LTC1564



10kHz to 150kHz Digitally Controlled Antialiasing Filter and 4-Bit P.G.A

The LTC[®]1564 is a new type of continuous time filter for antialiasing, reconstruction and other band-limiting appli-

cations. No other analog components or filter expertise are

needed to use it. There is one analog input pin and one

analog output pin. The cutoff frequency $(f_{\rm C})$ and gain are

programmable while the shape of the lowpass response is

fixed. A latching digital interface stores f_C and gain settings or it can be bypassed for control directly from the pins. The

LTC1564 operates from 2.7V to 10V total (single or split

The LTC1564 is a rail-to-rail high resolution 8th-order

lowpass filter with two stopband notches, giving approxi-

mately 100dB attenuation at 2.5 times the passband cutoff

frequency f_C (a de-facto standard for DSP front ends). Signals with low or variable levels can be normalized with

the built-in variable gain that reduces input-referred noise

with increasing gain for a typical dynamic range (maxi-

supplies) and comes in a 16-pin surface mount SSOP.

DESCRIPTION

FEATURES

- 4-Bit Digitally Controlled 8th-Order Lowpass Filter
 - f_{CUTOFF} Adjustable from 10kHz to 150kHz in 10kHz Steps
 - 100dB Attenuation at $2.5 \times f_{CUTOFF}$
- 4-Bit Digitally Controlled Programmable Gain Amplifier
 - G = 1 to 16 in 1V/V Steps
- Miniature 16-Pin SSOP Package
- No External Components
- 122dB Total System Dynamic Range
- Rail-to-Rail Input and Output Range
- 2.7V to 10V Operation
- Low Noise Mute Mode
- Low Power Shutdown Mode

APPLICATIONS

- Antialias or Reconstruction Filtering
- DSP Systems
- Communications Systems
- Scientific Instruments
- High Resolutions (16 Bits to 20 Bits)
- Processing Signals Buried in Noise
- Audio Signal Processing
- Programmable Data Rates
- Automatic Gain Control (AGC)
- Single Part Replacing Multiple Filters

mum signal level to minimum noise) of 122dB (20 equivalent bits) with 20kHz f_{C} and 118dB at 100kHz f_{C} on a ±5V

supply.

Other frequency-response shapes can be provided upon request. Please contact LTC Marketing.

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TYPICAL APPLICATION

Low Noise Programmable Filter with Variable Gain



V⁺ AND V⁻ SUPPLIES CAN BE FROM 1.35V TO 5.25V EACH TIE F AND G PINS TO V⁺ OR V⁻ TO SET FREQUENCY AND GAIN DYNAMIC RANGE 118dB TO 122dB AT \pm 5V DEPENDING ON FREQUENCY CODE

LTC1564 Programmable Range





ABSOLUTE MAXIMUM RATINGS

(Note 1)

Total Supply Voltage (V ⁺ to V ⁻) 11V Input Voltage V ⁺ + 0.3V to V ⁻ – 0.3V
Output Short-Circuit Duration Indefinite
Operating Temperature Range
LTC1564C 0°C to 70°C
LTC1564I40°C TO 85°C
Storage Temperature Range –65°C to 150°C
Lead Temperature (Soldering, 10 sec) 300°C

PACKAGE/ORDER INFORMATION



Consult factory for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS The \bullet denotes specifications that apply over the full operating temperature range, otherwise specifications are at T_A = 25°C. V_S = ±2.375V, f_C = 10kHz, gain = 1, R_L = 10k, unless otherwise noted.

