

LMX9830 Bluetooth™ Serial Port Module

1.0 General Description

The National Semiconductor LMX9830 Bluetooth Serial Port module is a highly integrated Bluetooth 2.0 baseband controller and 2.4 GHz radio, combined to form a complete small form factor (6.1 mm x 9.1 mm x 1.2 mm) Bluetooth node.

All hardware and firmware is included to provide a complete solution from antenna through the complete lower and upper layers of the Bluetooth stack, up to the application including the Generic Access Profile (GAP), the Service Discovery Application Profile (SDAP), and the Serial Port Profile (SPP). The module includes a configurable service database to fulfil service requests for additional profiles on the host. Moreover, the LMX9830 is pre-qualified as a Bluetooth Integrated Component. Conformance testing through the Bluetooth qualification program enables a short time to market after system integration by insuring a high probability of compliance and interoperability.

Based on National's CompactRISC™ 16-bit processor architecture and Digital Smart Radio technology, the LMX9830 is optimized to handle the data and link management processing requirements of a Bluetooth node.

The firmware supplied in the on-chip ROM memory offers a complete Bluetooth (v2.0) stack including profiles and command interface. This firmware features point-to-point and point-to-multipoint link management supporting data rates up to the theoretical maximum over RFCOMM of 704 kbps (**Best in Class** in the industry). The internal memory supports up to 7 active Bluetooth data links and one active SCO link.

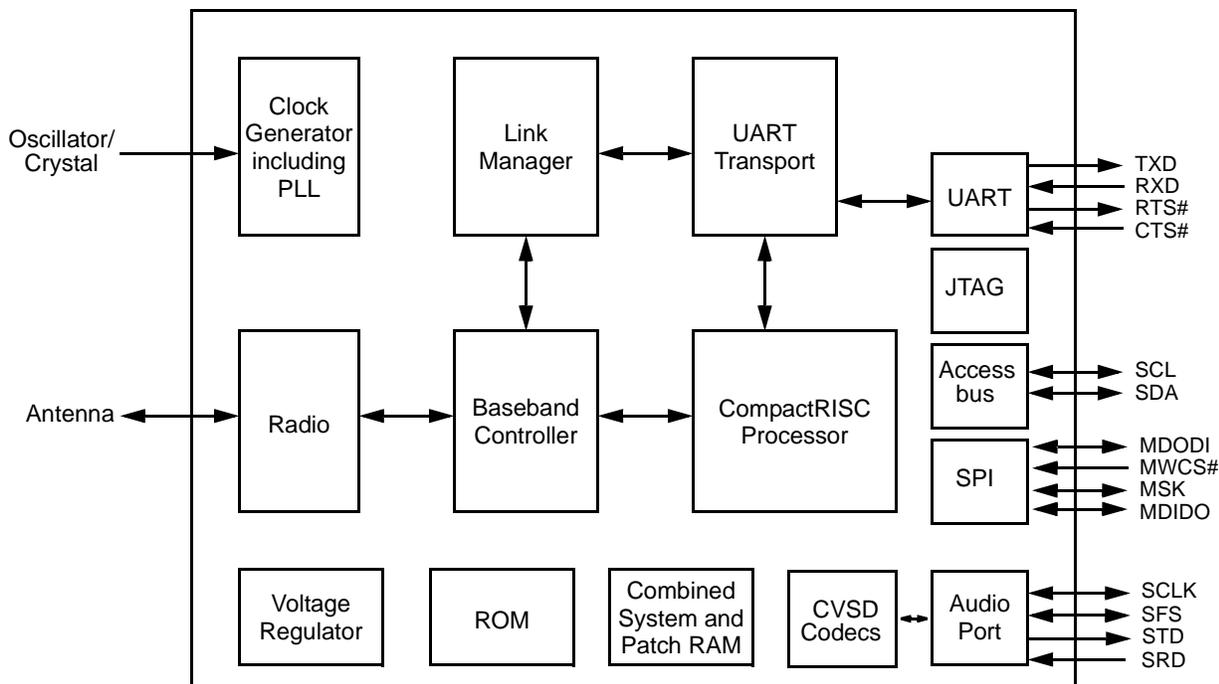
The on-chip Patch RAM provided for lowest cost and risk, allows the flexibility of firmware upgrade.

The LMX9830 module is lead free and RoHS (Restriction of Hazardous Substances) compliant. For more information on those quality standards, please visit our green compliance website at <http://www.national.com/quality/green/>

1.1 APPLICATIONS

- Personal Digital Assistants
- POS Terminals
- Data Logging Systems
- Audio Gateway applications
- Telemedicine/Medical, Industrial and Scientific.

2.0 Functional Block Diagram



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3.0 Features

- Compliant with the Bluetooth 2.0 Core Specification
- Better than -80 dBm input sensitivity
- Class 2 operation
- Low power consumption:
- High integration:
 - Implemented in 0.18 μm CMOS technology
 - RF includes antenna filter and switch on-chip

3.1 DIGITAL HARDWARE

- Baseband and Link Management processors
- CompactRISC Core
- Embedded ROM and Patch RAM memory
- UART Command/Data Port:
 - Support for up to 921.6k baud rate
- Auxiliary Host Interface Ports:
 - Link Status
 - Transceiver Status (Tx or Rx)
 - Three General Purpose I/Os, available through the API
 - Alternative IO functions:
 - Link Status
 - Transport layer activity
- Advanced Power Management (APM) features
 - Advanced power management functions
- Advanced Audio Interface for external PCM codec
- ACCESS.bus and SPI/Microwire for interfacing with external non-volatile memory

3.2 FIRMWARE

- Complete Bluetooth Stack including:
 - Baseband and Link Manager
 - L2CAP, RFCOMM, SDP
 - Profiles:
 - GAP
 - SDAP
 - SPP
- Additional Profile support on Host. e.g:
 - Dial Up Networking (DUN)
 - Facsimile Profile (FAX)
 - File Transfer Protocol (FTP)
 - Object Push Profile (OPP)
 - Synchronization Profile (SYNC)
 - Headset (HSP)
 - Handsfree Profile (HFP)
 - Basic Imaging Profile (BIP)
 - Basic Printing Profile (BPP)
- On-chip application including:
 - Default connections
 - Command Interface:
 - Link setup and configuration (also Multipoint)
 - Configuration of the module
 - Service database modifications
 - UART Transparent mode
 - Optimized cable replacement
 - Automatic transparent mode
 - Event filter

3.3 DIGITAL SMART RADIO

- Accepts external clock or crystal input:
 - 13 MHz Typical
 - Supports 10 - 20 MHz
 - Secondary 32.768 kHz oscillator for low-power modes
 - 20 ppm cumulative clock error required for Bluetooth
- Synthesizer:
 - Integrated VCO
 - Provides all clocking for radio and baseband functions
- Antenna Port (50 Ohms nominal impedance):
 - Embedded front-end filter for enhanced out of band performance
- Integrated transmit/receive switch (full duplex operation via antenna port)
- Better than -80 dBm input sensitivity
- 0 dBm typical output power

3.4 PHYSICAL

- Compact size - 6.1 mm x 9.1 mm x 1.2 mm
- Complete system interface provided in Ball Grid Array on underside for surface mount assembly

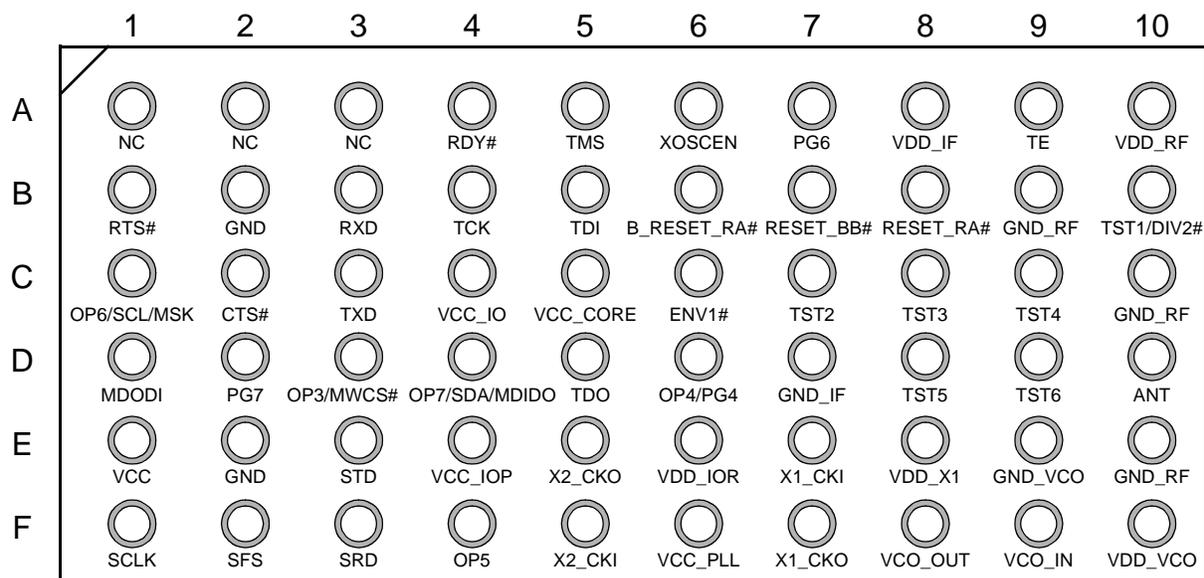
4.0 Order Information

Table 1. Order Information

| Order Number | Spec | Shipment Method |
|--------------|-------------------|----------------------|
| LMX9830SM | NOPB ¹ | 388 pcs Tray |
| LMX9830SMX | NOPB ¹ | 2500 pcs Tape & Reel |

