19-5200; Rev 3; 10/11

EVALUATION KIT AVAILABLE

# 

**Ordering Information** 

PACKAGE

8 TDFN-EP\*

8 µMAX

TOP

MARK

BML

\_

PIN-

8 SO

## 36V, Precision, Low-Noise, Wide-Band Amplifier

♦ 0.94nV/√Hz Ultra-Low Input Voltage Noise

Low Input Offset Voltage 125µV (max)

Gain-Bandwidth Product 55MHz

8-Pin SO and TDFN Packages

ESD 8kV HBM and 1kV CDM

+4.5V to +36V Wide Supply Range

**TEMP RANGE** 

-40°C to +125°C

-40°C to +125°C

-40°C to +125°C

+Denotes a lead(Pb)-free/RoHS-compliant package.

THD of -128dB at 10kHz

Rail-to-Rail Output

Unity-Gain Stable

PART

MAX9632ASA+

MAX9632ATA+

MAX9632AUA+

\*EP = Exposed pad.

Very Fast 600ns Settling Time to 16-Bit Accuracy

### **General Description**

The MAX9632 is a low-noise, precision, wide-band operational amplifier that can operate in a very wide +4.5V to +36V supply voltage range. The IC operates in dual (±18V) mode.

The exceptionally fast settling time and low distortion make the IC an excellent solution for precision acquisition systems. The rail-to-rail output swing maximizes the dynamic range when driving high-resolution 24-bit  $\Sigma\Delta$ ADCs even with low supply voltages.

The IC achieves 55MHz of gain-bandwidth product and ultra-low 0.94 nV/ $\sqrt{Hz}$  input voltage noise with only 3.9mA of quiescent current.

The IC is offered in 8-pin SO, µMAX®, and TDFN packages and is rated for operation over the -40°C to +125°C temperature range.

#### Applications

High-Resolution ADC Drivers High-Resolution DAC Buffers Medical Imaging Low-Noise Signal Processing Test and Measurement Systems ATE

µMAX is a registered trademark at Maxim Integrated Products, Inc.



#### **MIXI/N**

Maxim Integrated Products 1

**Features MAX9632** ♦ Low Input Offset Temperature Drift 0.5µV/°C (max)

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

VCC to VEE	0.3V to +40V
All Other Pins(VEE - 0.3V) to (	(VCC + 0.3V)
Short-Circuit (GND) Duration, OUT	10s
Continuous Input Current (any pin)	±20mA
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ ) (Note	1)
Multilever CO (derete 7 $4m/1/2$ C ebays $70^{\circ}$ C)	E O O ma\A/

Multilayer SO (derate 7.4mW/°C above +70°C) .......588mW Multilayer TDFN (derate 23.8mW/°C above +70°C)...1905mW Multilayer µMAX (derate 4.8mW/°C above +70°C)..387.8mW

ESD Protection	
HBM	8kV
CDM	1kV
Operating Temperature Range	40°C to +125°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

Junction-to-Ambient Thermal Resistance (0,1A) ..... 206.3°C/W Junction-to-Case Thermal Resistance ( $\theta_{JC}$ )......42°C/W

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

8 µMAX

### PACKAGE THERMAL CHARACTERISTICS (Note 1)

6 TDFN	•	
Junction-to-Ambient Thermal Resistance $(\theta_{JA})$	42°C/W	
Junction-to-Case Thermal Resistance (0JC)	8°C/W	
8 SO		
Junction-to-Ambient Thermal Resistance $(\theta_{JA})$	136°C/W	
Junction-to-Case Thermal Resistance (Aug)	38°C/W	

#### Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

### ELECTRICAL CHARACTERISTICS

(VCC = 15V, VEE = -15V, RL = 10kΩ to VGND, VIN+ = VIN- = VGND = 0V, VSHDN = VGND, TA = -40°C to +125°C. Typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER SUPPLY	-					
Supply Voltage Range	Vcc	Guaranteed by PSRR	4.5		36	V
Supply Current	Icc			3.9	6.5	mA
Power Supply Paination Patio	PSRR	$T_A = +25^{\circ}C$	125	140		dD
Power-Supply Rejection Ratio	Fonn	$-40^{\circ}C \le T_A \le +125^{\circ}C$	120			dB
SHUTDOWN						
Shutdown Input Voltage	VOUDN	Device disabled	VCC - 0.35		V <sub>CC</sub>	V
	VSHDN	Device enabled	VEE		VCC - 3.0	
Shutdown Current	ISHDN	V <sub>SHDN</sub> = V <sub>CC</sub>		1	15	μA
DC SPECIFICATIONS	·					
Input Offect Veltage	Vee	$T_A = +25^{\circ}C$		30	125	
Input Offset Voltage	Vos	$-40^{\circ}C \le T_A \le +125^{\circ}C$			165	μV
Input Offset Voltage Drift	±ΔVos	(Note 3)		0.15	0.5	µV/°C
Input Bias Current	IB			30	180	nA
Input Offset Current	los			15	100	nA
Input Common-Mode Range	VCM	Guaranteed by CMRR	V <sub>EE</sub> + 1.8		Vcc - 1.4	V

///XI//

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{CC} = 15V, V_{EE} = -15V, R_L = 10k\Omega$  to V<sub>GND</sub>, V<sub>IN+</sub> = V<sub>IN-</sub> = V<sub>GND</sub> = 0V, V<sub>SHDN</sub> = V<sub>GND</sub>, T<sub>A</sub> = -40°C to +125°C. Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CON	DITIONS	MIN	TYP	MAX	UNITS
		$V_{EE} + 1.8V \le V_{CM} \le$	$V_{CC} - 1.4V, T_{A} = +25^{\circ}C$	120	135		
Common-Mode Rejection Ratio	CMRR	$\begin{array}{l} V_{EE} + 1.8V \leq V_{CM} \leq \\ -40^{\circ}C \leq T_A \leq +125^{\circ}C \end{array}$		110			dB
Large-Signal Gain	Avol	$V_{EE}$ + 0.2V $\leq$ V <sub>OUT</sub> :	$V_{\text{EE}}$ + 0.2V $\leq$ V <sub>OUT</sub> $\leq$ V <sub>CC</sub> - 0.2V, R <sub>L</sub> = 10k $\Omega$		140		dB
	AVOL	$V_{EE} + 0.6V \le V_{OUT}$	$VEE + 0.6V \le VOUT \le VCC - 0.6V, RL = 600\Omega$		135		
	Vон	Vcc - Vout	$R_L = 10k\Omega$		50	150	
Output Voltage Swing	VOH	VCC - VOUI	$R_L = 600\Omega$		150	400	mV
Culput voltage Swing	Vol		$R_L = 10k\Omega$		50	150	
	VOL	Vout - Vee	$R_L = 600\Omega$		150	400	]
Short-Circuit Current	Isc	$T_A = +25^{\circ}C$			56		mA
AC SPECIFICATIONS							
Gain-Bandwidth Product	GBWP				55		MHz
Slew Rate	SR	$0 \le V_{OUT} \le 5V$			30		V/µs
Settling Time	ts	To 0.0015%, V <sub>OUT</sub> = AV = 1V/V	= 10V <sub>P-P</sub> , C <sub>L</sub> = 100pF,		600		ns
Total Harmonic Distortion	THD	f = 1kHz, V <sub>OUT</sub> = 3\ = 1V/V	/RMS, $R_L = 600\Omega$ , AV		-136		- dB
		$f = 10kHz, V_{OUT} = 3$ $= 1V/V$	$3V_{\text{RMS}}$ , $R_{\text{L}} = 600\Omega$ , AV		-128		
Input-Voltage Noise Density	eN	f = 1kHz			0.94		nV/√Hz
Input Voltage Noise		$0.1Hz \le f \le 10Hz$			65		nV <sub>P-P</sub>
Input-Current Noise Density	iN	f = 1kHz			3.75		pA/√Hz
Capacitive Loading	CL	No sustained oscilla	tion, AV = 1V/V		350		pF

**Note 2:** All devices are 100% production tested at  $T_A = +25$ °C. Temperature limits are guaranteed by design. **Note 3:** Guaranteed by design.