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Total Supply Voltage			2.7		10.5	V
Supply Current	$ \begin{array}{l} {\sf V}_S=\pm 1.35{\sf V},{\sf V}_{{\sf IN}}=0{\sf V}\\ {\sf V}_S=\pm 2.375{\sf V},{\sf V}_{{\sf IN}}=0{\sf V}\\ {\sf V}_S=\pm 5{\sf V},{\sf V}_{{\sf IN}}=0{\sf V} \end{array} $	•		15 16 22	17 18.5 25	mA mA mA
Output Voltage Swing	$R_L = 10k \text{ to } 0V$	•	4.5	4.65		V _{P-P}
Output Short-Circuit Current	$V_{\rm S} = \pm 5 V$	•		±10		mA
DC Offset Voltage Magnitude (Referred to Input)	Gain = 1, 0°C to 70°C Gain = 1, -40°C to 85°C Gain = 10, 0°C to 70°C Gain = 10, -40°C to 85°C	•		3 3 1 1	13 16 5 6	mV mV mV mV
DC AGND Reference Voltage	V _S = Single 5V Supply			2.5		V
Passband Gain	$f_C = 50$ kHz, $f_{IN} = 10$ kHz, Gain = 1 $f_C = 50$ kHz, $f_{IN} = 10$ KHz, Gain = 16	•	-0.1 23.5	0.3 24.2	0.8 25.3	dB dB
Passband Ripple	$ \begin{array}{l} f_C = 10 \text{kHz}, \ 0 \leq f_{\text{IN}} \leq \ 9 \text{kHz} \ (\text{Notes 2, 3}) \\ f_C = 150 \text{kHz}, \ 0 \leq f_{\text{IN}} \leq 135 \text{kHz} \ (\text{Notes 2, 3}) \end{array} $	•	-0.5 -0.6		0.5 1.6	dB dB
Roll Off at Cutoff Frequency (f_C) (Note 3)	$f_{C} = 10 \text{kHz} (F = 0001)$ $f_{C} = 150 \text{kHz} (F = 1111)$	•	-1.2 -1.5	-0.7 -0.5	-0.3 0.6	dB dB
Roll Off at 2f _C (Note 3)	f _C = 10kHz		-65	-62	-59	dB
Roll Off at 2.5f _C (Note 3)	f _C = 10kHz			-99		dB
Wideband Noise (Referred to Input)	$\begin{array}{l} BW = 20kHz, f_{C} = 10kHz, Gain = 1\\ BW = 20kHz, f_{C} = 10kHz, Gain = 16\\ BW = 200kHz, f_{C} = 100kHz, Gain = 1 \end{array}$			33 2.5 50		μV _{RMS} μV _{RMS} μV _{RMS}
Total Harmonic Distortion	f_{C} = 100kHz, f_{IN} = 10kHz, V_{IN} = 1 V_{RMS}			-86		dB
Input Impedance	$ Gain = 1, DC V_{IN} = 0V \\ Gain = 16, DC V_{IN} = 0V $			10 625		kΩ Ω
Output Impedance	f _C = 10kHz, f = 10kHz			30		Ω
Mute State (F = 0000) Gain	F = 0000, f _{IN} = 20kHz, V _{IN} = 1V _{RMS}			-103		dB



ELECTRICAL CHARACTERISTICS The \bullet denotes specifications that apply over the full operating temperature range, otherwise specifications are at T_A = 25°C. V_S = ±2.375V, f_C = 10kHz, gain = 1, R_L = 10k, unless otherwise noted.

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Mute State Output Noise	F = 0000, BW = 200kHz			5.4		μV _{RMS}
Shutdown Supply Current	$V_{S} = \pm 1.35V$, \overline{EN} to V ⁺ $V_{S} = \pm 1.35V$, \overline{EN} to V ⁺	•		45	75 150	μΑ μΑ
	$V_S = \pm 2.375V$, $\overline{EN} \text{ to } V^+$ $V_S = \pm 2.375V$, $\overline{EN} \text{ to } V^+$	•		100	150 180	μΑ μΑ
	$V_{\rm S} = \pm 5 V, \ \overline{\rm EN} \ {\rm to} \ V^+ \ ({\rm Note} \ 4)$			175		μA
Digital Input "High" Voltage	$V_{S} = \pm 1.35V$ $V_{S} = \pm 2.375V$ $V_{S} = \pm 5V$		1.08 1.90 4.50			V V V
Digital Input "Low" Voltage	$V_{S} = \pm 1.35V$ $V_{S} = \pm 2.375V$ $V_{S} = \pm 5V$				-1.08 -1.90 0.50	V V V
Digital Input Pull-Up or Pu <u>ll-</u> Down Current (Note 5) (Digital Inputs Other than EN)	$V_{S} = \pm 1.35V$ $V_{S} = \pm 5V$	•		3.5 13	6 20	μΑ μΑ
Digital Input Pull-Up Current (EN Input)	$V_S = \pm 1.35V$ $V_S = \pm 5V$	•		1 10	2 20	μΑ μΑ