1. NOPB = No Pb (No Lead)

5.0 Connection Diagram



X-ray - Top View

Figure 1. Connection Diagram

6.0 Pad Description

Table 2. Pin Description

| Pad Name | Pad Location | Type | Default Layout | Description |
|-------------|--------------|------|--------------------------|--|
| X1_CKO | F7 | O | | Crystal 10-20 MHz |
| X1_CKI | E7 | I | | Crystal or External Clock 10-20 MHz |
| X2_CKI | F5 | I | GND (if not used) | 32.768 kHz Crystal Oscillator. |
| X2_CKO | E5 | O | NC (if not used) | 32.768 kHz Crystal Oscillator. |
| RESET_RA# | B8 | I | | Radio Reset (active low) |
| B_RESET_RA# | B6 | O | NC | Buffered Reset Radio Output (active low) |
| RESET_BB# | B7 | I | | Baseband Reset (active low) |
| ENV1# | C6 | I | NC | ENV1: Environment Select (active low) used for manufacturing test only |
| TE | A9 | I | GND | Test Enable - Used for manufacturing test only |
| TST1/DIV2# | B10 | I | NC | TST1: Test Mode. Leave not connected to permit use with VTune automatic tuning algorithm DIV2#: No longer supported |
| TST2 | C7 | I | GND | Test Mode, Connect to GND |
| TST3 | C8 | I | GND | Test Mode, Connect to GND |
| TST4 | C9 | I | GND | Test Mode, Connect to GND |
| TST5 | D8 | I | GND | Test Mode, Connect to GND |
| TST6 | D9 | I | VCO_OUT | Test Input, Connect to VCO_OUT via 0Ohm resistor to permit use with VTune automatic tuning algorithm |

6.0 Pad Description (Continued)

Table 2. Pin Description

| Pad Name | Pad Location | Type | Default Layout | Description |
|--------------------|--------------|--|---|---|
| MDODJ ¹ | D1 | I/O | | SPI Master Out Slave In |
| OP6/SCL/MSK | C1 | OP6: I SCL/MSK : I/O | See Table 17 on page 17 | OP6: Pin checked during Startup Sequence for configuration option SCL: ACCESS.Bus Clock MSK: SPI Shift |
| OP7/SDA/MDIDO | D4 | OP7: I SDA/MDIDO: I/O | See Table 17 on page 17 | OP7: Pin checked during Startup Sequence for configuration option SDA: ACCESS.Bus Serial Data MDIDO: SPI Master In Slave Out |
| OP3/MWCS# | D3 | I | See Table 17 on page 17 and Table 18 on page 19 | OP3: Pin checked during Startup Sequence for configuration option MWCS#: SPI Slave Select Input (active low) |
| OP4/PG4 | D6 | OP4: I PG4: I/O | See Table 17 on page 17 and Table 18 on page 19 | OP4: Pin checked during Startup Sequence for configuration option PG4: GPIO |
| OP5 | F4 | I/O | See Table 17 on page 17 and Table 18 on page 19 | OP5: Pin checked during Startup Sequence for configuration option |
| SCLK | F1 | I/O | | Audio PCM Interface Clock |
| SFS | F2 | I/O | | Audio PCM Interface Frame Synchronization |
| SRD | F3 | I | | Audio PCM Interface Receive Data Input |
| STD | E3 | O | | Audio PCM Interface Transmit Data Output |
| XOSCEN | A6 | O | | Clock Request. Toggles with X2 (LP0) crystal enable/disable |
| PG6 | A7 | I/O | | GPIO |
| PG7 | D2 | I/O | | GPIO - Default setup RF traffic LED indication |
| CTS# ² | C2 | I | GND (if not used) | Host Serial Port Clear To Send (active low) |
| RXD | B3 | I | | Host Serial Port Receive Data |
| RTS# ³ | B1 | O | NC (if not used) | Host Serial Port Request To Send (active low) |
| TXD | C3 | O | | Host Serial Port Transmit Data |
| RDY# | A4 | O | NC | JTAG Ready Output (active low) |
| TCK | B4 | I | NC | JTAG Test Clock Input |
| TDI | B5 | I | NC | JTAG Test Data Input |
| TDO | D5 | O | NC | JTAG Test Data Output |
| TMS | A5 | I | NC | JTAG Test Mode Select Input |
| VCO_OUT | F8 | O | | Charge Pump Output, connect to Loop filter |
| VCO_IN | F9 | I | | VCO Tuning Input, feedback from Loop filter |
| ANT | D10 | I/O | | RF Antenna 50 ohm Nominal Impedance |
| VCC_PLL | F6 | O | | 1.8V Core Logic Power Supply Output |
| VCC_CORE | C5 | O | | 1.8V Voltage Regulator Output |

6.0 Pad Description (Continued)

Table 2. Pin Description

| Pad Name | Pad Location | Type | Default Layout | Description |
|----------|-----------------|------|----------------|---|
| VDD_X1 | E8 | I | | Power Supply Crystal Oscillator |
| VDD_VCO | F10 | I | | Power Supply VCO |
| VDD_RF | A10 | I | | Power Supply RF |
| VDD_IOR | E6 | I | | Power Supply I/O Radio/BB |
| VDD_IF | A8 | I | | Power Supply IF |
| VCC_IOP | E4 | I | | Power Supply Audio Interface |
| VCC_IO | C4 | I | | Power Supply I/O |
| VCC | E1 | I | | Voltage Regulator Input |
| GND_VCO | E9 | | | Ground |
| GND_RF | B9, C10, E10 | | | Ground |
| GND_IF | D7 | | | Ground |
| GND | B2,E2 | | | Ground |
| NC | A1,A2,A3 | | NC | Treat as no connect. Place pad for mechanical stability |

1. Must use 1k ohm pull up
2. Connect to GND if CTS is not use.
3. Treat as No Connect if RTS is not used. Pad required for mechanical stability.

7.0 General Specifications

Absolute Maximum Ratings (see Table 3) indicate limits beyond which damage to the device may occur. Operating Ratings (see Table 4) indicate conditions for which the device is intended to be functional.

This device is a high performance RF integrated circuit and is ESD sensitive. Handling and assembly of this device should be performed at ESD free workstations.

The following conditions are true unless otherwise stated in the tables below:

- $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$
- $V_{CC} = 3.3\text{V}$
- RF system performance specifications are guaranteed on National Semiconductor Mesa board rev 1.1 reference design platform.

Table 3. Absolute Maximum Ratings

| Symbol | Parameter | Min | Max | Unit |
|--------------------|---|------|------------------|------|
| VCC | Digital Voltage Regulator input | -0.2 | 4.0 | V |
| V _I | Voltage on any pad with GND = 0V | -0.2 | VCC + 0.2 | V |
| VDD_RF | Supply Voltage Radio | 0.2 | 3.3 | V |
| VDD_IF | | | | |
| VDD_X1 | | | | |
| VDD_VCO | | | | |
| P _{INRF} | | | | |
| V _{ANT} | Applied Voltage to ANT pad | | 1.95 | V |
| T _S | Storage Temperature Range | -65 | +150 | °C |
| T _L | Lead Temperature ¹ (solder 4 sec.) | | 225 | °C |
| T _{LNOPB} | Lead Temperature NOPB ^{1,2} (solder 40 sec.) | | 260 | °C |
| ESD _{HBM} | ESD - Human Body Model | | 2000 | V |
| ESD _{MM} | ESD - Machine Model | | 200 ³ | V |

1. Reference IPC/JDEC J-STD-20C spec.
2. NOPB = No Pb (No Lead)
3. A 200V ESD rating applies to all pins except OP3, OP6, OP7, MDODI, SCLK, SFS, STD, TDO, and ANT pins = 150V.

Table 4. Recommended Operating Conditions

| Symbol | Parameter | Min | Typ | Max | Unit |
|---------------------------|--|-----|------|------|------|
| VCC | Digital Voltage Regulator input | 2.5 | 2.75 | 3.6 | V |
| T _R | Digital Voltage Regulator Rise Time | | | 10 | us |
| T _A | Ambient Operating Temperature Range Fully Functional Bluetooth Node | -40 | +25 | +125 | °C |
| VCC_IO | Supply Voltage Digital I/O | 1.6 | 3.3 | 3.6 | V |
| VCC_PLL | Internally connected to VCC_Core | | | | |
| VDD_RF | Supply Voltage Radio | 2.5 | 2.75 | 3.0 | V |
| VDD_IF | | | | | |
| VDD_X1 | | | | | |
| VDD_VCO | | | | | |
| VDD_IOR | | | | | |
| VCC_IOP | Supply Voltage PCM Interface | 1.6 | 3.3 | 3.6 | V |
| VCC_CORE | Supply Voltage Output | | 1.8 | | V |
| VCC_CORE _{MAX} | Supply Voltage Output Max Load | | 5 | | mA |
| VCC_CORE _{SHORT} | When used as Supply Input (VCC grounded) | 1.6 | 1.8 | 2.0 | V |

