ( $ V_{2W}  >  V_{BIM}/2 $ ) VB/2 = 1 line voltage larger than half battery ( $ V_{2W}  <  V_{BIM}/2 $ )
ICON	Current limitation informationinterrupt masked in Power Denial and Ringing StateICON = 0Resistive FeedingICON = 1Constant Current Feeding
ΤΕΜΡΑ	Temperature alarm of the HV-SLIC which is signalled through the HV-SLIC Interface (see <b>chapter 7</b> ). TEMPA = 0 normal temperature TEMPA = 1 Temperature alarm from HV-SLIC detected
CFAIL	Clock Fail: Not the right count of clock cycles between two frame syncs CFAIL = 0 no clock fails detected CFAIL = 1 clock fails detected The CFAIL bit is not influenced by the DUP-counter (each failure is reported).
X	undefined

Any change of these bits is signalled via the interrupt-bit (SLCX) in the C/I-DU-channel. There are two types of generating an interrupt:

- Each toggling of a non-masked TCR1-bit combined with a DUP-counter

- Toggling of the non-masked CFAIL-bit (no filtering by the DUP-counter)

The status information is stored in the TCR1-register by an interrupt or - if there is no interrupt - before reading this register only.

The HOOK- and the GNK-input are directly filtered by an own DUP-/DUPGNK-counter too and they are also directly included in the C/I-DU-channel.

Reading the TCR1-register is possible in two ways:

- Reading only TCR1 (TOP-command with LSEL = 0b)
- Reading TCR1 with other TCR-registers (TOP-command with LSEL = 0b)

The first way gives the actual status of all TCR1-inputs if the internal interrupt is not active and actualizes the TCR1-register.

Is the interrupt active the content of TCR1-register is read and the interrupt is cleared.

The second way gives the content of TCR1-register and nothing will be changed.

The following figure shows the flow diagram of the interrupt logic.

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Figure 8 Flow Diagram of the Interrupt Logic

#### TCR2 and TCR3 Configuration Registers 2 and 3

TCR2 and TCR3 are the checksum of all the Coefficient bytes written into the Coefficient RAM (CRAM) of the SLICOFI by the COP-Command.

#### TCR2 7 5 4 3 1 Bit 6 2 0 LOW Byte of CRAM-checksum TCR3 5 4 3 2 1 Bit 7 6 0 OKCS HIGH Byte of CRAM-checksum **OKCS** shows, if the checksum is valid or the internal checksum calculation is

**OKCS** shows, if the checksum is valid or the internal checksum calculation is not yet finished <sup>1)</sup>

OKCS = 0 checksum is not valid

OKCS = 1 checksum is valid

Algorithm of defining the checksum:  $x^{16} x^{10} x^7 x 1$ 

With that algorithm you can reach a fault coverage of:  $(1 - 2^{-15})$ 

<sup>&</sup>lt;sup>1)</sup> After each change of the CRAM contents (COP-write or COP-read) the checksum has to be recalculated. During calculation time OKCS = 0.

#### TCR4 to TCR18: Configuration Register 4 to 18

These 15 bytes are the possible design information bytes which are described in **chapter 10.2** more detailed for the extended I0M-2 Channel Identification Command using an external ASIC.

#### TCR4 Bit 7 6 5 4 3 2 1 0 Byte 0 TCR5 3 2 Bit 7 6 5 4 1 0 Byte 1 **TCR18** Bit 7 6 5 4 3 2 1 0 Byte 14

TCR4 - TCR18 show the contents of the serial input of the ASIC via IDH-pin.

#### 5.5 COP Command

With a COP Command coefficients for the programmable filters can be written to the SLICOFI Coefficient RAM or read from the Coefficient RAM via the IOM-2 interface for verification. (Filter optimizing to different applications is supported by the software package SLICOS.)

Bit		7			6	5	4	3	2	1	0
					RW	CODE 4	CODE 3	CODE 2	0	CODE 1	CODE 0
RW		Read / WriteRW = 0Subsequent data is written to the SLICOFIRW = 1Read data from the SLICOFI									
COI	DE			incl	udes	number of f	ollowing b	ytes and fi	lter-addres	sses <sup>1)</sup>	
0	0	0	0	0	0	TH-Filter c	oefficients	(part 1)	(followed	by 8 bytes	s of data)
0	0	0	0	0	1	TH-Filter c	oefficients	(part 2)	(followed	by 8 bytes	s of data)
0	0	0	0	1	0	TH-Filter c	oefficients	(part 3)	(followed	by 8 bytes	s of data)
0	0	0	0	1	1	IM-Filter co	pefficients	(part 1)	(followed	by 8 bytes	s of data)
0	0	1	0	0	0	IM-Filter co	pefficients	(part 2)	(followed	by 8 bytes	s of data)
0	0	1	0	0	1	FRX-Fil	ter coeffic	ients	(followed	by 8 bytes	s of data)
0	0	1	0	1	0	FRR-Fil	ter coeffic	ients	(followed	by 8 bytes	s of data)
0	0	1	0	1	1	DC-Loop o	coefficient	(part 1)	(followed	by 8 bytes	s of data)
0	1	0	0	0	0	DC-Loop o	coefficient	(part 2)	(followed	by 8 bytes	s of data)
0	1	0	0	0	1	DC-Loop o	coefficient	(part 3)	(followed	by 8 bytes	s of data)
0	1	0	0	1	0	TTX and D	C-Loop co	oefficient	(followed	by 8 bytes	s of data)
0	1	0	0	1	1	AX-Filt	er coefficie	ents	(followed	by 8 bytes	s of data)
0	1	1	0	0	0	AR-Filt	er coefficio	ents	(followed	by 8 bytes	s of data)
0	1	1	0	0	1		er+BP1+L efficients	M-BP	(followed	by 8 bytes	s of data)
0	1	1	0	1	0	TG2-Filter	+BP2 coet	fficients	(followed	by 8 bytes	s of data)
0	1	1	0	1	1	Testing (leve	elmeter) co	oefficients	(followed	by 8 bytes	s of data)

<sup>&</sup>lt;sup>1)</sup> For generating a correct checksum all not used bits must be set to '0'.

#### 5.6 IOM<sup>®</sup>-2 Interface Command / Indication Byte

The Command/Indication (C/I) channel is used to communicate real time status information and for fast controlling of the SLICOFI. Data on the C/I channel is continuously transmitted in each frame until new data is to be sent.

#### Data Downstream C/I - Channel Byte (receive) - CIDD

Note that there is no address DD direction because there is only one SLICOFI per IOM2-channel. This byte is used for fast controlling of the SLICOFI. Each transfer to the SLICOFI has to last for at least 2 consecutive frames (FSC-cycles) so that it is accepted internally. Changes (spikes) of less than 2 FSC cycles are neglected.

Bit	7	6	5	4	3	2		
	RING	CONV	TIM	IO1	IO2	O1	]	
		see table belo see table belo	·	-	,			
Table			(		-,			
RING		CONV	Descri	ption				
0		0		Denial or Po (SCR1-7)	ower Down S	State (depe	nding on	
0		1	Active	State				
1		0	Power	Down or (au	itomatic) Po	wer Down F	Ring Pause	
1	1 (normal) Ringing State							
ТІМ		Timing bit to <b>chapter 6</b> ). TIM = 0 TIM = 1	SLICOFI is	s in the ringi	ng pause or	no meterpu		
IO1		Value for the as an output p the internally is switched to <b>chapter 6</b> , <b>p</b> IO1 = 0 IO1 = 1	oin. If the bit created Rin the IO1 p <b>age 51</b> ). The corres SLICOFI is The corres	REXTEN (S ig Burst On in instead o ponding pin s set to a log	CR5-2) is so Signal (for a of the IO1-bi at the digita gic 0.	et to 1 (exte an external t (for more al interface	rnal ringing) relay driver) details see of the	

# Programming the SLICOFI<sup>®</sup>

102	as an output	
	IO2 = 0	The corresponding pin at the digital interface of the SLICOFI is set to a logic 0.
	IO2 = 1	The corresponding pin at the digital interface of the SLICOFI is set to a logic 1.
01	Value for the	fixed Output Pin O1 (Pin 39).
	O1 = 0	The corresponding pin at the digital interface of the
	SLICOFI is se	et to a logic 0.
	O1 = 1	The corresponding pin at the digital interface of the SLICOFI is set to a logic 1.

### Data Upstream C/I - Channel Byte (transmit) - CIDU

Note that there is no address in DU direction too. This byte is used for fast transfer of the most important and time critical informations from the SLICOFI.