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

Note 2: Response is tested in production at discrete frequencies f_{IN} of 0.1, 0.5, 0.8 and 0.9 times $f_{\rm C}$.

Note 3: Relative to gain at 0.1f_C.

Note 4: All digital inputs driven rail-to-rail. When driving digital inputs with OV and 5V levels, the shutdown current will increase to 3.5mA (typ).

Note 5: Each digital input includes a small positive or negative current source to float the CMOS input to V⁺ or V⁻ potential if it is unconnected. The table shows the current due to this source when the input is driven at the supply voltage opposite from the float potential. Pins \overline{CS} /HOLD, F3, F2, F0 and G3 to G0 float to the V⁻ voltage, pins RST, EN and F1 to the V⁺ voltage. See "Floatable Digital Inputs" in Applications Information section.

TYPICAL PERFORMANCE CHARACTERISTICS





TYPICAL PERFORMANCE CHARACTERISTICS





Detail of Stopband Response



Rectangular Pulse Response



Short-Pulse Response



THD + Noise vs Input Voltage $(f_{C} = 10kHz)$



Triangular-Wave Time Response









TYPICAL PERFORMANCE CHARACTERISTICS



PIN FUNCTIONS

OUT (Pin 1): Analog Output. In normal filtering, this is the output of an internal operational amplifier and is capable of swinging essentially to any voltage between the power supply rails (that is, between V^+ and V^-). This output is designed to drive a nominal load of 5k and 50pF. For lowest signal distortion it should be loaded as lightly as possible. The output can drive lower resistances than 5k, but distortion may increase, and the output current will limit at approximately ±10mA. Capacitances higher than 50pF should be isolated by a series resistor of 500Ω to preserve AC stability. In the Mute state (F code 0000 or $\overline{RST} = 0$), the output operates as in normal filtering but the gain from the IN pin becomes zero and the output noise is reduced. In the shutdown state (EN = 1 or EN open circuited), most of the circuitry in the LTC1564 shuts off and the OUT pin assumes a high impedance state.

V⁻, V⁺ (Pins 2, 14): Power Supply Pins. The V⁺ and V⁻ pins should be bypassed with 0.1μ F capacitors to an adequate analog ground plane using the shortest possible wiring. Electrically clean supplies and a low impedance ground are important for the high dynamic range and high stopband suppression available from the LTC1564 (see further details under AGND). Low noise linear power supplies are recommended. Switching supplies are not recommended because of the inevitable risk of their

switching noise coupling into the signal path, reducing dynamic range.

EN (Pin 3): CMOS-Level Digital Chip Enable Input. Logic 1 or open circuiting this pin causes a shutdown mode with reduced supply current. The active circuitry in the LTC1564 shuts off and its output assumes a high impedance state. If F and G bits are latched (\overline{CS} /HOLD = 1) during the shutdown state, the latch will retain its contents.

A small pull-up current source at the $\overline{\text{EN}}$ input causes the LTC1564 to be in shutdown state if the $\overline{\text{EN}}$ pin is left open. Therefore, the user must connect the $\overline{\text{EN}}$ pin to logic 0 (V⁻ or optionally 0V with ±5V supplies) for normal filter operation.