7.0 General Specifications (Continued)

Table 5. Power Supply Requirements^{1,2}

| Symbol | Parameter | Min | Typ ³ | Max | Unit |
|-----------------------|--|-----|------------------|-----|------|
| I _{CC-TX} | Power supply current for continuous transmit | | | 65 | mA |
| I _{CC-RX} | Power supply current for continuous receive | | | 65 | mA |
| I _{RXSL} | Receive Data in SPP Link, Slave ⁴ | | 26 | | mA |
| I _{RXM} | Receive Data in SPP Link, Master ⁴ | | 23 | | mA |
| I _{SnM} | Sniff Mode, Sniff interval 1 second ⁴ | | 5.6 | | mA |
| I _{SC-TLDIS} | Scanning, No Active Link, TL Disabled ⁴ | | 0.43 | | mA |
| I _{Idle} | Idle, Scanning Disabled, TL Disabled ⁴ | | 100 | | μA |

1. Power supply requirements based on Class II output power.
2. Based on UART Baudrate 921.6kbit/s.
3. VCC = 3.3V, VCC_IO = 3.3V, Ambient Temperature = +25 °C.
4. Average values excluding IO

7.1 DC CHARACTERISTICS

Table 6. Digital DC Characteristics

| Symbol | Parameter | Condition | Min | Max | Units |
|------------------|---|--|---------------------|------------------------------|-------|
| V _{IH} | Logical 1 Input Voltage high (except oscillator I/O) | 1.6V ≤ VCC_IO ≤ 3.0V 3.0V ≤ VCC_IO ≤ 3.6V | 0.7 x VCC_IO 2.0 | VCC_IO + 0.2 VCC_IO + 0.2 | V |
| V _{IL} | Logical 0 Input Voltage low (except oscillator I/O) | 1.6V ≤ VCC_IO ≤ 3.0V 3.0V ≤ VCC_IO ≤ 3.6V | -0.2 -0.2 | 0.25 x VCC_IO 0.8 | V |
| V _{HYS} | Hysteresis Loop Width ¹ | | 0.1 x VCC_IO | | V |
| I _{OH} | Logical 1 Output Current | V _{OH} = 2.4V, VCC_IO = 3.0V | -10 | | mA |
| I _{OL} | Logical 0 Output Current | V _{OH} = 0.4V, VCC_IO = 3.0V | 10 | | mA |

1. Guaranteed by design.

7.0 General Specifications (Continued)

7.2 RF PERFORMANCE CHARACTERISTICS

In the performance characteristics tables the following applies:

- All tests performed are based on Bluetooth Test Specification revision 2.0.
- All tests are measured at antenna port unless otherwise specified

■ $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$

■ $V_{DD_RF} = 2.8\text{V}$ unless otherwise specified

RF system performance specifications are guaranteed on National Semiconductor Mesa Board rev 1.1 reference design platform.

Table 7. Receiver Performance Characteristics

| Symbol | Parameter | Condition | Min | Typ ¹ | Max | Unit |
|--------------------------------|---------------------------------------|--|-----------|------------------|-----|------|
| RX _{sense} | Receive Sensitivity | BER < 0.001 | 2.402 GHz | -80 | -76 | dBm |
| | | | 2.441 GHz | -80 | -76 | dBm |
| | | | 2.480 GHz | -80 | -76 | dBm |
| PinRF | Maximum Input Level | | -10 | 0 | | dBm |
| IMP ^{2,3} | Intermodulation Performance | F ₁ = + 3 MHz, F ₂ = + 6 MHz, P _{in} RF = -64 dBm | -38 | -36 | | dBm |
| RSSI | RSSI Dynamic Range at LNA Input | | -72 | | -52 | dBm |
| Z _{RFIN} ³ | Input Impedance of RF Port (RF_inout) | Single input impedance F _{in} = 2.5 GHz | | 32 | | Ω |
| Return Loss ³ | Return Loss | | | | -8 | dB |
| OOB ^{2,3} | Out Of Band Blocking Performance | P _{in} RF = -10 dBm, 30 MHz < F _{CWI} < 2 GHz, BER < 0.001 | -10 | | | dBm |
| | | P _{in} RF = -27 dBm, 2000 MHz < F _{CWI} < 2399 MHz, BER < 0.001 | -27 | | | dBm |
| | | P _{in} RF = -27 dBm, 2498 MHz < F _{CWI} < 3000 MHz, BER < 0.001 | -27 | | | dBm |
| | | P _{in} RF = -10 dBm, 3000 MHz < F _{CWI} < 12.75 GHz, BER < 0.001 | -10 | | | dBm |

1. Typical operating conditions are at 2.75V operating voltage and 25°C ambient temperature.
2. The f₀ = -64 dBm Bluetooth modulated signal, f₁ = -39dBm sine wave, f₂ = -39 dBm Bluetooth modulated signal, f₀ = 2f₁ - f₂, and |f₂ - f₁| = n * 1MHz, where n is 3, 4, or 5. For the typical case, n = 3.
3. Not tested in production

7.0 General Specifications (Continued)

Table 8. Transmitter Performance Characteristics

| Symbol | Parameter | Condition | Min | Typ ¹ | Max | Unit |
|---|---|--|-----|------------------|------|------|
| P _{OUTRF} | Transmit Output Power | 2.402 GHz | -4 | 0 | +3 | dBm |
| | | 2.441 GHz | -4 | 0 | +3 | dBm |
| | | 2.480 GHz | -4 | 0 | +3 | dBm |
| MOD ΔF _{1AVG} | Modulation Characteristics | Data = 00001111 | 140 | 165 | 175 | kHz |
| MOD ΔF _{2MAX} ² | Modulation Characteristics | Data = 10101010 | 115 | 125 | | kHz |
| ΔF _{2AVG} /ΔF _{1AVG} ³ | Modulation Characteristics | | 0.8 | | | |
| 20 dB Bandwidth | | | | | 1000 | kHz |
| P _{OUT2*f₀} ⁴ | PA 2 nd Harmonic Suppression | Maximum gain setting: f ₀ = 2402 MHz, P _{out} = 4804 MHz | | | -30 | dBm |
| Z _{RFOUT} ⁵ | RF Output Impedance/Input Impedance of RF Port (RF_inout) | P _{out} @ 2.5 GHz | | 47 | | Ω |

1. Typical operating conditions are at 2.75V operating voltage and 25°C ambient temperature.
2. ΔF_{2max} ≥ 115 kHz for at least 99.9% of all Δf_{2max}.
3. Modulation index set between 0.28 and 0.35.
4. Out-of-Band spurs only exist at 2nd and 3rd harmonics of the CW frequency for each channel.
5. Not tested in production.

Table 9. Synthesizer Performance Characteristics

| Symbol | Parameter | Condition | Min | Typ | Max | Unit |
|------------------------------------|-------------------------------------|-------------------------|------|-----|------|----------|
| f _{VCO} | VCO Frequency Range | | 2402 | | 2480 | MHz |
| t _{LOCK} | Lock Time | f ₀ ± 20 kHz | | 120 | | μs |
| Δf _{0offset} ¹ | Initial Carrier Frequency Tolerance | During preamble | -75 | 0 | 75 | kHz |
| Δf _{0drift} ¹ | Initial Carrier Frequency Drift | DH1 data packet | -25 | 0 | 25 | kHz |
| | | DH3 data packet | -40 | 0 | 40 | kHz |
| | | DH5 data packet | -40 | 0 | 40 | kHz |
| | | Drift Rate | -20 | 0 | 20 | kHz/50μs |
| t _{D-Tx} | Transmitter Delay Time | From Tx data to antenna | | 4 | | μs |

1. Frequency accuracy is dependent on crystal oscillator chosen. The crystal must have a cumulative accuracy of < +/- 20ppm to meet Bluetooth specifications.