 $(V_{CC} = 15V, V_{EE} = -15V, R_L = 10k\Omega$  to  $V_{GND}, V_{IN+} = V_{IN-} = V_{GND} = 0V$ ,  $V_{SHDN} = V_{GND}, T_A = -40^{\circ}C$  to  $+125^{\circ}C$ . Typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.)



### **Typical Operating Characteristics**

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 $(V_{CC} = 15V, V_{EE} = -15V, R_L = 10k\Omega$  to  $V_{GND}$ ,  $V_{IN+} = V_{IN-} = V_{GND} = 0V$ ,  $V_{SHDN} = V_{GND}$ ,  $T_A = -40^{\circ}C$  to  $+125^{\circ}C$ . Typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.)



### **Typical Operating Characteristics (continued)**

**POWER-UP TIME** 

 $V_{CC} = 5V, V_{EE} = -5V$ 

1µs/div

vs. FREQUENCY

1k FREQUENCY (Hz) 10k

100k

(V<sub>CC</sub> = 15V, V<sub>EE</sub> = -15V, R<sub>L</sub> = 10kΩ to V<sub>GND</sub>, V<sub>IN+</sub> = V<sub>IN-</sub> = V<sub>GND</sub> = 0V, V<sub>SHDN</sub> = V<sub>GND</sub>, T<sub>A</sub> = -40°C to +125°C. Typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.)





## Pin Configuration



### Pin Description

PIN	NAME	FUNCTION
1, 5	N.C.	Not Connected
2	IN-	Negative Input
3	IN+	Positive Input
4	VEE	Negative Supply Voltage
6	OUT	Output
7	Vcc	Positive Supply Voltage
8	SHDN	Active-High Shutdown
	EP	Exposed Pad (TDFN Only). Connect to a large VEE plane to maximize thermal performance. Not intended as an electrical connection point.

### \_Detailed Description

The MAX9632 is designed in a new 36V, high-speed complementary BiCMOS process that is optimized for excellent AC dynamic performance combined with high-voltage operation.

The IC offers precision, high-bandwidth, ultra-low noise and exceptional distortion performance.

The IC is unity-gain stable and operates either with single-supply voltage up to 36V or with dual supplies up to  $\pm 18V$ .

### **Applications Information**

#### **Operating Supply Voltage**

The IC can operate with dual supplies from  $\pm 2.25V$  to  $\pm 18V$  or with a single supply from  $\pm 4.5V$  to  $\pm 36V$  with respect to ground. Even though the IC supports high-voltage operation with excellent performance, the device can also operate in very popular applications at 5V.

#### Low Noise and Low Distortion

The IC is designed for extremely low-noise applications such as professional audio equipment, very high performance instrumentations, automated test equipment, and medical imaging. The low noise, combined with fast settling time, makes it ideal to drive high-resolution sigmadelta or SARs analog-to-digital converters.

The IC is also designed for ultra-low-distortion performance. THD specifications in the *Electrical Characteristics* table and *Typical Operating Characteristics* are calculated up to the fifth harmonic. Even when driving highvoltage swing up to 10VP-P, the IC maintains excellent low distortion operation over and above 100kHz of bandwidth.