CS/HOLD (Pin 4): CMOS-Level Digital Enable Input for the Latch Holding F and G Bits. Logic 0 makes the latch transparent so that the F and G inputs directly control the filter's cutoff frequency and gain. Logic 1 holds the last values of these inputs prior to the transition. This pin floats to logic 0 (V⁻) when open circuited because of a small current source (see Electrical Characteristics, Note 5).

F3, **F2**, **F1**, **F0** (**Pins 5**, **6**, **7**, **8**): CMOS-Level Digital Frequency Control ("F Code") Inputs. F3 is the most significant bit (MSB). These pins program the LTC1564's cutoff frequency f_C through the internal latch, which



PIN FUNCTIONS

passes the bits directly when the \overline{CS} /HOLD input is at logic 0. When \overline{CS} /HOLD changes to logic 1, the F pins cease to have effect and the latch holds the previous values. The F code controls the filter's cutoff frequency f_C in 10kHz steps up to 150kHz, as summarized in Table 1.

|--|

F3 (AT OUTPU	F2 T OF I		FO Al Latch)	NOMINAL F _C (CUTOFF FREQUENCY)
0	0	0	0	0 (Mute State: Filter Gain is Zero)
0	0	0	1	10kHz
0	0	1	0	20kHz
0	0	1	1	30kHz
0	1	0	0	40kHz
0	1	0	1	50kHz
0	1	1	0	60kHz
0	1	1	1	70kHz
1	0	0	0	80kHz
1	0	0	1	90kHz
1	0	1	0	100kHz
1	0	1	1	110kHz
1	1	0	0	120kHz
1	1	0	1	130kHz
1	1	1	0	140kHz
1	1	1	1	150kHz

Thus f_C is proportional to the binary value of the F code. Note that small current sources pull F1 to V⁺ and F3, F2 and F0 to V⁻ when these pins are left unconnected (see Electrical Characteristics, Note 5). This sets an F code input of 0010 (2, in decimal form) by default, giving an f_C of 20kHz in normal filtering operation, if CS/HOLD is logic 0 or is open circuited.

Tahlo 2

G0, G1, G2, G3 (Pins 9, 10, 11, 12): CMOS-Level Digital Gain Control ("G Code") Inputs. G3 is the most significant bit (MSB). These pins program the LTC1564's passband gain through the internal latch, which passes the bits directly when the \overline{CS} /HOLD input is at logic 0. When \overline{CS} /HOLD changes to logic 1, the G pins cease to have effect and the latch retains the previous input values. This gain control is linear in amplitude: nominal passband gain of the LTC1564 is the binary value of the G code, plus one as shown in Table 2.

Note that small current sources pull the G pins to V⁻ when these pins are left unconnected (see Electrical Characteristics, Note 5). This sets a G code input of 0000 by default, giving unity passband gain in normal filtering operation, if \overline{CS} /HOLD is logic 0 or is open circuited.

RST (Pin 13): CMOS-Level Asynchronous Reset Input. Logic 0 on this pin immediately resets the internal F and G latch to all zeros, regardless of the state of the CS/HOLD pin or the F or G input pins. This causes the LTC1564 to enter a mute state (powered but with zero signal gain) because of the resulting F = 0000 command. Logic 1 permits the other pins to control F and G. This pin floats to logic 1 (V⁺) when open circuited because of a small current source (see Electrical Characteristics, Note 5). A brief internal reset (shorter than the analog settling time of the filter) also occurs when power is first applied.