Bit	7	6	5	4	3	2	
	HOOK	GNK	SLCX	IO1	IO2	<b>I</b> 1	]
ноок		Indication of DUPGNK-cou HOOK = 0 HOOK = 1	unter in Pow Subscriber	•	tate).	e DUP-cou	inter or the
GNK	i	Indication if DUPGNK-cou s set to 0). GNK = 0 GNK = 1	unter). The f No ground		sabled in Po detected.	•	
SLCX	i t	Interrupt bit: S they are not Interrupt logic SLCX = 0 SLCX = 1	Summary of masked - f is describe No unmasl	utput of the filtered via	whole signa the DUP co a <b>chapter 5.</b> e signalling	ounter (see <b>4</b> , <b>page 36</b> ) register has	SCR7; the ). s toggled.
IO1	I	Logical state not programm IO1 = 0 IO1 = 1	ned as an in The corres SLICOFI is The corres		at the digita logic 0. at the digita	al interface	of the

102	0	e of the programmable Input/Output Pin IO2 (Pin 8) - even if nmed as an input pin. <sup>1)</sup>
	IO2 = 0	The corresponding pin at the digital interface of the
	IO2 = 1	SLICOFI is receiving a logic 0. The corresponding pin at the digital interface of the SLICOFI is receiving a logic 1.
11	Logical stat I1 = 0	e of the programmable Input Pin I1 (Pin 38). The corresponding pin at the digital interface of the SLICOFI is receiving a logic 0.
	1 = 1	The corresponding pin at the digital interface of the SLICOFI is receiving a logic 1.

The DUP- (DUPGNK) - counters filter the status-information and the input-pin I1 respectively. The counters count down and generate enable-signals for the registers if they are zero. Then they start counting again at the programmed value. If a status-information or an input-signal changes the proper counter is set and continues counting down. There are three different DUP-counters for HOOK, SLCX and the input-pin and one DUPGNK-counter for HOOK in PDen-mode or GNK in all other modes. Changing the mode freezes the actual status of HOOK and sets the actual HOOK-counter.

<sup>&</sup>lt;sup>1)</sup> If the Input/Output Pin is programmed as an output the corresponding bit in the CIDU is '1'

#### 6 Operating Modes

The SLICOFI supports 4 different Operating Modes: Power Denial (PDen), Power Down (PDown), Active and Ringing which are controlled via the upper 3 bits of the Data Downstream C/I channel byte (CIDD).

#### Table 9

RiING-(CIDD7)	CONV-(CIDD6)	TIM-(CIDD5)	Mode
0	0	x	PDen: PD (SCR1-7) = 0
0	0	x	PDown: PD (SCR1-7) = 1
1	0	0	PDown (Ring Pause)
0	1	0	Active
0	1	1	Active with Meterpulse on
1	x	1	Ringing: Ring Burst On
1	1	0	Ringing: Ring Pause

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PEB 3065 PEF 3065

#### **Operating Modes**



Figure 9

#### 6.1 Reset Behavior

The SLICOFI has 3 different reset sources that are all internally connected.

The Reset pin RES (pin 36), which works totally asynchronous to the external clocks.

The Reset bit (Within SOP - command, bit 4). The reset is valid for SOP-write only.

**Power On Reset.** If internal  $V_{\text{DDD}}$  gets above 1.5 Volts the SLICOFI is Reset by Power On Reset.

All 3 different sources set the SLICOFI to the basic setting modes (see below).

After a reset caused by any of the sources mentioned above, the reset bit (SCR0-4 = RSTST) in read direction is set to one. This bit is cleared (RSTST = 0) after it has been read by a SOP-read operation with the LSEL bits set to 00b (means: read only SCR0 byte). A SOP-read with other LSEL bits reads the actual RSTST value, but does not clear it.

The Reset pin RES has a Schmitt-Trigger input to reduce the sensitivity for spikes. In addition the pin RES has a spike rejection. All spikes smaller than typ. 70 ns are neglected. The pin RES can be set to 1 for an unlimited time but at least 125  $\mu$ s is recommended; during that, the DU pin is set to high impedance.

The SLICOFI leaves this mode automatically with the beginning of the next 8 kHz-frame (or after pin RES is released).

### 6.2 Basic Setting Modes

After RESET, the SLICOFI automatically is switched to its basic settings in which it uses internal default values for all filters and settings (AC and DC), so that the SLMA still works in a kind of "emergency mode" and can be handled by C/I-Interface commands only.

This means that for an (un-)determined reset (e.g. Power On Reset) the SLICOFI is reset, but can be switched to or return automatically to any operating mode presented to the C/I-channel after 2 FSC cycles. In all modes the SLMA stays stable, supervision and DC-feeding are still working and conversation can go on in a proper way until all filters and settings have been reloaded by SOP and COP-commands.

So what happens internally after reset?

- all configuration registers are set to their default values (note that the Coefficient RAM is **not** reset)
- the RSTST-bit (SCR0-4) is set to 1 to indicate that a reset has taken place
- The IOM-2 interface is reset. Running communication is stopped
- DU is in high impedance state
- AC- and DC-loop use the default values and not the programmed ones (see below)

Parameter	Values	Unit	Test Condition/Result
Const I	26	mA	limit for Constant Current (for Active and Power Down)
RFS	2×150		Feeding Resistance (for Active and Power Down - excluding the external Fuse resistors)
$V_{drop}$	10	V	Overall voltage drop (to reach maximum length and there is no Teletax)
$\overline{f_{Ring}}$	25	Hz	Ring Frequency
ARing	1.7	V	Ring rms-value at $V_{2W}$
PDen	1.45	Vrms	Power Denial Voltage for Offhook
Offhook	8	mA	Offhook Detection (for Power Down, Ringing and Active without hysteresis)
DC-Lowpass	0.3/5	Hz	DC- Lowpass set to 0.3 and 5 Hz respectively
Levelmeter			undefined (parameters stored in CRAM)
DUP	10	ms	Data Upstream Persistency Counter is set to 10 ms
DUPGNK	20	ms	Data Upstream Persistency Counter for GNK is set to 20 ms

#### Table 10 DC

Boosted Battery is reset to normal feeding

Reverse Polarity is reset to Normal Polarity

all bits of the Signalling Register are masked and reset to 0

the Data Upstream C/I channel byte is reset to 0 (and IO's are set to Input pins)

C1 and C2 are set to PDNR and PDN is set high

A-Law is chosen

# Table 11 AC

Parameter	Values	Unit	est Condition/Result				
IM-Filter	900		Approximately 900 Real Input Impedance				
TH-Filter	TH <sub>BRD</sub>	Approximately BRD-Impedance for Balanced Network					
AX	10	dB	Attenuation Transmit (this means about 0 dB for SLMA)				
AR	- 15.11	dB	Attenuation Receive (this means about – 7 dB for SLMA)				
ATTX	190	mV	Teletax Generator Amplitude at $V_{2W}$ ; but note that the SLICOFI is set to TTXNO = 1 with reset				

Parameter	Values	Unit	Test Condition/Result
$\overline{f_{\text{TTX}}}$ SOREV	16	kHz	Teletax Generator frequency; but note that the SLICOFI is set to TTXNO = 1 with reset for Metering with Polarity Reversal: Hard Reversal is used.
TG1	1008	Hz	Tone Generator 1 and AC-levelmeter Bandpass
TG2	2000	Hz	Tone Generator 2 (+ 2 dB compared to TG1)

Table 11AC (cont'd)

# 6.3 Power Denial (PDen)

After a Reset (including the Power On Reset) the SLICOFI is set to Power Denial State. In Power Denial all functions that are not necessary are disabled to minimize power consumption. Via the two pins  $V_{\text{LINE1}}$  and  $V_{\text{LINE2}}$  the SLICOFI is directly connected to the a - and b - wire, while the PDN-Pin is set high (which turns off the HV-SLIC). While the interface is fully working - including programmability of the registers with SOP- or TOP commands and the Coefficient RAM (COP commands) the rest of the SLICOFI is turned off except the supervision of the line. The change of the line state is reported via the HOOK-bit in the IOM-2 Data upstream channel. To avoid spurious Offhook - informations caused by longitudinal induction the HOOK - bit is low pass filtered (programmable with the DUPGNK - counter in PDen state only). The HV-interface pins C1, C2 are switched off. The voice channel Data Downstream is directly fed into the voice channel Data Upstream. The HOOK-indication in PDen is optimized for longitudinal suppression up to 65 Vrms for the Offhook transition.

### 6.4 Power Down (PDown)

In Power Down Mode the DC-Loop of the SLICOFI is fully working; the AC-Loop is still turned off. The output voltage at the V2W pin is controlled via the IT input in such a way that it behaves like a programmable constant current source. Current limitation is used for detecting Offhook, too. The change of the line state is reported via the HOOK-bit in the IOM-2 Data upstream channel. To avoid spurious Offhook-informations the HOOK-bit is lowpass filtered (programmable with DUP-counter).

The ternary HV-interface (C1, C2) is set to Power Down mode. If Offhook is detected the HV-interface is set to one of the active modes. This can be avoided by setting PDADIS = 1 (SCR3-4). Then the HV-SLIC interface is set to Power Down anyway.

The longitudinal current supervision via the IL pin is activated in this mode.

The voice channel Data Downstream is directly fed into the voice channel Data Upstream.