#### **Rail-to-Rail Output Stage**

The output stage swings to within 50mV (typ) of either power-supply rail with a 10k $\Omega$  load and provides a 55MHz GBW with a 30V/s slew rate. The device is unity-gain stable and can drive a 100pF capacitive load without compromising stability. Stability with higher capacitive loads can be improved by adding an isolation resistor in series with the op-amp output. This resistor improves the circuit's phase margin by isolating the load capacitor from the amplifier's output. The *Typical Operating Characteristics* show a profile of the isolation resistor and capacitive load values that maintain the device into the stable region.



Figure 1. Input Protection Circuit

#### **Input Differential Voltage Protection**

During normal op-amp operation, the inverting and noninverting inputs of the IC are at essentially the same voltage. However, either due to fast input voltage transients or other fault conditions, these inputs can be forced to be at two different voltages.

Internal back-to-back diodes protect the inputs from an excessive differential voltage (Figure 1). Therefore, IN+ and IN- can be any voltage within the range shown in the *Absolute Maximum Ratings* section. Note the protection time is still dependent on the package thermal limits.

If the input signal is fast enough to create the internal diodes' forward bias condition, the input signal current must be limited to 20mA or less. If the input signal current is not inherently limited, an input series resistor can be used to limit the signal input current. Care should be taken in choosing the input series resistor value, since it degrades the low-noise performance of the device.

#### Shutdown

The shutdown is referenced to the positive supply. See the *Electrical Characteristics* table for the proper levels of functionality. A high level (above V<sub>CC</sub> - 0.35V) disables the op amp and puts the output into a high-impedance state. A low level (below V<sub>CC</sub> - 3V) enables the device. As an example, if the op amp is powered with dual supplies of  $\pm 15V$ , the device is enabled when shutdown is at or below 12V. The device is disabled when shutdown is at or above 14.65V. If the op amp is powered with a single supply of 36V, the device is enabled when shutdown is at or below 33V. The device is disabled when shutdown is at or above 35.65V. This input must be connected to a valid high or low voltage and should not be left disconnected.

#### **Power Supplies and Layout**

The MAX9632 can operate with dual supplies from  $\pm 2.25V$  to  $\pm 18V$  or with a single supply from  $\pm 4.5V$  to  $\pm 36V$  with respect to ground. When used with dual supplies, bypass both VCC and VEE with their own 0.1µF capacitor to ground. When used with a single supply, bypass VCC with a 0.1µF capacitor to ground.



Careful layout technique helps optimize performance by decreasing the amount of stray capacitance at the op amp's inputs and outputs. To decrease stray capacitance, minimize trace lengths by placing external components close to the op amp's pins.

For high-frequency designs, ground vias are critical to provide a ground return path for high-frequency signals and should be placed near the decoupling capacitors. Signal routing should be short and direct to avoid parasitic effects. Avoid using right angle connectors since they may introduce a capacitive discontinuity and ultimately limit the frequency response.

#### **Electrostatic Discharge (ESD)**

The IC has built-in circuits to protect it from ESD events. An ESD event produces a short, high-voltage pulse that is transformed into a short current pulse once it discharges through the device. The built-in protection circuit provides a current path around the op amp that prevents it from being damaged. The energy absorbed by the protection circuit is dissipated as heat.

ESD protection is guaranteed up to  $\pm 8$ kV with the Human Body Model (HBM). The Human Body Model simulates the ESD phenomenon wherein a charged body directly transfers its accumulated electrostatic charge to the ESD-sensitive device. A common example of this phenomenon is when a person accumulates static charge by walking across a carpet and then transfers all of the charge to an ESD-sensitive device by touching it.

Not all ESD events involve the transfer of charge into the device. ESD from a charged device to another body is also a common form of ESD.

If a charged device comes into contact with another conductive body that is at a lower potential, it discharges into that body. Such an ESD event is known as Charged Device Model (CDM) ESD, which can be even more destructive than HBM ESD (despite its shorter pulse duration) because of its high current. The IC guarantees CDM ESD protection up to  $\pm 1$ kV.