Table 2									
G3 (AT OUTP	G2 Put of In	G1 Ternal	GO Latch)	NOMINA Passband (Volt/Volt)	GAIN		/I INPUT SIG Is Peak-to Single 5v		NOMINAL INPUT IMPEDANCE (kΩ)
0	0	0	0	1	0	10	5.0	3.0	10
0	0	0	1	2	6.0	5	2.5	1.5	5
0	0	1	0	3	9.5	3.33	1.67	1.0	3.33
0	0	1	1	4	12	2.5	1.25	0.75	2.5
0	1	0	0	5	14.0	2	1	0.6	2
0	1	0	1	6	15.6	1.67	0.83	0.5	1.67
0	1	1	0	7	16.9	1.43	0.71	0.43	1.43
0	1	1	1	8	18.1	1.25	0.63	0.38	1.25
1	0	0	0	9	19.1	1.1	0.56	0.33	1.11
1	0	0	1	10	20.0	1.0	0.50	0.30	1
1	0	1	0	11	20.8	0.91	0.45	0.27	0.91
1	0	1	1	12	21.6	0.83	0.42	0.25	0.83
1	1	0	0	13	22.3	0.77	0.38	0.23	0.77
1	1	0	1	14	22.9	0.71	0.36	0.21	0.71
1	1	1	0	15	23.5	0.67	0.33	0.20	0.66
1	1	1	1	16	24.1	0.63	0.31	0.19	0.63



PIN FUNCTIONS

Table 3. Summary of LTC1564 Digital Controls and Modes

EN	RST	CS/HOLD	F3	F2	F1	FO	G3	G2	G1	GO	FUNCTION
1	1	1	Х	Х	Х	Х	Х	X X X X Shutdown Mode. Filter Disabled. Latch Holds F and G when Last \overline{CS} /HOLD = 0		Shutdown Mode. Filter Disabled. Latch Holds F and G Inputs Present when Last $\overline{\text{CS}}/\text{HOLD}$ = 0	
1	1	0	Х	Х	Х	Х	Х	X X X X Shutdown Mode. Filter Disabled. Latch Accepts F and G Inputs		Shutdown Mode. Filter Disabled. Latch Accepts F and G Inputs	
1	0	Х	Х	Х	Х	Х	Х	Х	Х	X X Shutdown Mode. Filter Disabled. Latch Contents (F and G) Reset to Al	
0	1	0	0	0	0	0	Х	Х	Х	Х	Mute Mode. Filter Active, Zero Gain, Reduced Noise
0	0	Х	Х	Х	Х	Х	X X X X X Mute Mode. Filter Active, Zero Gain, Reduced Noise. L (F and G) Reset to All Zeros		Mute Mode. Filter Active, Zero Gain, Reduced Noise. Latch Contents (F and G) Reset to All Zeros		
0	1	1	Ot	her Tl	han O	0000 X		Х	Х	Х	Normal Filtering Operation. Latch Holds F and G Inputs Present when Last $\overline{\text{CS}}$ /HOLD = 0
0	1	0	Ot	her T	han O	000	1		Х	Normal Filtering Operation. Filter Responds Directly to F and G Input Pins (See Separate Pin Descriptions)	

X = Doesn't Matter



Figure 1. Dual Supply Ground Plane Connection

AGND (Pin 15): Analog Ground. The AGND pin is at the midpoint of an internal resistive voltage divider, developing a potential halfway between the V⁺ and V⁻ pins, with an equivalent series resistance to the pin of nominally 7k. (In the shutdown state, analog switch FETs interrupt the voltage-divider resistors and the AGND pin assumes a high impedance.) AGND also serves as the internal half-supply reference in the LTC1564, tied to the noninverting inputs of all internal op amps and establishing the ground reference voltage for the IN and OUT pins. Because of this, very "clean" grounding is recommended, including an





Figure 2. Single Supply Ground Plane Connection

analog ground plane surrounding the package. For dual supply operation, this ground plane will be tied to the OV point and the AGND pin should connect directly to the ground plane (Figure 1). For single supply operation, in contrast, if the system signal ground is at V⁻, the ground plane should tie to V⁻ and the AGND pin should be AC-bypassed to the ground plane by at least a 0.1 μ F high quality capacitor (at least 1 μ F for best AC performance) (Figure 2). As with all high dynamic range analog circuits, performance in an application will reflect the quality of the grounding.