7.0 General Specifications (Continued)

7.3 PERFORMANCE DATA (TYPICAL)

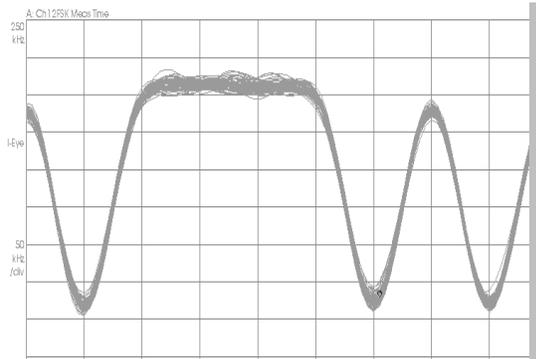


Figure 2. Modulation

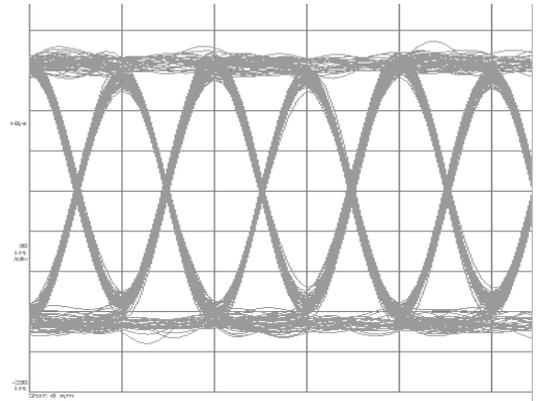


Figure 4. Corresponding Eye Diagram

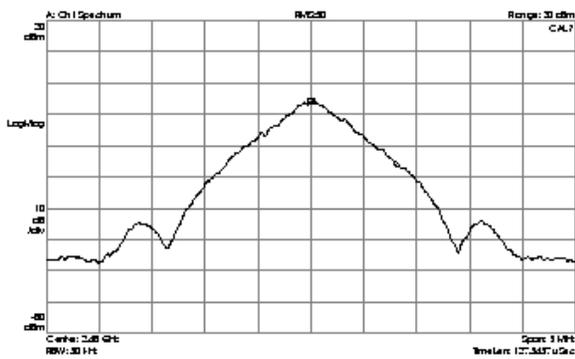


Figure 3. Transmit Spectrum

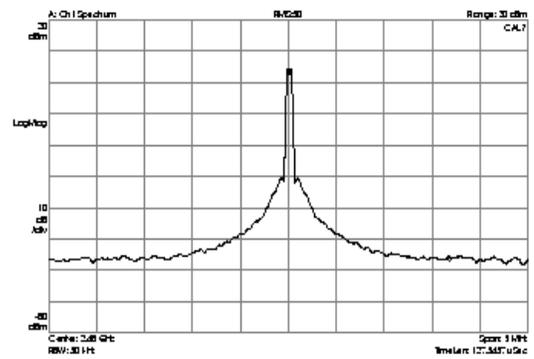


Figure 5. Synthesizer Phase Noise

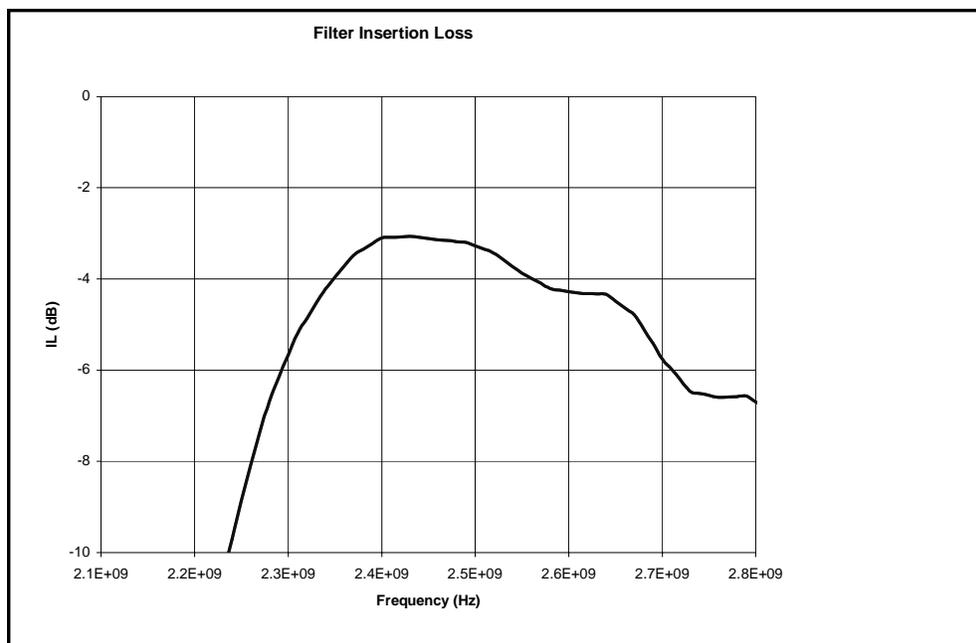


Figure 6. Front-End Bandpass Filter Response

7.0 General Specifications (Continued)

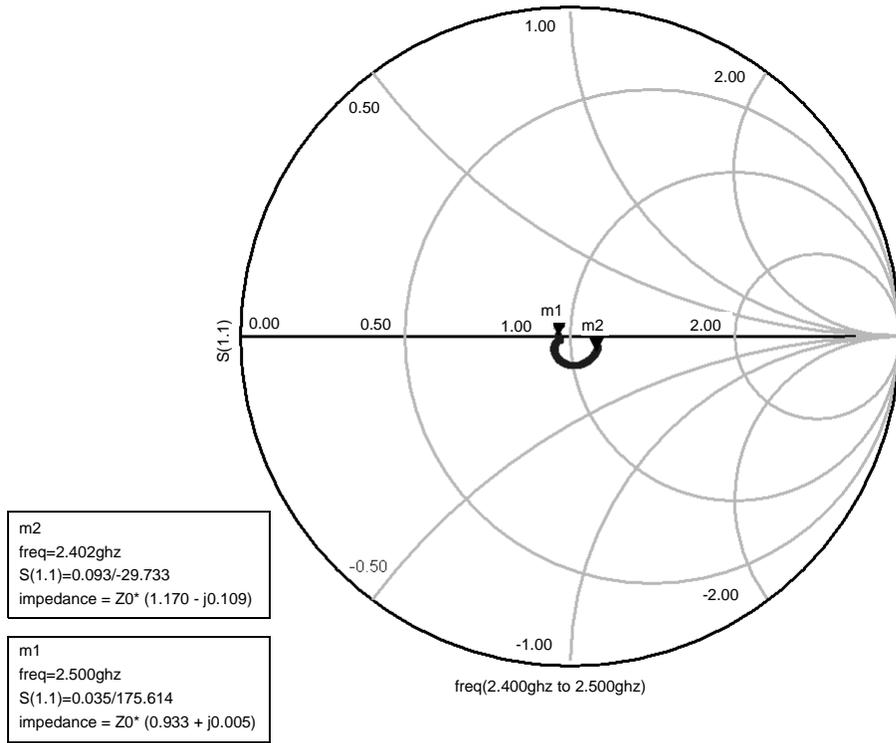


Figure 7. TX and RX Pin 50Ω Impedance Characteristics

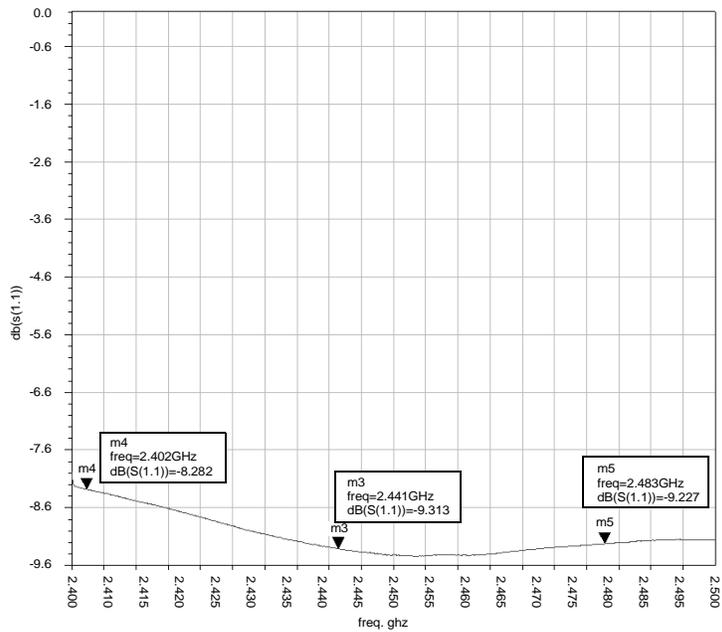


Figure 8. Transceiver Return Loss

8.0 Functional Description

8.1 BASEBAND AND LINK MANAGEMENT PROCESSORS

Baseband and Lower Link control functions are implemented using a combination of National Semiconductor's CompactRISC 16-bit processor and the Bluetooth Lower Link Controller. These processors operate from integrated ROM memory and RAM and execute on-board firmware implementing all Bluetooth functions.

8.1.1 Bluetooth Lower Link Controller

The integrated Bluetooth Lower Link Controller (LLC) complies with the Bluetooth Specification version 2.0 and implements the following functions:

- Adaptive Frequency Hopping
- Interlaced Scanning
- Fast Connect
- Support for 1, 3, and 5 slot packet types
- 79 Channel hop frequency generation circuitry
- Fast frequency hopping at 1600 hops per second
- Power management control
- Access code correlation and slot timing recovery

8.1.2 Bluetooth Upper Layer Stack

The integrated upper layer stack is prequalified and includes the following protocol layers:

- L2CAP
- RFComm
- SDP

8.1.3 Profile support

The on-chip application of the LMX9830 allows full stand-alone operation, without any Bluetooth protocol layer necessary outside the module. It supports the Generic Access Profile (GAP), the Service Discovery Application Profile (SDAP), and the Serial Port Profile (SPP).

The on-chip profiles can be used as interfaces to additional profiles executed on the host. The LMX9830 includes a configurable service database to answer requests with the profiles supported.

8.1.4 Application with command interface

The module supports automatic slave operation eliminating the need for an external control unit. The implemented transparent option enables the chip to handle incoming data raw, without the need for packaging in a special format. The device uses a pin to block unallowed connections. This pincode can be fixed or dynamically set.

Acting as master, the application offers a simple but versatile command interface for standard Bluetooth operation like inquiry, service discovery, or serial port connection. The firmware supports up to seven slaves. Default Link Policy settings and a specific master mode allow optimized configuration for the application specific requirements. See also Section "Integrated Firmware" on page 30.

8.1.5 Memory

The LMX9830 introduces 16 kB of combined system and Patch RAM memory that can be used for data and/or code upgrades of the ROM based firmware. Due to the flexible startup used for the LMX9830 operating parameters like the Bluetooth Device Address (BD_ADDR) are defined during boot time. This allows reading out the parameters of an external EEPROM or programming them directly over UART.