Together with the bits Hi-a and Hi-b of the configuration register 1 (SCR1-2 and SCR1-3) simple handling of Ground Start function is possible.

# SIEMENS

#### **Operating Modes**

#### Table 12

		Pin No./Pin Name								
	CIDD7	CIDD6	CIDD5	SCR1-7	SCR1-3	SCRI1-2	PIN 28	PIN 9	PIN 10	
	RING	CONV	тім	PD	HI-b	HI-a	PDN	C1	C2	
PDNH - Loop open (lab < 30 μA)	0	0	1	x	x	x	1	V <sub>OL</sub>	V <sub>OL</sub>	
PDNR	0	0	0	0	not	(11)	1	V <sub>OZ</sub>	V <sub>OZ</sub>	
PDown	0	0	0	1	0	0	0	$V_{OH}$	V <sub>OH</sub>	
PDown (with Hi-a)	0	0	0	1	0	1	0	$V_{OL}$	V <sub>OH</sub>	
PDown (with Hi-b)	0	0	0	1	1	0	0	V <sub>OZ</sub>	V <sub>OH</sub>	
b-line high impedance (Ground Start)	0	0	0	x	1	1	0	V <sub>OZ</sub>	V <sub>OH</sub>	

#### 6.5 Active Mode (Act)

In Active Mode ("Conversation State") both AC-and DC-Loop are fully working. The output voltage at the  $V_{2W}$  pin is controlled via the IT input pin in such way, that it behaves like a constant current source which turns automatically into a programmable resistive feeding source due to the DC-Characteristic values (see **chapter 3.2**, **page 13** for more details).

The ternary HV-interface is set to one of the active modes.

#### Polarity

The SLICOFI supports either normal or reverse Polarity which is set by the POLNR-bit (SOP-5). The information is transferred to the HV-Interface and simultaneously a 180 degree phase shift of the AC- and DC-Loop is done. The performance and the functionality is not influenced by that.

#### **Boosted Battery**

To feed subscriber lines with enhanced loop resistance the SLICOFI supports the Boosted Battery mode. The HV-Interface pins are set to Boosted Battery (BB) mode and the maximum  $V_{2W}$  output voltage is extended to – 3.2 V.

#### Meterpulses

The SLICOFI supports two different kinds of meterpulses: Meterpulses with 12/16 kHz (Teletax Metering) and with polarity reversal. In the Active Mode the Timing bit (TIM) controls the meterpulse which might be 12/16 kHz **or** reversal. The decision between

these two ways is made by the bit TTXNO (SCR3-7). If bit TTXNO is set to 1, then the meterpulse is reversal. In this case the Timing bit is linked to POLNR (SOP-5) by an EXOR gate. If bit TTXNO is set to 0, then the Timing bit and POLNR are completely independent from another and Teletax Metering is used.

#### Metering with Polarity Reversal

#### Hard or Soft (SOREV, SCR 3-5)

As long as the TIM bit of the C/I-channel is set to 1, the SLICOFI is changing the actual polarity of the HV-Interface and performs an immediate 180 degree phase shift of the AC- and DC-Loop.

### **Teletax Metering Injection**

For countries with Teletax Metering, the SLICOFI provides either a 12 or 16 kHz Signal (switchable with the bit TTX12 (SCR3-6))<sup>1)</sup> which amplitude is free programmable up to 250 mVrms at  $V_{2W}$ . The SLICOFI filters the Teletax pulses in transmit direction, too. The slope of the pulses are internally shaped, so that the noise during switching and transmission is less than 50 mV at  $V_{2W}$  and 1 mV at the IOM-2 interface (psophometrically weighted). With the bit NOSL (SCR2-2) the slope can be switched off. In that case the switching noise is not defined (for signalling only).

### 6.6 Ringing Mode

The SLICOFI generally supports balanced ringing.

If the SLICOFI is set to Ringing Mode, the HV-Interface is set to Ringing Mode, the AC-loop is turned off and the DC-Loop is automatically opened.

The voice channel Data Downstream is directly fed into the voice channel Data Upstream.

### Balanced Ringing

The sine wave of the ringing is generated in the SLICOFI. The frequency and the amplitude are free programmable between 16 and 70 Hz and up to 2.125 Vrms at  $V_{2W}$ , respectively<sup>2)</sup>. In Ring Pause 0 V is provided at  $V_{2W}$ . If the Ring Burst On (RBO) command is sent to the SLICOFI via the C/I-channel (RING and TIM = 1) the begin and end (TIM = 0) of the ring burst is automatically synchronized at the voltage zero crossing. If the DC-current at the IT-pin exceeds the programmed value, Offhook is detected within 2 periods of the ringing frequency and the Ring Burst at  $V_{2W}$  is switched off within 3 periods. During Offhook the Ring Burst On command is neglected.

<sup>&</sup>lt;sup>1)</sup> Note, that the right Teletax Coefficient Set (via COP-command) must be provided, too.

<sup>&</sup>lt;sup>2)</sup> Note that the DC-value is 0. So DC injection has to be performed by the HV-SLIC.

# **Unbalanced (external) Ringing**

The sine wave for ringing is generated by an external ring generator. To coordinate with the SLICOFI following settings must be done:

- 1. IO1 set as an output
- 2. SCR5-2 (REXTEN) = 1
- 3. RING-(CIDD7) = 1 (PDown: Ring Pause)
- 4. TIM-(CIDD5) = 1 (Ringing: Ring Burst On)

Pin REXT: a positive puls according to zerocrossing of the ringer voltage

RINGING:

- 5. signal for relays on IO1
- 6. HV-SLIC in PDen Mode
- 7. SLICOFI in PDown Mode, Offhook-detection via  $V_{\text{LINE1,2}}$

RING PAUSE:

8. TIM-(CIDD5) = 0 (PDown: Ring Pause), Offhook-detection via IT (in the same way as balanced ringing)

#### **SLIC Interface**

#### 7 SLIC Interface

#### 2 Wire Output Voltage ( $V_{2W}$ )

The  $V_{2W}$  output voltage pin (26) represents the sum for AC- and DC-loop together with Teletax info or Ring Burst at the receive path. The buffer is designed for a load of  $R_{\rm L} > 600$  and  $C_{\rm L} < 10$  pF and directly connected to the HV-SLIC in typical applications.

#### Transversal Current Sense AC - Input (ITAC)

The pin ITAC (21) is the input voltage pin for the AC transversal current information from the HV-SLIC in the transmit path. AC/DC separation is done by an external highpass filter (capacitor range between 680 nF - 1  $\mu$ F). The input resistance is larger than 20 k. Current/voltage conversion is done via an external resistor (same for pin IT).

#### Transversal Current Sense DC - Input (IT)

The pin IT (19) is the input voltage pin for the DC transversal current information from the HV-SLIC in the transmit path. The signal is internally filtered via a 0.3 Hz lowpass. The input resistance is larger than 20 k. Current/voltage conversion is done via an external resistor (same for pin ITAC).

#### Longitudinal Current Sense - Input (IL)

The scaled longitudinal current information transferred from the HV-SLIC – the current-voltage conversion is done by an external resistor – is lowpass filtered (time programmable using DUPGNK-counter) and is reported via the Data Upstream C/I-channel (CIDU-6). In Power Denial, the GNK-bit is set to '0' and the setting of the Interrupt bit (CIDU-5) caused by GNK is prohibited. Changing from PDen to any other mode, the DUPGNK-counter is set to the programmed value; so the change of the GNK information (CIDU-6) is lowpass filtered anyway.

### Battery Image Input ( $V_{\text{BIM}}$ )

The information about the actually used battery voltage ( $V_{BAT}$ ) of the SLMA is transferred from the HV-SLIC via the  $V_{BIM}$  pin to the SLICOFI. In order to give some information about the operating point of the SLMA there is a comparison of the actual battery voltage and the output voltage  $V_{2W}$  of the SLICOFI. This information is transferred via the Signalling register (TCR1-5: VB/2).

If  $|V_{2W}| < |V_{BIM}/2|$  the VB/2-bit is set to 1, else to 0.

### Ternary Interface (C1, C2) and HV-SLIC Switch Off Output (PDN)

In order to set the HV-SLIC to the different operating states, the information of the SLMA-controller is passed through from the IOM-2-channel to the ternary HV-SLIC-Interface pins C1 and C2.

#### **SLIC Interface**

#### Table 13

		C2 (PIN 10)						
		VOL	VOZ	VOH				
	VOL	RING RP/PDNH	RING NP	HI-a				
C1 (PIN 9)	VOZ	BB RP	BB NP/PDNR	HI-b				
	VOH	Active RP	Active NP	PDown				

- BB Boosted battery
- RP Reverse Polarity
- NP Normal Polarity
- HI-b High Impedance b-leg
- HI-a High Impedance a-leg
- PDNH Power Denial High Impedance
- PDNR Power Denial Resistive

For signalling "Over temperature" the HV-SLIC drains a current ( $I_{OT}$ ) from pin 9. The message is transferred via the Signalling register (TCR1-3). This is possible in any operating states of the HV-Interface except for Power Denial.