PIN FUNCTIONS

IN (Pin 16): Analog Input. The filter in the LTC1564 senses the voltage difference between the IN and AGND pins. In normal filtering (EN = 0, RST = 1, F code other than 0000), the IN pin connects within the LTC1564 to a digitally controlled resistance whose other end is a current-summing point at the AGND potential. At unity gain (G code 0000), the value of this input resistance is nominally 10k and the IN voltage range is rail-to-rail (V^+ to V^-). When filtering at gain settings above unity (G code \neq 0000), the input resistance falls as (1/gain) to nominally 625Ω at a gain of 16 (G code 1111) and the linear input range also falls in inverse proportion to gain. (The variable gain capability is designed to boost lower level input signals with good noise performance.) Input resistance does not vary significantly with the frequency-setting F code except in the mute state (F code 0000). In either the mute state (F code 0000 or $\overrightarrow{RST} = 0$) or the shutdown state (\overrightarrow{EN} = 1 or EN open circuited), analog switches disconnect the IN pin internally and this pin presents a very high input resistance. Circuitry driving the IN pin must be compatible with the LT1564's input resistance and with the variation of this resistance in the event that the LTC1564 is used in multiple modes. Signal sources with significant output resistance may introduce a gain error as the source's output resistance and the LTC1564's input resistance form a voltage divider. This is especially true at the higher gain or G code settings where the LTC1564's input resistance is lowest.

In single supply voltage applications with elevated gain settings (G code \neq 0000) it is important to keep in mind that the LTC1564's ground reference point is AGND, not V⁻. With increasing gains, the LTC1564's linear input voltage range is no longer rail-to-rail but converges toward AGND. Similarly the OUT pin swings positive or negative with respect to AGND. At unity gain (G code 0000), both IN and OUT voltages can swing from rail-to-rail.



BLOCK DIAGRAM

Figure 3. Block Diagram



APPLICATIONS INFORMATION

Functional Description

The LTC1564 is a self-contained, continuous time, variable gain, high order analog lowpass filter. The gain magnitude between IN and OUT pins is approximately constant for signal frequency components up to the cutoff frequency $f_{\rm C}$ and falls off rapidly for frequencies above $f_{\rm C}$. The pins IN, OUT and AGND (analog ground) are the sole analog signal connections on the LTC1564; the others are power supplies and digital control inputs to select f_{C} (and to select gain if desired). The f_C range is 10kHz to 150kHz in 10kHz steps. The form of the lowpass frequency response is an 8-pole elliptic type with two stopband notches (Figure 4). This response rolls off by approximately 100dB from $f_{\rm C}$ to 2.5 $f_{\rm C}$. The LTC1564 is laser trimmed for $f_{\rm C}$ accuracy, passband ripple, gain and offset. It delivers a combination of 100+dB stopband attenuation, 100+dB signal-to-noise ratio (SNR) and 100+kHz f_C.



Figure 4. General Shape of Frequency Response

Figure 3 is a block diagram showing analog signal path, digital control latch, and analog ground (AGND) circuitry. A proprietary active-RC architecture filters the analog signal. This architecture limits internal noise sources to near the fundamental "kT/C" bounds for a filter of this order and power consumption. The variable gain capability at the input is an integral part of the filter, and allows boosting of low level input signals with little increase in output referred noise. This permits the input noise floor to drop steadily with increasing gain, enhancing the SNR at lower signal levels. Such a property is difficult to achieve in practice by combining separate variable gain amplifier and filter circuits.

Digital Control

Logic levels for the LTC1564 digital inputs are nominally rail-to-rail CMOS. (Logic 1 is V⁺, logic 0 is V⁻ or alternatively 0V with ±5V supplies). The part is tested with 10% and 90% of full excursion on the inputs, thus ±1.08V at ±1.35V supplies, ±1.9V at ±2.375V and 0.5V and 4.5V at ±5V.