**Driving High-Resolution Sigma-Delta ADCs** The MAX9632's excellent AC specifications and 55MHz bandwidth are a good fit for driving high-speed, precision delta-sigma ADCs. These ADCs require an ultra-low noise op amp to achieve signal-to-noise ratios (SNR) better than 100dB. The MAX11040 is a 24-bit, 4-channel, simultaneous-sampling ADC with 117dB SNR at 1ksps and 106dB at 16ksps. The MAX11040 measures analog inputs up to  $\pm 2.2$ V. Sampling up to 64ksps, the MAX11040 achieves better than -94dB THD and 94dB SFDR.

The MAX11040 measures four differential inputs simultaneously, outputting the data through an SPI™ interface to allow daisy-chaining the data outputs and inputs together. Therefore, up to eight MAX11040 devices can be placed in parallel to measure up to 32 inputs simultaneously. This is ideal for 3-phase power monitoring that requires multiple current and voltage readings and very wide dynamic range.

The *Typical Application Circuit* shows an example of the MAX9632 driving the MAX11040.

**Chip Information** 

PROCESS: BICMOS

SPI is a trademark of Mototrola, Inc.



**Typical Application Circuit** 



### **Package Information**

For the latest package outline information and land patterns (footprints), go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 SO	S8+2	<u>21-0041</u>	<u>90-0096</u>
8 TDFN-EP	T833+3	<u>21-0137</u>	<u>90-0060</u>
8 µMAX	U8+3	<u>21-0036</u>	<u>90-0092</u>



## Package Information (continued)

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COMMON	I DIMEN	SIONS		PACKAGE V	ARIAT	IONS					
SYMBOL	MIN.	MAX.		PKG. CODE	N	D2	E2	е	JEDEC SPEC	b	[(N/2)-1] x e
А	0.70	0.80		T633-2	6	1.50±0.10	2.30±0.10	0.95 BSC	MO229 / WEEA	0.40±0.05	1.90 REF
D	2.90	3.10		T833-2	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
Е	2.90	3.10		T833-3	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
A1	0.00	0.05		T1033-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF
L	0.20	0.40		T1033MK-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF
k	0.25	MIN.		T1033-2	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF
A2	0.20	REF.		T1433-1	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
				T1433-2	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
				T1433-3F	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
NOTES:			- 161		DEOE						
1. ALL D 2. COPL 3. WARP 4. PACK 5. DRAW 6. "N" IS 7. NUMB 8. MARK	ANARITY AGE SH AGE LEI ING CO S THE ER OF ING IS	SHALL NGTH/P/ NFORMS TOTAL N LEADS FOR PA	NOT EXC F EXCEEL ACKAGE N TO JED UMBER O SHOWN A CKAGE O	OF LEADS. ARE FOR REF RIENTATION R	DNSID XCEP EREN REFER	ERED AS S T DIMENSIO CE ONLY. ENCE ONLY	NS "D2" AN	ND "E2", AN	C(S). ID T1433-1 & T <sup>1</sup>	1433–2.	
1. ALL D 2. COPL 3. WARP 4. PACK 5. DRAW 6. "N" IS 7. NUMB 8. MARK	ANARITY AGE SH AGE LEI ING CO S THE ER OF ING IS	SHALL NGTH/P/ NFORMS TOTAL N LEADS FOR PA	NOT EXC F EXCEEL ACKAGE N TO JED UMBER O SHOWN A CKAGE O	CEED 0.08 m 0 0.10 mm. WIDTH ARE CO EC MO229, E DF LEADS. ARE FOR REF	DNSID XCEP EREN REFER	ERED AS S T DIMENSIO CE ONLY. ENCE ONLY	NS "D2" AN	ND "E2", AN	ND T1433-1 & T		XI/V

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**WAX9632** 

### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	10/10	Initial release	—
1	4/11	Updated short-circuit current spec	3
2	8/11	Updated TDFN land pattern number	11
3	10/11	Added µMAX package	1, 2, 7

**MAX9632** 

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