The HV-SLIC (PEB 4065) has two different Power Denial Modes:

- 1. PDNR, the resistive mode which provides a connection of 15 k $\Omega$  from TIP and RING to BGND and  $V_{\text{BAT}}$ , respectively
- 2. PDNH, offers high impedance at TIP and RING

In this mode (PDN = 1) the HV-SLIC is completely turned off. Line supervision is done via the  $V_{\text{LINE1,2}}$  pins. In all other modes, PDN is set to GND ( $R_{\text{ON}} < 250 \Omega$ ).

### Line Sense Pins ( $V_{\text{LINE1,2}}$ )

In Power Denial state the line supervision is done via the  $V_{\text{LINE1,2}}$  pins. If the voltage  $V_{\text{LINE}}$  between the two pins exceeds the programmed value, Offhook is reported via the Data Upstream C/I-channel (CIDU-7)<sup>1)</sup>. To reach the longitudinal voltage suppression, the incoming signal is low pass filtered using the values that are programmed by the DUPGNK counter (no longitudinal current information present in PDen, but the same interferences).

<sup>&</sup>lt;sup>1)</sup> Note:  $V_{\text{LINE}} = V_{\text{LINE1}} - V_{\text{LINE2}}$ ; so the voltage of  $V_{\text{LINE1}}$  has to be higher than  $V_{\text{LINE2}}$  for correct external indication

#### 8 Transmission Characteristics

The target figures in this specification are based on the subscriber-line board requirements. The proper adjustment of the programmable filters (transhybrid balancing, impedance matching, frequency-response correction) needs a complete knowledge of the SLICOFI's analog environment. Unless otherwise stated, the transmission characteristics are guaranteed within the test conditions.

### **Test Conditions**

 $T_{A} = 0 \text{ °C to } 70 \text{ °C};$   $V_{DDD} = V_{DDA} = 5 \text{ V} \pm 5\%; V_{SS} = -5 \text{ V} \pm 5\%; \text{ GNDA} = \text{GNDD} = 0 \text{ V}$   $R_{L} > 600 \Omega; C_{L} < 10 \text{ pF} (at V_{2W}); H_{IM} = H_{TH} = 0; H_{FRX} = H_{FRR} = 1$  AR = 0 dB AX = 0 dB $f = 1004 \text{ Hz}; 0 \text{ dBm0}; \text{ A-Law or }\mu\text{-Law};$ 

In Transmit direction for  $\mu$ -law an additional gain of 1.94 dB is implemented.

The 0 dBm0 definitions for Receive and Transmit are different.

A 0 dBm0 signal in Transmit direction is equivalent to 206 mVrms [165 mVrms].

(A -Law, [μ-Law]).

A 0 dBm0 signal in Receive direction is equivalent to 118 mVrms.



### Figure 10

With  $V_{IT} = 0 \text{ dBm0}|_{SLICOFI} = -11.51 \text{ dBm0}|_{600} = 206 \text{ mV}$  for transmit With  $V_{V2W} = 0 \text{ dBm0}|_{SLICOFI} = -16.34 \text{ dBm0}|_{600} = 118 \text{ mV}$  for receive

# Table 14

Parameter	rameter Symbol Limi		nit Valu	es	Unit	Test Condition
		min.	typ.	max.	-	
Gain absolute						
transmit	$G_{X}$	- 0.20	± 0.05	0.20	dB	
receive	$G_{R}$	- 0.20	± 0.05	0.20	dB	
IMAN-Loop	$G_{IMAN}$	- 0.5	± 0.1	0.5	dB	adding to – 7.2 dB Loop
TTX-injection	$G_{TTX}$	- 0.7	± 0.3	0.7	dB	gain
Total Harmonic distortion						
transmit	THD <sub>T</sub>		- 56	- 48	dB	at 0 dBm0;
receive	THD <sub>R</sub>		- 56	- 48	dB	$f = 1 \text{ kHz}; 2^{\text{nd}}, 3^{\text{rd}} \text{ order}$ at 0 dBm0; $f = 1 \text{ kHz}; 2^{\text{nd}}, 3^{\text{rd}} \text{ order}$
Ringing injection	THD <sub>Rng</sub>		- 35	- 34	dB	f = 16.3-70  Hz
TTX injection			- 60	- 40	dB	f = 12 kHz and 16 kHz
Idle channel						
noise transmit	$N_{TP}$		- 69	- 67	dBm0p	Teletax countries, burst off A-law, psophometric:
	$N_{TTX\_TP}$		- 65	- 60	dBm0p	$V_{IN} = 0 V$ Teletax burst on A-law, psophometric: $V_{IN} = 0 V$
	$N_{\rm G_{TP}}$		- 58	- 53	dBm0p	AX = 30 dB Teletax countries, burst off A-law, psophometric:
receive	$N_{RP}$		- 88	- 81	dBm0p	$V_{\rm IN} = 0$ V Teletax countries, burst off A-law, psophometric idle code +0
	$N_{TTX\_RP}$		- 87	- 80	dBm0p	Teletax burst on A-law, psophometric idle code +0

### 8.1 Frequency Response

**Receive:** reference frequency 1 kHz, signal level 0 dBm0, H<sub>FRR</sub> = 1



# Figure 11

**Transmit:** reference frequency 1 kHz, signal level 0 dBm0,  $H_{FRX} = 1$ 



#### Figure 12

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#### 8.2 Group Delay

Maximum delays when the SLICOFI is operating with  $H_{TH} = H_{IM} = 0$  and  $H_{FRR} = H_{FRX} = 1$  including delay through A/D- and D/A converters. Specific filter programming may cause additional group delays.

Group Delay deviations stay within the limits in the figures below.

Group Delay absolute values: Signal level 0 dBm0

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.	-	
Transmit delay	D <sub>XA</sub>	250	312	375	μs	f <sub>Test</sub> @ T <sub>Gmin</sub>
Receive delay	$D_{RA}$	250	312	375	μs	f <sub>Test</sub> @ T <sub>Gmin</sub>
Digital loop back	D <sub>RX</sub>			630	μs	f <sub>Test</sub> @ T <sub>Gmin</sub>

#### Table 15

# Group Delay Distortion receive and transmit: Signal level 0 dBm0, f<sub>Test</sub> @ T<sub>Gmin</sub>



# 8.3 Out-of-Band Signals at Analog Output (receive)

With a 0 dBm0 sine wave with frequency f(300 Hz to 3.4 kHz) applied to the digital input, the level of any resulting out-of-band signal at the analog output will stay at least X dB below a 0 dBm0, 1 kHz sine wave reference signal at the analog output



### Figure 13

3.4...4.6 kHz: X = 
$$-14\left(\sin\left(\left(\pi\frac{4000-f}{1200}\right)-1\right)\right)$$

# 8.4 Out-of-Band Signals at Analog Input (transmit)

With a 0 dBm0 out-of-band sine wave signal with frequency f (< 100 Hz or 3.4 kHz to 100 kHz) applied to the analog input, the level of any resulting frequency component at the digital output will stay at least X dB below a 0 dBm0, 1 kHz sine wave reference signal at the analog input.<sup>1)</sup>



### Figure 14

3.4...4.0 kHz: X= 
$$-14\left(\sin\left(\pi\frac{4000-f}{1200}\right)-1\right)$$

4.0...4.6 kHz: X = 
$$-18\left(\sin\left(\pi\frac{4000-f}{1200}\right) - \frac{7}{9}\right)$$

<sup>&</sup>lt;sup>1)</sup> Poles at 12 kHz  $\pm$  150 Hz respectively 16 kHz  $\pm$  150 Hz and harmonics will be provided.

# **SIEMENS**

#### **Transmission Characteristics**

# 8.5 Overload Compression

**Transmit:** measured with sine wave f = 1004 Hz.



Figure 15

#### 8.6 Gain Tracking (receive or transmit)

The gain deviations stay within the limits in the figures below.

**Receive:** measured with sine wave f = 1004 Hz reference level is -10 dBm0.  $A_{R} = 6$  dB



#### Figure 16

Transmit: measured with sine wave f = 1004 Hz reference level is -10 dBm0.  $A_x = 0$  dB



#### Figure 17

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### 8.7 Total Distortion

The signal to distortion ratio exceeds the limits in the following figure:

**Receive:** measured with sine wave f = 1004 Hz (C-message weighted for  $\mu$ -law, psophometrically weighted for A-law).



#### Figure 18

$$\Sigma(A_{R1} + A_{R2}) = 7 \text{ dB}$$

#### Table 16

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
Signal to Distortion at full attenuation	$SD_{\text{att}_{R}}$		– 13	-7		Signal S = $-40 \text{ dB}$ A <sub>R</sub> = 30 dB

**Transmit:** measured with sine wave f = 1004 Hz (C-message weighted for  $\mu$ -law, psophometrically weighted for A-law).