The f_C and gain settings are always controlled by the output of an on-chip CMOS latch. Inputs to this latch are the pins F3 through F0, G3 through G0, the latch-enable control CS/HOLD and the asynchronous reset input RST. A logic-0 input to \overline{CS} /HOLD makes the latch transparent so that the F and G input pins pass directly to the latch outputs and therefore control the filter directly. Raising CS/HOLD to logic 1 freezes the latch's output so that the F and G input pins have no effect. Logic 0 at the RST input at any time resets the latch outputs to all zeros. The all-zero state, in turn, imposes a mute mode with zero gain and low output noise if the filter is powered on (EN = 0). The all-zeros condition will persist until RST is returned to logic 1, nonzero F and G inputs are set up and the latch outputs are updated by CS/HOLD = 0. EN is a chip-enable input causing a shutdown state. Specific details on the digital controls appear in the Pin Functions section of this data sheet.

Floatable Digital Inputs

Every digital input of the LTC1564 includes a small current source (roughly $10\mu A$) to float the CMOS input to V⁺ or V⁻ potential if the pin is unconnected. Table 4 summarizes the open-circuit default levels.

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INPUT	FLOATING LOGIC LEVEL	EFFECT								
EN	1	Shutdown State								
CS/HOLD	0	F and G Pins Enabled								
RST	1	Latch Not Reset								
F3 F2 F1 F0	0010	f _C = 20kHz								
G3 G2 G1 G0	0 0 0 0	Unity Passband Gain								

Note particularly that the pull-up current source at the \overline{EN} pin forces the LTC1564 to the shutdown state if this pin is left open. Therefore the user *must* connect \overline{EN} deliberately to a logic-0 level (V⁻, or optionally 0V with ±5V supplies) for normal filter operation. The other digital inputs float to



APPLICATIONS INFORMATION

levels that program the part for enabled F and G pins (\overline{CS} /HOLD = 0), 20kHz f_C and unity passband gain. Therefore six connections (power pins, \overline{EN} to logic 0, AGND, IN and OUT) are enough to set up a working 20kHz lowpass filter, and additional pins can be connected as necessary to select different f_C or gain.

This feature of floatable logic inputs is intended for rapid prototyping and experimentation. Floating the logic inputs is not recommended for production designs because, depending on construction details, the high impedances of these inputs may permit unwanted interference coupling and consequent erroneous digital inputs to the LTC1564.

Also, it may be necessary to consider the effect of the pullup and pull-down current sources on the logic that drives the LTC1564. In particular, if the LTC1564 operates from \pm 5V but receives digital inputs from logic using 5V and 0V, CMOS logic levels will be compatible but the possibility exists of the LTC1564 pulling current out of the driving logic at those LTC1564 inputs that are capable of floating to logic 0. That is because the small current sources at these inputs return to V⁻, not to 0V. If the driving logic presents a high impedance or three-state output, the LTC1564's input current may pull this output below 0V, although the current is limited to about 10µA. The system designer should be aware of this possibility and ensure that any such current flow is compatible with the driving logic.

Mute State

The Mute mode keeps the filter powered as in normal filtering but "turns off" the signal path for minimal signal transmission (approximately -100dB) and reduced output noise. This feature may be useful for gating a signal source on and off, or for system calibration procedures. Note however that the DC output in the Mute state may shift by some millivolts compared to normal filtering because the internal signal path changes. Recovery from Mute, like other transient responses in a filter, proceeds at the time scale of the filter's pole-zero time constants and therefore is faster at the higher f_C settings (that is, at the higher F codes).

The LTC1564 enters the Mute state when the F bits at the latch output (Figure 3) become 0000. (It can be remem-

bered as a "zero-bandwidth" frequency setting.) This is achieved either by presenting a 0000 code to the F inputs and lowering the \overline{CS} /HOLD input to enable the latch, or alternatively at any time by lowering \overline{RST} , which immediately resets the latch contents to all zeroes. Such a reset also occurs normally at the application of power, unless \overline{CS} /HOLD is low and a nonzero pattern at the F inputs overrides the brief power-on reset. In the Mute state, the G gain-control inputs have no effect.