8.1.6 External memory interfaces

As the LMX9830 is a ROM based device with no on-chip non volatile storage, the operation parameters will be lost after a power cycle or hardware reset. In order to prevent re initializing such parameters, patches or even user data, the LMX9830 offers two interfaces to connect an external EEPROM to the device:

- μ -wire/SPI
- Access.bus (I²C compatible)

The selection of the interface is done during start up based on the option pins. See Table 17 on page 17 for the option pin descriptions.

8.1.7 μ -wire/SPI interface

In case the firmware is configured by the option pins to use a μ -wire/SPI EEPROM, the LMX9830 will activate that interface and try to read out data from the EEPROM. The external memory needs to be compatible to the reference listed in Table 10 on page 13. The largest size EEPROM supported is limited by the addressing format of the selected NVM.

The device must have a page size equal to N x 32 bytes.

The firmware requires that the EEPROM supports Page write. Clock must be HIGH when idle.

Table 10. M95640-S EEPROM 8Kx8

| Parameter | Value |
|-----------------------------|---|
| Supplier | ST Microelectronics |
| Supply Voltage ¹ | 1.8 - 3.6V |
| Interface | SPI compatible (positive clock SPI Modes) |
| Memory Size | 8K x 8, 64kbit |
| Clock Rate ¹ | 2 MHz |
| Access | Byte and Page Write (up to 32bytes) |

1. Parameter range reduced to requirements of National reference design

8.0 Functional Description (Continued)

8.1.8 Access.bus interface

In case the firmware is configured by the option pins to use an access.bus or i²c compatible EEPROM, the LMX9830 will activate that interface and try to read out data from the EEPROM. The external memory needs to be compatible to the reference listed in Table 11 on page 14.

The largest size EEPROM supported is limited by the addressing format of the selected NVM. The device must have a page size equal to N x 32 bytes.

The device uses a 16 bit address format. The device address must be "000".

Table 11. 24C64 EEPROM 8Kx8

| Parameter | Value |
|-----------------------------|-------------------------|
| Supplier | Atmel |
| Supply Voltage ¹ | 2.7 - 5.5 V |
| Interface | 2 wire serial interface |
| Memory Size | 8K x 8, 64kbit |
| Clock Rate ¹ | 100 KHz |
| Access | 32 Byte Page Write Mode |

1. Parameter range reduced to requirements of National reference design

8.2 TRANSPORT PORT - UART

The LMX9830 provides one Universal Asynchronous Receiver Transmitter (UART). The UART interface consists out of Receive (RX), Transmit (TX), Ready-to-Send (RTS) and Clear-to-Send signals. RTS and CTS are used for hardware handshaking between the host and the LMX9830. Since the LMX9830 acts as gateway between the bluetooth and the UART interface, National Semiconductor recommends to use the handshaking signals especially for transparent operation. In case two signals are used CTS needs to be pulled to GND. Please refer also to "LMX9830 Software User's Guide" for detailed information on 2-wire operation.

The UART interface supports formats of 8-bit data with or without parity, with one or two stop bits. It can operate at standard baud rates from 2400bits/s up to a maximum baud rate of 921.6kbits/s. DMA transfers are supported to allow for fast processor independent receive and transmit operation.

The UART baudrate is configured during startup by checking option pins OP3, OP4 and OP5 for reference clock and baudrate. In case Auto baud rate detect is chosen, the firmware check the NVS area if a valid UART baudrate has been stored in a previous session. In case, no useful value can be found the device will switch to auto baud rate detection and wait for an incoming reference signal.

The UART offers wakeup from the power save modes via the multi-input wakeup module. When the LMX9830 is in low power mode, RTS# and CTS# can function as Host_WakeUp and Bluetooth_WakeUp respectively. Table 12 on page 14 represents the operational modes supported by the firmware for implementing the transport via the UART.

Table 12. UART Operation Modes

| Item | Range | Default at Power-Up | With Auto-Detect |
|--------------|----------------------|---|----------------------|
| Baud Rate | 2.4 to 921.6 kbits/s | Either configured by option pins, NVS parameter or auto baud rate detection | 2.4 to 921.6 kbits/s |
| Flow Control | RTS#/CTS# or None | RTS#/CTS# | RTS#/CTS# |
| Parity | Odd, Even, None | None | None |
| Stop Bits | 1,2 | 1 | 1 |
| Data Bits | 8 | 8 | 8 |

8.3 AUDIO PORT

8.3.1 Advanced Audio Interface

The Advanced Audio Interface (AAI) is an advanced version of the Synchronous Serial Interface (SSI) that provides a full-duplex communications port to a variety of industry-standard 13/14/15/16-bit linear or 8-bit log PCM codecs, DSPs, and other serial audio devices.

The interface allows the support one codec or interface. The firmware selects the desired audio path and interface configuration by a parameter that is located in RAM (imported from non-volatile storage or programmed during boot-up). The audio path options include the Motorola

MC145483 codec, the OKI MSM7717 codec, the Winbond W681360/W681310 codecs and the PCM slave through the AAI.

In case an external codec or DSP is used the LMX9830 audio interface generates the necessary bit and frame clock driving the interface.

Table 13 on page 15 summarizes the audio path selection and the configuration of the audio interface at the specific modes.

The LMX9830 supports one SCO link.

8.0 Functional Description (Continued)

Table 13. Audio path configuration

| Audio setting | Interface | Freq | Format | AAI Bit Clock | AAI Frame Clock | AAI Frame Sync Pulse Length |
|--------------------------------------|--------------------------|------------------|-------------------------------|----------------|-----------------|-----------------------------|
| OKI MSM7717 | Advanced audio interface | ANY ¹ | 8-bit log PCM (a-law only) | 480 KHz | 8 KHz | 14 Bits |
| Motorola MC145483² | Advanced audio interface | | 13-bit linear | 480 KHz | 8 KHz | 13 Bits |
| OKI MSM7717 | Advanced audio interface | 13MHz | 8-bit log PCM (a-law only) | 520 KHz | 8 KHz | 14 Bits |
| Motorola MC145483³ | Advanced audio interface | | 13-bit linear | 520 KHz | 8 KHz | 13 Bits |
| Winbond W681310 | Advanced audio interface | 13MHz | 8 bit log PCM A-law and u-law | 520 KHz | 8 KHz | 14 Bits |
| Winbond W681360 | Advanced audio interface | 13MHz | 13-bit linear | 520 KHz | 8 KHz | 13 Bits |
| PCM slave⁴ | Advanced audio interface | ANY ¹ | 8/16 bits | 128 - 1024 KHz | 8 KHz | 8/16 Bits |

- For supported frequencies see Table 22 on page 24
- Due to internal clock divider limitations the optimum of 512KHz, 8KHz can not be reached. The values are set to the best possible values. The clock mismatch does not result in any discernible loss in audio quality.
- Due to internal clock divider limitations the optimum of 512KHz, 8KHz can not be reached. The values are set to the best possible values. The clock mismatch does not result in any discernible loss in audio quality.
- In PCM slave mode, parameters are stored in NVS. Bit clock and frame clock must be generated by the host interface.

PCM slave configuration example: PCM slave uses the slot 0, 1 slot per frame, 16 bit linear mode, long frame sync, normal frame sync. In this case, 0x03E0 should be stored in NVS. See “LMX9830 Software User’s Guide” for more details.

8.4 AUXILIARY PORTS

8.4.1 RESET#

There are two reset inputs: RESET_RA# for the radio and RESET_BB# for the baseband. Both are active low.

There is also a reset output, B_RESET_RA# (Buffered Radio Reset) active low. This output follows input RESET_RA#.

When RESET_RA# is released, going high, B_RESET_RA# stays low until the clock has started.

Please see Section 8.5 "System Power Up" on page 15 for details.

8.4.2 General Purpose I/Os

The LMX9830 offers 3 pins which either can be used as indication and configuration pins or can be used for General Purpose functionality. The selection is made out of settings derived out of the power up sequence.

In General Purpose configuration the pins are controlled hardware specific commands giving the ability to set the direction, set them to high or low or enable a weak pull-up.

In alternate function the pins have pre-defined indication functionality. Please see Table 14 on page 15 for a description on the alternate indication functionality.

Table 14. Alternate GPIO pin configuration

| Pin | Description |
|---------|---|
| OP4/PG4 | Operation Mode pin to configure Transport Layer settings during boot-up |
| PG6 | GPIO |
| PG7 | RF Traffic indication |

8.5 SYSTEM POWER UP

In order to correctly power-up the LMX9830 the following sequence is recommended to be performed:

Apply VCC_IO and VCC to the LMX9830.

The RESET_RA# should be driven high. Then RESET_BB# should be driven high at a recommended time of 1ms after the LMX9830 voltage rails are high. The LMX9830 is properly reset.

Please see timing diagram, Figure 9 on page 16.

8.0 Functional Description (Continued)

ESR of the crystal also has impact on the startup time of the crystal oscillator circuit of the LMX9830 (See Table 15 and Table 16 on page 16).