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# **Transmission Characteristics**



# Figure 19

 $A_X = -7 \text{ dB}$ 

# Table 17

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
Signal to Distortion at full gain	SD <sub>att_T</sub>		- 17	- 12	dB	Signal S = $-40 \text{ dB}$ A <sub>X</sub> = $-30 \text{ dB}$
Signal to Distortion in IMAN Loop	SD <sub>IMAN</sub>		- 39	- 30	dB	Signal S = – 45 dB

#### 8.8 Transhybrid Loss

The quality of Transhybrid-Balancing is very sensitive to deviations in gain and group delay – deviations inherent to the SLICOFI A/D- and D/A-converters as well as to all external components used on a line card (HV-SLIC).

Measurement of SLICOFI Transhybrid-Loss: A 0 dBm0 sine wave signal with a frequency in the range between 300-3400 Hz is applied to the digital input. The resulting analog output signal at pin  $V_{2W}$  is connected to the pin ITAC via a 1 le filters FRR, A<sub>R</sub>, FRX, A<sub>X</sub> and IM are disabled, the balancing filter TH is enabled with coefficients optimized for this configuration ( $V_{2W}$  = ITAC).

The resulting echo measured at the digital output is at least X dB below the level of the digital input signal as shown in the table below

	COP-write				Coef	ficient	S		
TH-Filter Part 1	00 <sub>H</sub>	00	80	80	18	00	08	00	88
TH-Filter Part 2	01 <sub>H</sub>	08	00	AF	84	04	AC	2B	90
TH-Filter Part 3	02 <sub>H</sub>	DA	AB	B3	22	DB	37	88	00

#### Table 18

#### Table 19

Parameter	Symbol	Limit Values		Unit	Test Condition
		min.	typ.	Ī	
Transhybrid Loss at 500 Hz	THL <sub>500</sub>	33	50	dB	
Transhybrid Loss at 2500 Hz	THL <sub>2500</sub>	29	44	dB	
Transhybrid Loss at 3000 Hz	THL <sub>3000</sub>	27	42	dB	

#### 9 Electrical Characteristics

#### 9.1 Absolute Maximum Ratings

#### Table 20

Parameter	Symbol	Limit Values		Unit	Test Condition
		min.	max.		
$V_{\text{DDA}}$ referred to GNDA $V_{\text{DDD}}$ referred to GNDD $V_{\text{SS}}$ referred to GNDA GNDA with respect to GNDD $V_{\text{DDA}}$ with respect to $V_{\text{DDD}}$ $V_{\text{LINE1,2}}$ referred to GND		- 0.3 - 0.3 - 5.5 - 0.3 - 0.3 - 75	7.0 7.0 0.3 0.3 0.3 75	V V V V V	
Analog input and output voltage referred to $V_{\text{DDA}} = 5 \text{ V}; (V_{\text{SS}} = -5 \text{ V})$ referred to $V_{\text{SS}} = -5 \text{ V}; (V_{\text{DDA}} = 5 \text{ V})$		- 10.3 - 0.3	0.3 10.3	v v	
All digital input voltages referred to GNDD = 0 V; $(V_{DDD} = 5 V)$ referred to $V_{DDD} = 5 V$ ; (GNDD = 0 V)		- 0.3 - 5.3	5.3 0.3	v v	
DC input and output current at any input or output pin (free from latch -up)			100	mA	
Storage temperature Ambient temperature under bias	$T_{\rm STG}$ $T_{\rm A}$	- 65 - 10	125 80	°C ℃	
Power dissipation	P <sub>D</sub>		1	W	
ESD-integrity (according MIL-Std 883D, method 3015.7) <sup>1)</sup>	V <sub>ESD</sub>	1000		V	

<sup>1)</sup> All Pins except  $V_{\text{LINE1}}$  and  $V_{\text{LINE2}}$  (11, 12); for these Pins  $V_{\text{ESD}}$  < 500 V due to process limitation

Note: Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device.

Functional operation under these conditions is not implied.

Exposure to conditions beyond those indicated in the recommended operational conditions of this specification may effect device reliability.

# 9.1.1 Operating Range

 $T_{\rm A}$  = - 40 to 85 °C;  $V_{\rm DD}$  =  $V_{\rm DDD}$  =  $V_{\rm DDA}$  = 5 V  $\pm$  5%;  $V_{\rm SS}$  = - 5 V  $\pm$  5%; GNDD = GNDA = 0 V

### Table 21

Parameter	Symbol	Li	mit Val	ues	Unit	Test Condition
		min.	typ.	max.		
V <sub>DD</sub> supply current <sup>1)</sup> Power Denial Power Down Active Active with TTX Ringing	IDD <sub>PDen</sub> IDD <sub>PDown</sub> IDD <sub>Act</sub> IDD <sub>TTX</sub> IDD <sub>Rng</sub>		4 11 21 25 11	6 15 30 34 15	mA mA mA mA mA	
V <sub>SS</sub> supply current <sup>1)</sup> Power Denial Power Down Active Active with TTX Ringing	ISS <sub>PDen</sub> ISS <sub>PDown</sub> ISS <sub>Act</sub> ISS <sub>TTX</sub> ISS <sub>Rng</sub>		1 3,5 4,5 7 3.5	2 6 7 10 6	mA mA mA mA mA	
Power supply rejection-ratio receive $V_{\text{DD}}$ receive $V_{\text{SS}}$ transmit $V_{\text{DD}}$ transmit $V_{\text{SS}}$	PSRR	56 56 40 40	70 65 70 50		dB dB dB dB	ripple: 1 kHz, 70 mVrms at $V_{2W}$ at $V_{2W}$ at IOM-2 at IOM-2
Power dissipation <sup>1)</sup> Power Denial Power Down Active Active with TTX Ringing	PDen PDown Act TTX Rng		25 73 128 160 73	42 110 195 231 110	mW mW mW mW mW	

<sup>1)</sup> Power dissipation and supply currents are target values.

# 9.2 Digital Interface

 $T_{\rm A}$  = - 40 to 85 °C;  $V_{\rm DD}$  =  $V_{\rm DDD}$  =  $V_{\rm DDA}$  = 5 V ± 5%;  $V_{\rm SS}$  = - 5 V ± 5%; GNDD = GNDA = 0 V

### Table 22

Parameter	Symbol	Limit	Values	Unit	Test condition	
		min	max.			
For all input pins (including IO-Pins):						
Low-input posgoing	$V_{T+}$	- 0,3	3.15	V	see figure below	
Low-input neggoing	<i>V</i> <sub>T-</sub>	1.35	V <sub>DD</sub> + 0,3	V	see figure below	
Low-input Hysteresis	$V_{H}$	0.5		V	$V_{\rm H} = V_{\rm T+} - V_{\rm T-}$	
Input leakage current	I <sub>IL</sub>	– 1	1	μΑ	$-0.3 \le V_{\rm in} \le V_{\rm DD}$	
Spike rejection for RESET (pin 36)	t <sub>rej</sub>	50	200	ns		
Ternary Inputs: ID-L, ID-M (pins 31, 32)						
High level	$V_{IHID}$	2.0		V		
Zero level	$V_{IMID}$	- 0.8	0.8	V		
Low level	$V_{ILID}$		- 2.0	V		
For all output pins except DU (Pin 6; including IO-Pins):						
Low-output voltage	$V_{OL}$		0.45	V	$I_{\rm O} = -2 \text{ mA}$ (typ. at $I_{\rm O} = -3.5 \text{ mA}$ )	
High-output voltage	V <sub>OH</sub>	3.5		V	$I_{\rm O} = 400 \mu{\rm A}$	
for DU-pin (Pin 6)						
Low-output voltage	$V_{OLDU}$		0.45	V	$I_{O} = -4 \text{ mA}$ (typ. at $I_{O} = -7 \text{ mA}$ )	
High-output voltage	$V_{OHDU}$	3.5		V	$I_{\rm O} = 400 \mu{\rm A}$	

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# **Electrical Characteristics**



# Figure 20

### 9.3 DC-Feeding

# 9.3.1 DC-Feeding ( $T_A = 0$ to 70 °C)

 $T_{\rm A}$  = - 0 to 70 °C;  $V_{\rm DD}$  =  $V_{\rm DDD}$  =  $V_{\rm DDA}$  = 5 V  $\pm$  5%;  $V_{\rm SS}$  = - 5 V  $\pm$  5%; GNDD = GNDA = 0 V

#### Table 23

Parameter	Symbol	Limit Values U		Unit	Test condition	
		min.	typ.	max.		
"Line Current" Measurement:						
Transmit	$V_{ m IT~offset}$ $V_{ m IT~gain}$ $V_{ m IT~gain}$ $V_{ m IT~THD}$ -	- 25 0.94 - 1.06 40	50	25 1.06 - 0.94	mV dB	direct/reverse polarity f < 50 Hz, direct polarity f < 50 Hz, reverse polarity direct/reverse polarity
"Line Voltage" Feeding:						
Receive	$\begin{array}{c} V_{2\rm W~offset} \\ V_{2\rm W~gain} \\ V_{2\rm W~THD} \end{array}$	- 25 0.94 40	50	25 1.06	mV dB	normal battery, $f = 300$ Hz normal battery, $f = 300$ Hz normal battery
Receive Boosted		- 40 1.5 40	1.6 50	40 1.7	mV dB	boosted battery, $f = 300$ Hz boosted battery, $f = 300$ Hz boosted battery