Output noise in Mute is largely thermal and wideband (unlike in normal filtering, where the filter's response affects the noise spectrum). Typical Mute-state output noise is $5.4\mu V_{RMS}$ in 200kHz measurement bandwidth and less than $3\mu V_{RMS}$ in 40kHz bandwidth. It has occasionally happened elsewhere in the electronics industry that someone would characterize a circuit or system by comparing its output level in normal operation to the noise level in a Mute state as though this were a normal signalto-noise ratio (SNR), which it is not, because this signal and noise exist only at different times. A scrupulous name for such a measure is SMR, signal-to-mute ratio. Accordingly in a 40kHz bandwidth, the LTC1564 can exhibit an SMR exceeding 120dB.

Construction and Instrumentation Cautions

Electrically clean construction is important in applications seeking the full dynamic range or high stopband rejection of the LTC1564. Short, direct wiring will minimize parasitic capacitance and inductance. High quality supply bypass capacitors of 0.1µF near the chip provide good decoupling from a clean, low inductance power source. But several inches of wire (i.e., a few microhenrys of inductance) from the power supplies, unless decoupled by substantial capacitance ($\geq 10\mu F$) near the chip, can cause a high-Q LC resonance in the hundreds of kHz in the chip's supplies or ground reference. This may impair stopband rejection and other specifications at those frequencies. In stringent filter applications we have often found that a compact, carefully laid out printed circuit board with good ground plane makes a difference in both stopband rejection and distortion. Finally, equipment to measure filter performance can itself introduce distortion or noise floors. Checking for these limits with a wire replacing the filter is a prudent routine procedure.



TYPICAL APPLICATIONS



2-Chip Flexible DSP Front End with Amplification, Antialias Filtering and A/D Conversion

16-Bit Output, Sampling Rate to 500ksps, Analog Bandwidth to 150kHz, Gain to 24dB. (For More Information, See *Linear Technology* Magazine, May 2001)



Boosting a 100mV_{RMS} Input Signal to Nearly Fill the Input Range of the LTC1608 ADC. Input Frequency of 40kHz, LTC1608 f_{SAMPLE} = 204.8ksps, LTC1564 is Set for f_C = 50kHz and Gain of 16 (F = 0101, G = 1111). Measured THD is 86dB, HD₂ = -88dB, SNR = 85dB with 100mV_{RMS} Input. Dynamic Range of Approximately 115dB



TYPICAL APPLICATIONS





Single Supply Differential Output Driver



PACKAGE DESCRIPTION



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LTC1560-1	1MHz/500kHz Continuous Time, Lowpass Elliptic Filter	f _{CUTTOFF} = 500kHz or 1MHz
LTC1562/LTC1562-2	Universal 8th Order Active RC Filters	$f_{CUTOFF(MAX)} = 150$ kHz (LTC1562), $f_{CUTOFF(MAX)} = 300$ kHz (LTC1562-2)
LTC1563-2/LTC1563-3	4th Order Active RC Lowpass Filters	$f_{CUTOFF(MAX)} = 256 kHz$
LTC1565-31	650kHz Continuous Time, Linear Phase Lowpass Filter	7th Order, Differential Inputs and Outputs
LTC1566-1	2.3MHz Continuous Time Lowpass Filter	7th Order, Differential Input and Outputs
LTC1569-6/LTC1569-7	Self Clocked, 10th Order Linear Phase Lowpass Filters	$ \begin{array}{l} f_{\text{CLK}}/f_{\text{CUTOFF}} = 64/1, \ f_{\text{CUTOFF}(\text{MAX})} = 75 \text{kHz} \ (\text{LTC1569-6}), \\ f_{\text{CLK}}/f_{\text{CUTOFF}} = 32/1, \ f_{\text{CUTOFF}(\text{MAX})} = 300 \text{kHz} \ (\text{LTC1569-7}) \end{array} $