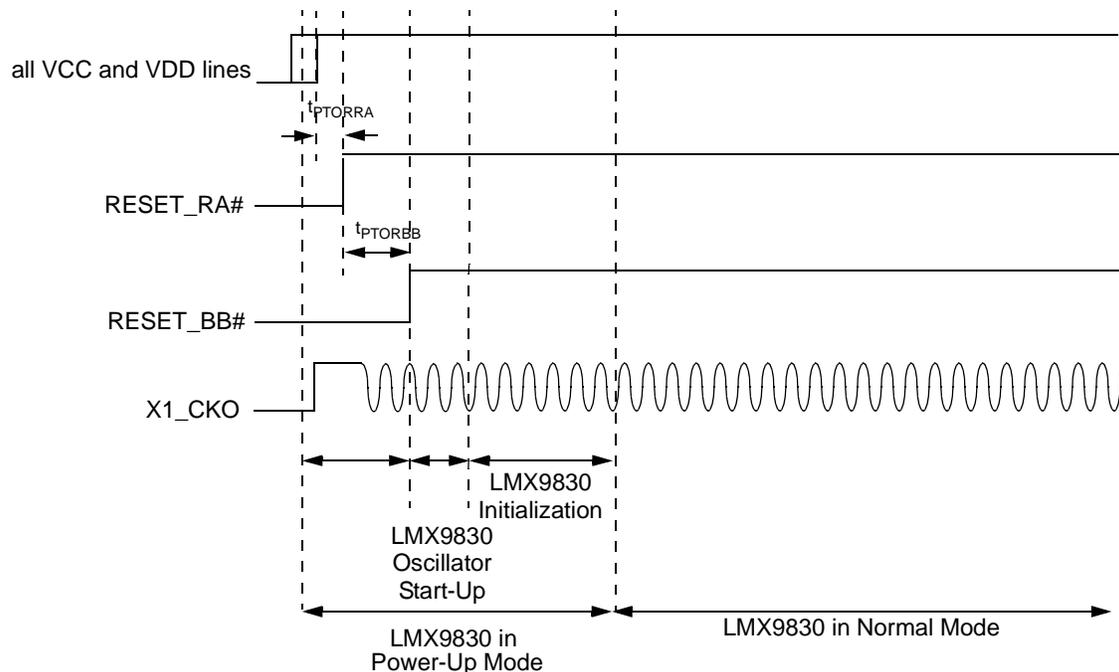


Figure 9. LMX9830 Power on Reset Timing

Table 15. LMX9830 Power to Reset timing

| Symbol | Parameter | Condition | Min | Typ | Max | Unit |
|--------------|------------------------|---|-------------------|-----|-----|------|
| t_{PTORRA} | Power to Reset_RA# | V_{CC} and V_{CC_IO} at operating voltage level to valid reset | <500 ¹ | | | us |
| t_{PTORBB} | Reset_RA# to Reset_BB# | V_{CC} and V_{CC_IO} at operating voltage level to valid reset | 1 ² | | | ms |

1. Rise time on power must switch on fast, rise time <500us.
2. Recommended value.

Table 16. ESR vs. Startup Time

| ESR (Ohm) | Typical ^{1,2} | Unit |
|-----------|------------------------|------|
| 10 | 12 | ms |
| 25 | 13 | ms |
| 40 | 16 | ms |
| 50 | 24 | ms |
| 80 | 30 | ms |

1. Frequency, loading caps and ESR all must be considered for determining startup time.
2. For reference only, must be tested on each system to accurately design POR and correctly startup system.

8.0 Functional Description (Continued)

8.6 STARTUP SEQUENCE

During startup the LMX9830 checks the options register pins OP3 to OP7 for configuration on operation mode, external clock source, transport layer and available non volatile storage PROM.

The different options for startup are described in Table 17 on page 17.

8.6.1 Options Register

External pads in Table 17 on page 17 are latched in this register at the end of Reset. The Options register can be read by firmware at any time.

All pads are inputs with weak on-chip pull-up/down resistors during Reset. Resistors are disconnected at the end of RESET_BB#.

1 = Pull-up resistor connected in application

0 = Pull-down resistor connected in application

x = Don't care

8.6.2 Startup With External PROM Available

To be able to read out information from an external PROM the option pins have to be set according to Table 17 on page 17.

Startup sequence activities:

1. From the Options registers OP6 and OP7, the LMX9830 checks if a serial PROM is available to use (ACCESS.bus or Microwire).
2. If serial PROM is available, the permanent parameter block, patch block, and non-volatile storage (NVS) are

read from it. If the BD Address is not present, enter the BD address to be saved in the NVS. For more information see Section 8.6.4 "Configuring the LMX9830 through transport layer" on page 19

3. From the Options register OP3, OP4 and OP5, the LMX9830 checks for clocking information and transport layer settings. If the NVS information are not sufficient, the LMX9830 will send the "Await Initialization" event on the TL (Transport Layer) and wait for additional information (see Section 8.6.3 "Startup Without External PROM Available" on page 17.)
4. The LMX9830 compensates the UART for new BBCLK information from the NVS.
5. The LMX9830 starts up the Bluetooth core.

8.6.3 Startup Without External PROM Available

The following sequence will take place if OP6 and OP7 have been set to "No external memory" as described in Table 17 on page 17.

Startup sequence activities:

1. From the Options registers OP6 and OP7, the LMX9830 checks if a serial PROM is available to use.
2. From the Options register OP3, OP4 and OP5, the LMX9830 checks for clocking mode and transport layer.
3. The LMX9830 sends the "Await Initialization" Event on the TL (Transport Layer) and waits for NVS configuration commands. The configuration is finalized by sending the "Enter Bluetooth Mode" command.
4. The LMX9830 compensates the UART for new BBCLK information from the NVS.
5. The LMX9830 starts up the Bluetooth core.

Table 17. Startup Sequence Options¹

| Package Pad | | | | | | Comment |
|-------------|-----|-----|------------------|------------------|-------------------|--|
| OP3 | OP4 | OP5 | OP6 ² | OP7 ³ | ENV1# | |
| PD | PD | PD | PD | PD | PU | PD = Internal Pull-down during Reset PU = Internal Pull-up during Reset |
| x | x | x | Open (0) | Open (0) | Open (1) BBCLK | No serial memory |
| x | x | x | 1 | Open (0) | Open (1) BBCLK | TBD |
| x | x | x | Open (0) | 1 | Open (1) BBCLK | Microwire serial memory |
| x | x | x | 1 | 1 | Open (1) BBCLK | ACCESS.bus serial memory |
| T_SCLK | x | x | T_RFDAT A | T_RFCE | 0 BBCLK | Test mode |

1. 1/0 pull-up/down resistor connected in application.
2. If OP6 is 1, must use 1k ohm pull up
3. If OP7 is 1, must use 1k ohm pull up