### 9.3.2 DC-Feeding ( $T_A = -40$ to 85 °C)

 $T_{A} = -40$  to 85 °C;  $V_{DD} = V_{DDD} = V_{DDA} = 5 V \pm 5\%$ ;  $V_{SS} = -5 V \pm 5\%$ ; GNDD = GNDA = 0 V

#### Table 24

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
"Line Current" Measurement:						
Transmit	$V_{ m IT~offset}$ $V_{ m IT~gain}$ $V_{ m IT~gain}$ $V_{ m IT~THD}$ -	- 30 0.94 - 1.06 40	50	30 1.06 - 0.94	mV dB	direct/reverse polarity f < 50 Hz, direct polarity f < 50 Hz, reverse polarity direct/reverse polarity

Parameter	Symbol	Limit Values		Unit	Test Condition	
		min.	typ.	max.		
"Line Voltage"						
Feeding:						
Receive	$V_{\rm 2W \; offset}$	-30		30	mV	normal battery, $f = 300 \text{ Hz}$
	$V_{\rm 2W\;gain}$	0.927		1.073		normal battery, $f = 300 \text{ Hz}$
	$V_{2W \text{ THD}}$	40	50		dB	normal battery
<b>Receive Boosted</b>	V <sub>2W offset</sub>	-48		48	mV	boosted battery, $f = 300 \text{ Hz}$
	$V_{2W \text{ gain}}$	1.48	1.6	1.72		boosted battery, $f = 300 \text{ Hz}$
	$V_{2W \text{ THD}}$	40	50		dB	boosted battery

Table 24(cont'd)

# 9.4 HV-SLIC Interface

 $T_{A} = -40$  to 85 °C;  $V_{DD} = V_{DDD} = V_{DDA} = 5 \text{ V} \pm 5\%$ ;  $V_{SS} = -5 \text{ V} \pm 5\%$ ; GNDD = GNDA = 0 V

#### Table 25

Parameter	Symbol	Limit Values			Unit	Test Condition/Result
		min.	typ.	max.		
Ground Key Detection						
at Pin IL	$V_{ILLo}$	- 217		217	mV	GNK = 0
	$V_{_{\rm ILHi}}$	293		- 293	mV	GNK = 1
Half Battery Information						$V_{\text{BIM}} = -3 \text{ V}$
at Pin $V_{2W}$	$V_{\rm V2WLo}$			- 1.65	V	VB/2 = 0
200	$V_{V2WHi}$	- 1.35			V	VB/2 = 1
PDN-Pin						
max. R <sub>on</sub>	R <sub>on</sub>		90	250		in Active-Mode to GND

# Table 25(cont'd)

Parameter	Symbol	Limit Values			Unit	Test Condition/Result
		min.	typ.	max.		
Output voltage: HV-SLIC-Interface Pins 9, 10 (C1, C2) High level Zero level Low level Current drained from pin 9 (C1) in all 3 states	$V_{ m OHHV}$ $V_{ m OMHV}$ $V_{ m OLHV}$ $I_{ m OTLo}$ $I_{ m OTHi}$	2.5 - 0.8 480		0.8 - 2.5 320	V V μΑ μΑ	$I_{out} < 10$ $I_{out} < 10$ $I_{out} < 10$ TEMPA = 0 TEMPA = 1
External Indication	$V_{EXT\_off}$ $V_{EXT\_0}$ $V_{EXT\_6}$	- 200 0.5 0.3		200 1.3 1.5	mV V V	measured at IOM-2 without DC $V_{\text{LINE}} = 0 \text{ V}$ without DC $V_{\text{LINE}} = 6 \text{ V}$ with DC = 30 V $V_{\text{LINE}}^{1)} = 6 \text{ V}$

<sup>1)</sup>  $V_{\text{LINE}} = V_{\text{LINE1}} - V_{\text{LINE2}}$ 

# 9.5 IOM<sup>®</sup>-2 Interface Timing



# Figure 21
#### **Electrical Characteristics**

## Table 26 Switching Characteristics

Parameter	Symbol		Limit Values		
		min.	typ.	max.	
Period DCL "slow" mode <sup>1)</sup> Period DCL "fast" mode <sup>2)</sup> DCL Duty Cycle Period FSC FSC set-up time FSC hold time	$ \begin{array}{c} t_{\rm DCL} \\ t_{\rm DCL} \\ t_{\rm DCLh} \\ t_{\rm FSC} \\ t_{\rm FSC\_S} \\ t_{\rm FSC\_H} \end{array} $	40 70 40	1/2048 1/4096 125 t <sub>DCLh</sub>	60	kHz kHz % μs ns ns
DD data in set-up time DD data in hold time DU data out delay (intrinsic) DU data out delay	$t_{\text{DD}\_S}$ $t_{\text{DD}\_H}$ $t_{\text{dDUintr.}}$ $t_{\text{dDU}}$	20 50	40 150 <sup>3)</sup>	70 250	ns ns ns ns

<sup>1)</sup> DCL = 2048 kHz:  $t_{FSC} = 256 \times t_{DCL}$ 

<sup>2)</sup> DCL = 4096 kHz:  $t_{FSC} = 512 \times t_{DCL}$ 

<sup>3)</sup> Depending on Pull up resistor (typical 1...10 k)

#### **Electrical Characteristics**



# 9.6 IOM<sup>®</sup>-2 Command/Indication Interface Timing (DCL = 4096 kHz)

## Figure 22

Table 27	Switching	Characteristics
----------	-----------	-----------------

Parameter	Symbol	Limit Values			Unit
		min.	typ.	max.	
Command out delay	t <sub>dCout</sub>			0	ns
Command out high impedance Command out active	t <sub>dCZ</sub> t <sub>dCA</sub>		150 150	200 200	ns ns
Indication in set-up time Indication in hold time	$t_{ m lin\_s}$ $t_{ m lin\_h}$	50 200			ns ns

9.7

#### **Electrical Characteristics**



# IOM<sup>®</sup>-2 Command/Indication Interface Timing (DCL = 2048 kHz)

## Figure 23

Table 28	Switching	Characteristics
----------	-----------	-----------------

Parameter	Symbol	Limit Values			Unit
		min.	typ.	max.	
Command out delay	t <sub>dCout</sub>			0	ns
Command out high impedance Command out active	$t_{dCZ}$ $t_{dCA}$		150 150	200 200	ns ns
Indication in set-up time Indication in hold time	t <sub>lin_s</sub> t <sub>lin_h</sub>	50 200			ns ns

## **Electrical Characteristics**

# 9.8 External Masterclock



# Figure 24

## Table 29 Switching Characteristics

Parameter	Symbol		Limit Values		
		min.	typ.	max.	
Period MCLK MCLK Duty Cycle	t <sub>MCLK</sub> t <sub>MCLKh</sub>	40	1/16.384	60	MHz %

#### Appendix

## 10 Appendix

#### 10.1 IOM<sup>®</sup>-2 Interface Monitor Transfer Protocol

#### **Monitor Channel Operation**

The monitor channel is used for the transfer of maintenance information between two functional blocks. Using two monitor control bits (MR and MX) per direction, the data are transferred in a complete handshake procedure. The MR and MX bits in the fourth octet (C/I channel) of the IOM2 frame are used for the handshake procedure of the monitor channel

The monitor channel transmission operates on a pseudo-asynchronous basis:

- Data transfer (bits) on the bus is synchronized to Frame Sync FSC
- Data flow (bytes) are asynchronously controlled by the handshake procedure.

For example: Data is placed onto the DD-Monitor-Channel by the Monitor-transmitter of the master device (DD-MX-Bit is activated i.e. set to '0'). This data transfer will be repeated within each frame (125  $\mu$ s rate) until it is acknowledged by the SLICOFI Monitor-receiver by setting the DU-MR-bit to '0', which is checked by the Monitor-transmitter of the master device. Thus, the data rate is not 8-Kbytes/s.



#### Figure 25

#### Monitor Handshake Procedure

The monitor channel works in 3 states

Idle state	A pair of inactive (set to '1') MR- and MX-bits during two or more consecutive frames: End of Message (EOM)
Sending state	MX-bit is activated (set to '0') by the Monitor-transmitter, together with data-bytes (can be changed) on the Monitor-channel
Acknowledging	MR-bit is set to active (set to '0') by the Monitor-receiver, together with a data-byte remaining in the Monitor-channel.

A start of transmission is initiated by a Monitor-transmitter in sending out an active MX-bit together with the first byte of data (the address of the receiver) to be transmitted in the Monitor-channel.