8.0 Functional Description (Continued)

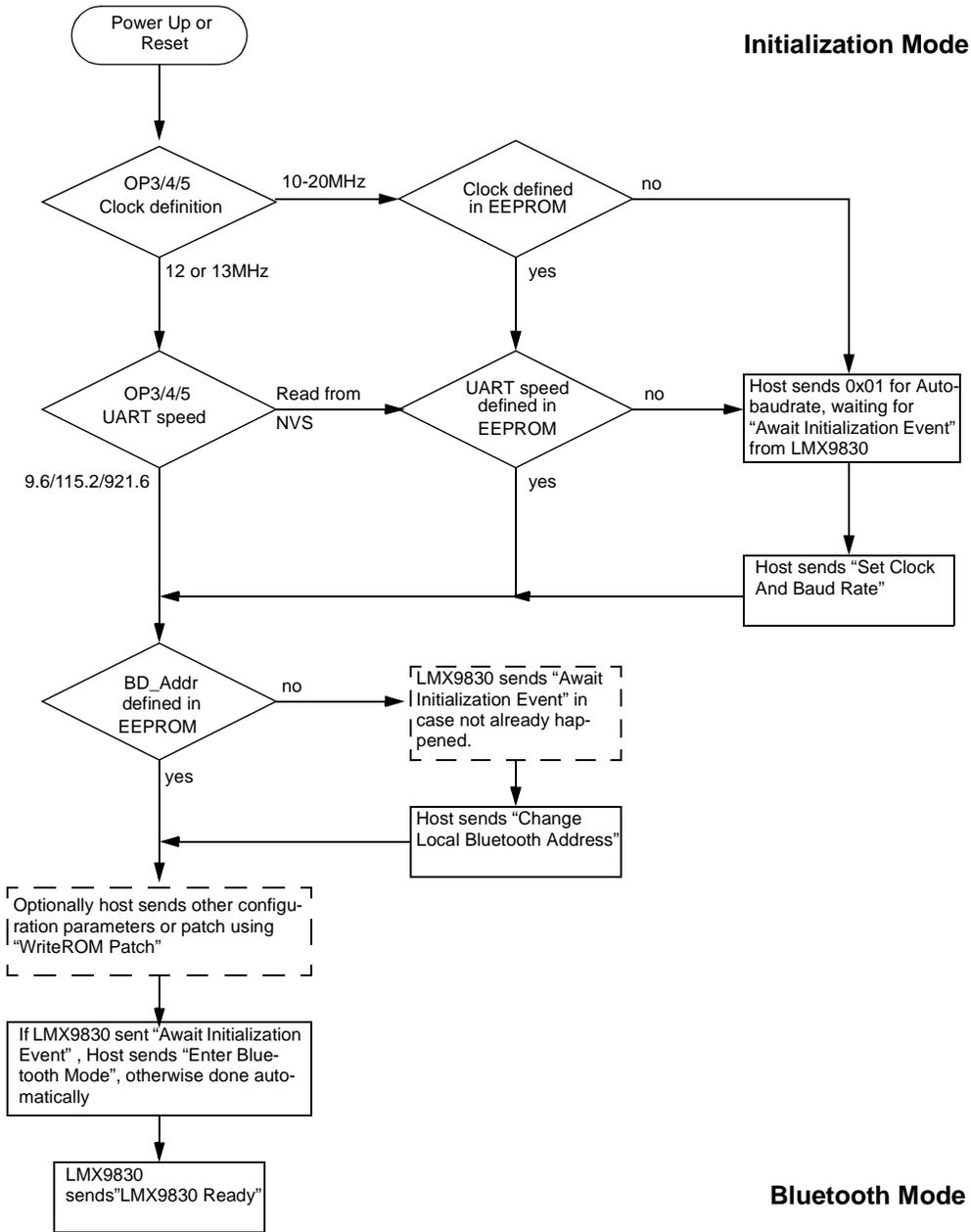


Figure 10. Flow Diagram for the Start-up Sequence

8.0 Functional Description (Continued)

Table 18. Fixed Frequencies

| Osc Freq. (MHz) | BBCLK (MHz) | PLL (48 MHz) | OP3 ¹ | OP4 ² | OP5 ³ | Function |
|--------------------|--------------------|--------------|------------------|------------------|------------------|-----------------------------------|
| 12 | 12 | OFF | 0 | 0 | 0 | UART speed read from NVS |
| 10-20 ⁴ | 10-20 ¹ | ON | 0 | 1 | 0 | Clock and UART baudrate detection |
| 13 | 13 | OFF | 1 | 0 | 0 | UART speed read from NVS |
| 13 | 13 | OFF | 1 | 0 | 1 | UART speed 9.6 kbps |
| 13 | 13 | OFF | 1 | 1 | 0 | UART speed 115.2 kbps |
| 13 | 13 | OFF | 1 | 1 | 1 | UART speed 921.6 kbps |

1. If OP3 is 1, must use 1k ohm pull up.
2. If OP4 is 1, must use 1k ohm pull up.
3. If OP5 is 1, must use 1k ohm pull up.
4. Supported frequencies see Table 22 on page 24

8.6.4 Configuring the LMX9830 through transport layer

As described in Section 8.5 "System Power Up" on page 15, the LMX9830 will check during startup the Options Registers if an external PROM is available. If the information on the PROM are incomplete or no PROM is installed the LMX9830 will boot into the "initialization Mode".

The mode is confirmed by the "Await Initialization" Event.

The following information are needed to enter Bluetooth Mode:

- Bluetooth Device Address (BD_Addr)
- External clock source (only if 10 - 20 MHz has been selected)
- UART Baudrate (only if Auto baudrate detection has been selected)

In general the following procedure will initialize the LMX9830:

1. Wait for "Await initialization" Event
 - Event will only appear if transport layer speed is set or after successful baudrate detection.
2. Send "Set Clock and Baudrate" Command only if the clock speed is not known through hardware configuration (i.e only if **OP3 OP4 OP5 = 0 1 0**).
3. Send "Write BD_Addr" to Configure Local Bluetooth Device Address.
4. Send "Enter Bluetooth Mode"
 - LMX9830 will use configured clock and UART speed and start the command interface.

NOTE: In case no EEPROM is used, BDAAddr, clock source and Baudrate are only valid until the next power-cycle or hardware reset.

8.6.5 Auto Baud Rate Detection

The LMX9830 supports an Automatic Baudrate Detection in case the external clock is different to 12, 13MHz or the range 10-20 MHz or the baudrate is different to 9.6 kbps, 115.2 or 921.6kbit/s.

The baudrate detection is based on the measurement of a single character. The following issues need to be considered:

- The flow control pin CTS must be low or else the host is in flow stop.
- The Auto Baudrate Detector measures the length of the 0x01 character from the positive edge of bit 0 to the positive edge of stop bit.
- Therefore the very first received character must always be a **0x01**.
- The host can restrict itself to send only a 0x01 character or also can send a command.
- The host must flush the TX buffer within 50-100 milliseconds depend on clock frequency on the host controller.
- After 50-100 milliseconds the UART is about to be initialized and short after the host should receive a "Await Initialization" Event or an "Command Status" Event.

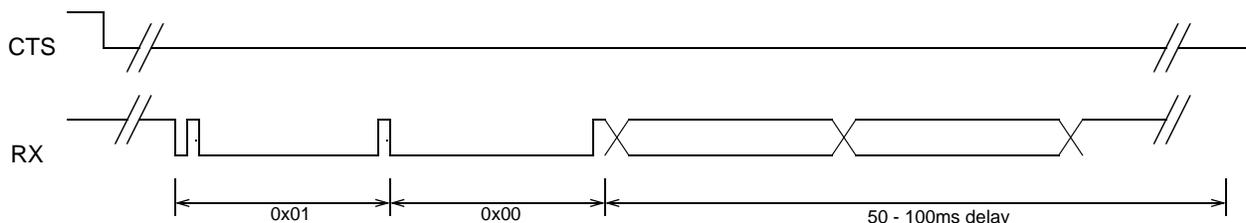


Figure 11. Auto baudrate detection timing diagram

9.0 Digital Smart Radio (Continued)

8.7 USING AN EXTERNAL EEPROM FOR NON-VOLATILE DATA

The LMX9830 offers two interfaces to connect to external memory. Depending on the EEPROM used, the interface is activated by setting the correct option pins during start up. See Table 17 on page 17 for the option pin settings.

The external memory is used to store mandatory parameters like the BD_Address as well as many optional parameters like Link Keys or even User data.

The NVM is organized with fixed addresses for the parameters. Because of that the EEPROM can be preprogrammed with default parameters in manufacturing. Refer to "Operation Parameters Stored in LMX9830" for the organization of the NVS map.

In case the external memory is empty on first startup the LMX9830 will behave as like no memory is connected. (See Section 8.6.3 "Startup Without External PROM Available" on page 17). During the startup process parameters can be written directly to the EEPROM to be available after next bootup. On first bootup, the EEPROM will be automatically programmed to default values, including the UART speed of 9600 BPS. Patches supplied over the TL will be stored automatically into the EEPROM.

9.0 Digital Smart Radio

9.1 FUNCTIONAL DESCRIPTION

The integrated Digital Smart Radio utilizes a heterodyne receiver architecture with a low intermediate frequency (2 MHz) such that the intermediate frequency filters can be integrated on chip. The receiver consists of a low-noise amplifier (LNA) followed by two mixers. The intermediate frequency signal processing blocks consist of a poly-phase bandpass filter (BPF), two hard-limiters (LIM), a frequency discriminator (DET), and a post-detection filter (PDF). The received signal level is detected by a received signal strength indicator (RSSI).

The received frequency equals the local oscillator frequency (fLO) plus the intermediate frequency (fIF):

$$f_{RF} = f_{LO} + f_{IF} \text{ (supradyn)}$$

The radio includes a synthesizer consisting of a phase detector, a charge pump, an (off-chip) loop-filter, an RF-frequency divider, and a voltage controlled oscillator (VCO).

The transmitter utilizes IQ-modulation with bit-stream data that is gaussian filtered. Other blocks included in the transmitter are a VCO buffer and a power amplifier (PA).

9.2 RECEIVER FRONT-END

The receiver front-end consists of a low-noise amplifier (LNA) followed by two mixers and two low-pass filters for the I- and Q-channels.

The intermediate frequency (IF) part of the receiver front-end consists of two IF amplifiers that receive input signals from the mixers, delivering balanced I- and Q-signals to the poly-phase bandpass filter. The poly-phase bandpass filter

is directly followed by two hard-limiters that together generate an AD-converted RSSI signal.

9.2.1 Poly-Phase Bandpass Filter

The purpose of the IF bandpass filter is to reject noise and spurious (mainly adjacent channel) interference that would otherwise enter the hard limiting stage. In addition, it takes care of the image rejection.

The bandpass filter uses both the I- and Q-signals from the mixers. The out-of-band suppression should be higher than 40 dB ($f < 1$ MHz, $f > 3$ MHz). The bandpass filter is tuned over process spread and temperature variations by the autotuner circuitry. A 5th order Butterworth filter is used.

9.2.2 Hard-Limiter and RSSI

The I- and Q-outputs of the bandpass filter are each followed by a hard-limiter. The hard-limiter has its own reference current. The RSSI (Received Signal Strength Indicator) measures the level of the RF input signal.

The RSSI is generated by piece-wise linear approximation of the level of the RF signal. The RSSI has a mV/dB scale, and an analog-to-digital converter for processing by the baseband circuit. The input RF power is converted to a 5-bit value. The RSSI value is then proportional to the input power (in dBm).

The digital output from the ADC is sampled on the BPK-TCTL signal low-to-high transition.

9.3 RECEIVER BACK-END

The hard-limiters are followed by a two frequency discriminators. The I-frequency discriminator uses the 90° phase-shifted signal from the Q-path, while the Q-discriminator uses the 90° phase-shifted signal from the I-path. A poly-phase bandpass filter performs the required phase shifting. The output signals of the I- and Q-discriminator are subtracted and filtered by a low-pass filter. An equalizer is added to improve the eye-pattern for 101010 patterns.

After equalization, a dynamic AFC (automatic frequency offset compensation) circuit and slicer extract the RX_DATA from the analog data pattern. It is expected that the Eb/No of the demodulator is approximately 17 dB.

9.3.1 Frequency Discriminator

The frequency discriminator gets its input signals from the limiter. A defined signal level (independent of the power supply voltage) is needed to obtain the input signal. Both inputs of the frequency discriminator have limiting circuits to optimize performance. The bandpass filter in the frequency discriminator is tuned by the autotuning circuitry.