This state remains until the addressed Monitor-Receiver acknowledges the received data by sending out an active MR-bit, which means that the data-transmission is repeated each  $125 \,\mu s$  frame (minimum is one repetition). During this time the Monitor-transmitter evaluates the MR-bit.

Flow control, means in the form of transmission delay, can only take place when the transmitters MX and the receivers MR bit are in active state.

Since the receiver is able to receive the monitor data at least twice (in two consecutive frames), it is able to check for data errors. If two different bytes are received the receiver will wait for the receipt of two identical successive bytes (last look function)

A collision resolution mechanism (check if another device is trying to send data during the same time) is implemented in the transmitter. This is done by looking for the inactive ('1') phase of the MX-bit and making a per bit collision check on the transmitted monitor data (check if transmitted '1's are on DU/DD-line; DU/DD-line are open-drain lines).

Any abort leads to a reset of the SLICOFI command stack, the device is ready to receive new commands.

To obtain a maximum speed data transfer, the transmitter anticipates the falling edge of the receivers acknowledgment.

Due to the inherent programming structure, duplex operation is not possible. It is not **allowed** to send any data to the SLICOFI, while transmission is active.

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## Appendix



## Figure 26 State Diagram of the SLICOFI Monitor Transmitter

- MR ... MR bit received on DD line
- MX ... MX bit calculated and expected on DU line
- MXR ... MX bit sampled on DU line
- CLS ... Collision within the monitor data byte on DU line
- RQT ... Request for transmission form internal source
- ABT ... Abort request/indication

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## Appendix



## Figure 27 State Diagram of the SLICOFI Monitor Receiver

- MR ... MR bit calculated and transmitted on DU line
- MX ... MX bit received data downstream (DD line)
- LL ... Last lock of monitor byte received on DD line
- ABT ... Abort indication to internal source

## **Monitor Channel Data Structure**

The monitor channel is used for the transfer of maintenance information between two functional blocks. By use of two monitor control bits (MR and MX) per direction, the data are transferred in a complete handshake procedure.

## **Address Byte**

Messages to and from the SLICOFI are started with the following byte:

Bit	7	6	5	4	3	2	1	0
	1	0	0	0	0	0	0	1

Thus providing information for only one analog line, the SLICOFI is one device on one IOM-2 time slot. Monitor data for the analog channel is selected by the SLICOFI specific command (SOP, TOP or COP) following.

## 10.2 Channel Identification Command (CIC)

In order to unambiguously identify different devices by software, a two Byte identification command is defined for analog lines IOM-2 devices. A device requesting the identification of the SLICOFI will send the following 2 byte code:

1	0	0	0	0	0	0	0
0	0	0	0	0	0	0	0

Each device will then respond with its specific identification code. For the SLICOFI this two byte identification code is:

1	0	0	0	CONF			
1	0	0	0	0	0	0	0

**CONF** an optional 4-bit code indicating the specific hardware environment. A typical application of the CONF code is the differentiation of various types of line circuits that use the same SLICOFI/SLIC hardware within the same system.

For the realization of the Channel Identification Commands on the line card, it needs 3 pins at the SLICOFI. There are two inputs that can handle a ternary code (ID-L and ID-M). One pin is a binary input (ID-H) which is switchable as a digital serial interface of a shift register, to transfer special line card design informations up to 15 bytes into the monitor channel of the IOM-2 interface.

There are two different solutions of the CIC for the SLICOFI to identify the version of the line card.

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## Appendix



#### Figure 28

## Solution 1 ("Normal" Channel Identification Command):

The input of the 3 pin interface (ID-H, ID-L, ID-M) is transferred to the 4 bit CONF information using the following truth-table:

#### Table 30

	SLICOFI P	CONF-inf.		
ID-H	ID-M	ID-L	(4 bits)	
+ 5 V	– 5 V	– 5 V	0000	
+ 5 V	– 5 V	0 V	0001	
+ 5 V	– 5 V	+ 5 V	0010	
+ 5 V	0 V	– 5 V	0011	
+ 5 V	0 V	0 V	0100	
+ 5 V	0 V	+ 5 V	0101	
+ 5 V	+ 5 V	– 5 V	0110	
+ 5 V	+ 5 V	0 V	0111	
0 V	+ 5 V	0 V	1000	
0 V	+ 5 V	– 5 V	1001	
0 V	0 V	+ 5 V	1010	
0 V	0 V	0 V	1011	
0 V	0 V	– 5 V	1100	
0 V	– 5 V	+ 5 V	1101	
0 V	– 5 V	0 V	1110	
0 V	– 5 V	– 5 V	1111	

This is a 16 possible individual line card design information or an address pointer for the system to get more basic information.

The information is read through the IOM-2 monitor channel with the CIC command.

## Solution 2 (Extended Channel Identification Command):

The second realization step is that the combination of ports (M + L) = +5 V changes the input port ID-H to a shift register input.

#### Table 31

	SLICOFI Ports		CONF-inf.
ID-H	ID-M	ID-L	(4 bits)
X	+ 5 V	+ 5 V	1111

An external shift register on the line card transmits up to 15 bytes of special HW + FW line card design information (TCR4 - TCR18).

The information is read through the IOM2 monitor channel with the TOP Command. The LSEL bits TOP Command's register must be '10' - code for reading extended line card design and configuration information from TCR4 - TCR18 registers, which are sequential reading using two shift register. The CONF code is '1111' by this extended identification.

The first schematic gives an overview of the different timings for the extended channel identification.

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## Appendix



Figure 29 General Timing

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#### Expected Input of the ASIC (via ID - H; ID - L = ID - M 0 = +5 V)

If - for example - the SLICOFI has the time slot 6 (TSx = 110, see **chapter 4**, **page 16**, too), the Monitor Channel of TS6 looks like the following (for all other time slots equivalent).



#### Figure 30

Expected Input Timing and IOM-2 Interface Timing and Switching characteristic: To be defined.

#### 10.3 Test Modes

Various loops and tests (to cut off at different points or disable some filters) for testing either the chip or the board and the line are implemented in the SLICOFI.

LB	ТМ	Т3	T2	T1	T0	Testloop
SCR1-5	SCR2-3	SCR6-3	SCR6-2	SCR6-1	SCR6-0	
1	0	0	0	0	1	ALB_ADC
1	0	0	1	0	1	DLB_4M
1	0	1	0	0	0	DLB_PCM
1	0	1	1	0	0	DC_ALB
1	0	all	other combi	nations of	T3: T0	don't use
1	1	0	0	0	1	RVP
1	1	0	0	1	0	TVP

#### Table 32

LB	ТМ	Т3	T2	T1	Т0	Testloop
SCR1-5	SCR2-3	SCR6-3	SCR6-2	SCR6-1	SCR6-0	
1	1	0	0	1	1	LC
1	1	0	1	0	0	RC
1	1	0	1	0	1	ILT
1	1	0	1	1	0	DC-THRU
1	1	all	other comb	inations of	T3: T0	don't use
0	X	X	X	Х	X	all loops off

Table 32(cont'd)

#### Testregister (STCR1 to 8) - Summary

The Testregisters (accessed by the SOP-command with LSEL = 11b) are for internal use only. The 8 Testregisters can only be read or written en bloc. They are enabled/disabled by the Enable Testregister bit ENTR (SCR5-1). For ENTR = 0 the STCRs are set to the basic settings - so no refresh is necessary.

But note there are complex internal connections; so do use only the following two commands: ACDACDIS and EXT\_MCLK. All other bits MUST be set as described below.

#### **STCR1 Test Configuration Register 1**

Bit	7	6	5	4	3	2	1	0
	0	0	0	0	0	0	0	0

Reset value: 00<sub>H</sub>

#### STCR2 Test Configuration Register 2

Bit	7	6	5	4	3	2	1	0
	0	0	EXT_MCLK	0	0	ACDACDIS	0	0

Reset value: 00<sub>H</sub>

general remark All bits of STCR1 are set if necessary automatically by regular

testloops. So setting STCR1-bits to '1', together with a testloop, the certain action is inverted.

**EXT\_MCLK** Possibility to provide the SLICOFI with external clock (see also **page 35**, EXT\_MCLK, SCR8-4; There are no functional differences between these two settings!)

EXT\_MCLK = 0 Internal masterclock is used

EXT\_MCLK = 1 External masterclock is used

To use an external masterclock of 16 MHz following steps must be done:

- IO1 must be set to input and becomes the input-pin of the masterclock
- 2) Enable the testregisters (Configuration Register 5: SCR5-1 (ENTR) =1)
- The testregisterblock must be programmed (Test Configuration Register 2: STCR2-5 (EXT\_MCLK) = 1)
- ACDACDIS Disables AC-DAC ACDACDIS = 0 normal operation ACDACDIS = 1 disables AC-DAC

## **STCR3 Test Configuration Register 3**

Bit	7	6	5	4	3	2	1	0
	0	0	0	0	0	0	0	0

Reset value: 00<sub>H</sub>

## **STCR4 Test Configuration Register 4**

Bit	7	6	5	4	3	2	1	0
	0	1	0	1	1	1	1	1

Reset value:  $5F_H$ 

## **STCR5 Test Configuration Register 5**

Bit	7	6	5	4	3	2	1	0
	0	0	0	0	0	0	0	0

Reset value: 00<sub>H</sub>

## STCR6 to STCR8 Test Configuration Register 6 to 8

Bit	7	6	5	4	3	2	1	0
	0	0	0	0	0	0	0	0

Reset value: 00<sub>H</sub>

#### Appendix

# ALB\_ADC

(Analog loop with ADC and DAC)

This testloop feasibles the test of AC analog parts including ADC and DAC.

Initializing the testloop:

Reset

Active Mode

Disable Impedance matching filter



Figure 31

## DLB\_4M

(Digital loop up to 4 MHz)

This testloop feasibles the test of AC digital parts including DSP.

Initializing the testloop:

Reset

Store owns coefficients (generated by SLICOS)

Active Mode

Select programmed coefficients (FIXC (SCR5\_5) = 0)

Open Impedance matching and Transhybrid loop

(OPIM4M (SCR6\_4) = 1, IM (SCR4\_6) = 0, TH (SCR4\_7) = 0)



Figure 32

## Appendix

## DLB\_PCM

(Digital loop only PCM-interface)

This testloop is the basic setting after Reset and the NOT Active Mode.

It releases a shortcut between DD and DU. In Active Mode this loop can be programmed.

Initializing the testloop:

Reset

or in Active Mode:





# DC\_ALB

(DC analog loop)

This testloop feasibles the test of the analog DC parts (max. frequency of the testsignal 4 kHz).

Initializing the testloop:

Reset

Active Mode

Open analog loop (OPIMAN (SCR6\_5) = 1, ACDACDIS (STCR2\_2) = 1) Testloop



Figure 34

### Appendix

## RVP

(Ringer voltage present)

This testloop feasibles the test of the ringer burst level.

Initializing the testloop:

Reset

Store owns coefficients and voltage level for measurement

(generated by SLICOS)

Select programmed coefficients (FIXC (SCR5\_5) = 0)

Open analog loop (OPIMAN (SCR6\_5) = 1, ACDACDIS (STCR2\_2) = 1)

Ringing Mode, Ring Burst On (RBO) command

Testloop

Test condition is indicated in MVA (SCR2\_7) and result of the comparison is stored in OKRNG (SCR2\_4). The mean value can get at PCM Output, too.





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## TVP

(Teletax voltage present)

This testloop feasibles the test of the teletax burst level which includes the test of TTX adaptation and basic functions of HV-SLIC.

Initializing the testloop:

Reset

Store owns coefficients and voltage level for measurement (generated by SLICOS)

Select programmed coefficients (FIXC (SCR5\_5) = 0)

Active Mode, Teletax Burst On: TTXNO (SCR3\_7) = 0

Testloop

Test condition is indicated in MVA (SCR2\_7) and result of the comparison is stored in OKTTX (SCR2\_5). The rectified value can get at PCM Output, too. (During the testloop the last DC value is hold.)



#### Figure 36

# LC

(Loop current measurement)

This testloop feasibles a DC test of the line (shortcut, resistance, operating point) and basic function of the HV-SLIC.

Initializing the testloop:

Reset

Store owns coefficients (generated by SLICOS)

Select programmed coefficients (FIXC (SCR5\_5) = 0)

Open analog loop (OPIMAN (SCR6\_5) = 1, ACDACDIS (STCR2\_2) = 1)

- Active Mode
- Testloop





#### Appendix

# RC

(Ringer capacitance measurement)

This testloop feasibles the test of the line concerning the ringer.

Initializing the testloop:

Reset

Store owns coefficients (generated by SLICOS)

Select programmed coefficients (FIXC (SCR5\_5) = 0)

Open analog loop (OPIMAN (SCR6\_5) = 1, ACDACDIS (STCR2\_2) = 1)

Ringing Mode, Ring Burst On (RBO) command





## Appendix

## ILT

(Longitudinal current measurement)

This testloop feasibles the test of the line.

Initializing the testloop:

Reset

Store owns coefficients (generated by SLICOS)

Select programmed coefficients (FIXC (SCR5\_5) = 0)

Open analog loop (OPIMAN (SCR6\_5) = 1, ACDACDIS (STCR2\_2) = 1)

Active Mode

Testloop



## Figure 39

# DC\_THRU (DC loop)

This testloop feasibles the test of the DC parts.

Initializing the testloop:

Reset

PDown Mode (AC-Loop disactivated)



Figure 40

10.4	List of Abbreviations
Act	Active Mode
ADC	Analog Digital Converter
AGDCR	Attenuation DC Receive
AGDCX	Attenuation DC Transmit
AGR	Attenuation Receive
AGX	Attenuation Transmit
AGTTX	Attenuation Teletax
AR	Attenuation Receive
ASIC	Application Specific Integrated Circuit
AX	Attenuation Transmit
BB	Boosted Battery
BiCMOS	Bipolar Complementary Metal Oxid Semiconductor
BP	Band Pass
C/I-DD	Channel Identification-Data Downstream
C/I-DU	Channel Identification-Data Upstream
C1, 2	Digital Interface to HV-SLIC
CAP	External Capacitor to GNDA
CCITT	Commité Consultatif International de Telephone et Telegraph
CHOP	Chopper (see SCR8_6)
CMP	Compander
CODEC	Coder Decoder
COMP	Comparator (Testloops, Levelmetering)
COP	Coefficient Operation
CRAM	Coefficient RAM
DAC	Digital Analog Converter
DAC-HOL	D DC DAC Hold (Testloop TVP)
DBP	Deutsche Bundes Post
DCCHAR	DC Characteristic block
DCL	Data Clock
DD	Data Downstream

SI	Ε	Μ	Eľ	NS

DHP_R	Disable Receive Highpass (SCR5_7)
DHP_X	Disable Transmit Highpass (SCR1_1)
DSP	Digital Signal Processor
DU	Data Upstream
DUP	Data Upstream Persistency Counter
DUPGNK	Data Upstream Persistency Counter for GNK
EXP	Expander
FRR	Frequency Response Receive Filter
FRX	Frequency Response Transmit Filter
FSC	Frame Sync.
GNDIT	Analog Ground
GNK	Ground Key
HV-SLIC	High Voltage Subscriber Line Interface Circuit
11	Fixed Input Pin
ID-L	External Identification
ID-M	External Identification
IH-H	External Identification
IL	Longitudinal Current Input
IM	Impedance Matching Filter (programmable)
IMFIX	Impedance Matching Filter (fixed)
IO	User Programmable I/O Pin
IOM-2-Interface	ISDN Oriented Modular Interface
ISDN	Integrated Service Digital Network
IT	Transversal Current Input (for AC and DC)
ITAC	Transversal Current Input (for AC)

# Appendix

LP03 LP5 LSSGR	Low Pass 0.3 Hz Low Pass 5 Hz Local area transport access Switching System Generic Requirements
MEAN VAL.	Mean Value (Testloops, Levelmetering)
MR	Monitor Receive
MX	Monitor Transmit
O1	Fixed Output Pin
PCM	Pulse Code Modulation
PDen	Power Denial
PDN	Power Down
PDN	PDN Pin (Sets the HV SLIC to Power Denial)
POFI	Post Filter
PREFI	Antialiasing Pre Filter
RB	Ring Burst
RECT	Rectifier (Testloops, Levelmetering)
RES	Reset
REXT	External Ring Sync. Input
RFIX	Receive Filter (fixed)
RNG	Ring Generator
RREF	External Resistor to GNDA
SCR	Status Configuration Register
SEL24	Select Data Clock 2 or 4 MHz
SLIC	Subscriber Line Interface Circuit
SLICOS	SLICOFI Oriented Software
SLMA	Subscriber Loop Marging
SLXC	Summary Line Card Outputs
SOP	Status Operation
STCR	Status Test Configuration Register

# Appendix

TCR	Transfer Configuration Register
TE 1-3	Test Pin
TG	Tone Generator
ТН	Transhybrid Balancing
THFIX	Transhybrid Balancing Filter (fixed)
THRESH	Threshhold (Testloops, Levelmetering)
ТОР	Transfer Operation
TS	Time Slot
TS 0-2	Time Slot selection Pin
ТТХ	Teletax
TTXFI	Teletax Adaptation
TTXGEN	Teletax Generator
$V_{2W}$	Two Wire Output Voltage
V <sub>BIM</sub>	Battery Image Input
V <sub>LINE1, 2</sub>	Offhook-Detection in Power Denial Mode
LINE I, Z	
х	Transmit Filter (programmable)
	<b>U U V</b>

XFIX Transmit Filter (fixed)

PEB 3065 PEF 3065

#### **Package Outlines**

#### 11 Package Outlines





#### Sorts of Packing

Package outlines for tubes, trays etc. are contained in our Data Book "Package Information" SMD = Surface Mounted Device

Dimensions in mm