9.3.2 Post-Detection Filter and Equalizer

The output signals of the FM discriminator first go through a post-detection filter and then through an equalizer. Both the post-detection filter and equalizer are tuned to the proper frequency by the autotuning circuitry. The post-detection filter is a low-pass filter intended to suppress all remaining spurious signals, such as the second harmonic (4 MHz) from the FM detector and noise generated after the limiter.

9.0 Digital Smart Radio (Continued)

The post-detection filter also helps for attenuating the first adjacent channel signal. The equalizer improves the eye-opening for 101010 patterns. The post-detection filter is a third order Butterworth filter.

9.4 AUTOTUNING CIRCUITRY

The autotuning circuitry is used for tuning the bandpass filter, the detector, the post-detection filter, the equalizer, and the transmit filters for process and temperature variations. The circuit also includes an offset compensation for the FM detector.

9.5 SYNTHESIZER

The synthesizer consists of a phase-frequency detector, a charge pump, a low-pass loop filter, a programmable frequency divider, a voltage-controlled oscillator (VCO), a delta-sigma modulator, and a lookup table.

The frequency divider consists of a divide-by-2 circuit (divides the 5 GHz signal from the VCO down to 2.5 GHz), a divide-by-8-or-9 divider, and a digital modulus control. The delta-sigma modulator controls the division ratio and also generates an input channel value to the lookup table.

9.5.1 Phase-Frequency Detector

The phase-frequency detector is a 5-state phase-detector. It responds only to transitions, hence phase-error is independent of input waveform duty cycle or amplitude variations. Loop lockup occurs when all the negative transitions on the inputs, F_REF and F_MOD, coincide. Both outputs (i.e., Up and Down) then remain high. This is equal to the zero error mode. The phase-frequency detector input frequency range operates at 12MHz.

9.6 TRANSMITTER CIRCUITRY

The transmitter consists of ROM tables, two Digital to Analog (DA) converters, two low-pass filters, IQ mixers, and a power amplifier (PA).

The ROM tables generate a digital IQ signal based on the transmit data. The output of the ROM tables is inserted into IQ-DA converters and filtered through two low-pass filters. The two signal components are mixed up to 2.5 GHz by the TX mixers and added together before being inserted into the transmit PA.

9.6.1 IQ-DA Converters and TX Mixers

The ROM output signals drive an I- and a Q-DA converter. Two Butterworth low-pass filters filter the DA output signals. The 6 MHz clock for the DA converters and the logic circuitry around the ROM tables are derived from the autotuner.

The TX mixers mix the balanced I- and Q-signals up to 2.4-2.5 GHz. The output signals of the I- and Q-mixers are summed.

9.7 CRYSTAL REQUIREMENTS

The LMX9830 contains a crystal driver circuit. This circuit operates with an external crystal and capacitors to form an oscillator. Figure 12 on page 22 shows the recommended

crystal circuit. Table 22 on page 24 specifies system clock requirements.

The RF local oscillator and internal digital clocks for the LMX9830 is derived from the reference clock at the CLK+ input. This reference may either come from an external clock or a dedicated crystal oscillator. The crystal oscillator connections require an Xtal and two grounded capacitors.

It is also important to consider board and design dependent capacitance in tuning crystal circuit. Equations that follow allow a close approximation of crystal tuning capacitance required, but actual values on board will vary with capacitive properties of the board. As a result, there is some fine tuning of crystal circuit that has to be done that can not be calculated, must be tuned by testing different values of load capacitance.

Many different crystals can be used with the LMX9830. Key requirements from Bluetooth specification is ± 20 ppm. Additionally, ESR (Equivalent Series Resistance) must be carefully considered. LMX9830 can support maximum of 230ohm ESR, but it is recommended to stay <100ohms ESR for best performance over voltage and temperature. Reference Figure 17 on page 25 for ESR as part of crystal circuit for more information.

9.7.1 Crystal

The crystal appears inductive near its resonant frequency. It forms a resonant circuit with its load capacitors. The resonant frequency may be trimmed with the crystal load capacitance.

1. Load Capacitance

For resonance at the correct frequency, the crystal should be loaded with its specified load capacitance, which is the value of capacitance used in conjunction with the crystal unit. Load capacitance is a parameter specified by the crystal, typically expressed in pF. The crystal circuit shown in Figure 13 on page 22 is composed of:

- C1 (motional capacitance)
- R1 (motional resistance)
- L1 (motional inductance)
- C0 (static or shunt capacitance)

The LMX9830 provides some of the load with internal capacitors C_{int} . The remainder must come from the external capacitors and tuning capacitors labeled Ct1 and Ct2 as shown in Figure 12 on page 22. Ct1 and Ct2 should have the same the value for best noise performance.

The LMX9830 has an additional internal capacitance C_{TUNE} of 2.6pF. Crystal load capacitance (C_L) is calculated as the following:

$$C_L = C_{int} + C_{TUNE} + Ct1//Ct2$$

The C_L above does not include the crystal internal self-capacitance C_0 as shown in Figure 13 on page 22, so the total capacitance is:

$$C_{total} = C_L + C_0$$

9.0 Digital Smart Radio (Continued)

Based on crystal spec and equation:

$$C_L = C_{int} + C_{TUNE} + Ct1//Ct2$$

$$C_L = 8\text{pF} + 2.6\text{pF} + 6\text{pF} = 16.6\text{pF}$$

16.6pF is very close to the TEW crystal requirement of 16pF load capacitance. With the internal shunt capacitance C_{total} :

$$C_{total} = 16.6\text{pF} + 5\text{pF} = 21.6\text{pF}$$

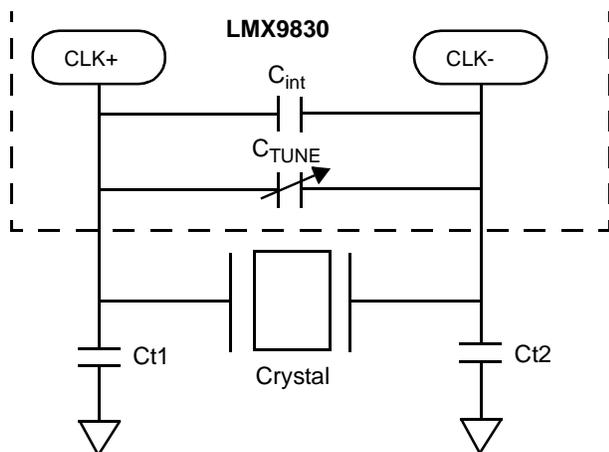


Figure 12. LMX9830 Crystal Recommended Circuit

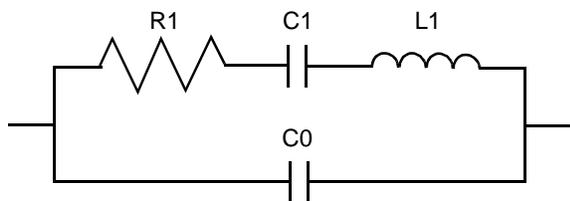


Figure 13. Crystal Equivalent Circuit

2. Crystal Pullability

Pullability is another important parameter for a crystal, which is the change in frequency of a crystal with units of ppm/pF, either from the natural resonant frequency to a load resonant frequency, or from one load resonant frequency to another. The frequency can be pulled in a parallel resonant circuit by changing the value of load capacitance. A decrease in load capacitance causes an increase in frequency, and an increase in load capacitance causes a decrease in frequency.

3. Frequency Tuning

Frequency Tuning is achieved by adjusting the crystal load capacitance with external capacitors. It is a Bluetooth requirement that the frequency is always within ± 20 ppm. Crystal/oscillator must have cumulative accuracy specifications of ± 15 ppm to provide margin for frequency drift with aging and temperature.

TEW Crystal

The LMX9830 has been tested with the TEW TAS-4025A crystal, reference Table 19 on page 22 for specification. Since the internal capacitance of the crystal circuit is 8 pF and the load capacitance is 16 pF, 12 pF is a good starting point for both Ct1 and Ct2. The 2480 MHz RF frequency offset is then tested. Figure 14 on page 23 shows the RF frequency offset test results.

Figure 14 on page 23 shows the results are -20 kHz off the center frequency, which is -1 ppm. The pullability of the crystal is 2 ppm/pF, so the load capacitance must be decreased by about 1.0 pF. By changing Ct1 or Ct2 to 10 pF, the total load capacitance is decreased by 1.0 pF. Figure 15 on page 23 shows the frequency offset test results. The frequency offset is now zero with Ct1 = 10 pF, Ct2 = 10 pF.

Reference Table 20 on page 22 for crystal tuning values used on Mesa Development Board with TEW crystal.

Table 19. TEW TAS-4025A

| Specification | Value |
|-------------------------|------------------------------|
| Package | 4.0x2.5x0.65 mm - 4 pads |
| Frequency | 13.000 MHz |
| Mode | Fundamental |
| Stability | > ± 15 ppm @ -40 to +85C |
| C_L Load Capacitance | 16pF |
| ESR | 80 Ω max. |
| C_0 Shunt Capacitance | 5pF |
| Drive Level | 50 ± 10 uV |
| Pullability | 2 ppm/pF min |
| Storage Temperature | -40 to +85C |

Table 20. TEW on LMX9830 DONGLE

| Reference | LMX9830 |
|-----------|---------|
| Ct1 | 12pF |
| Ct2 | 12pF |

9.0 Digital Smart Radio (Continued)

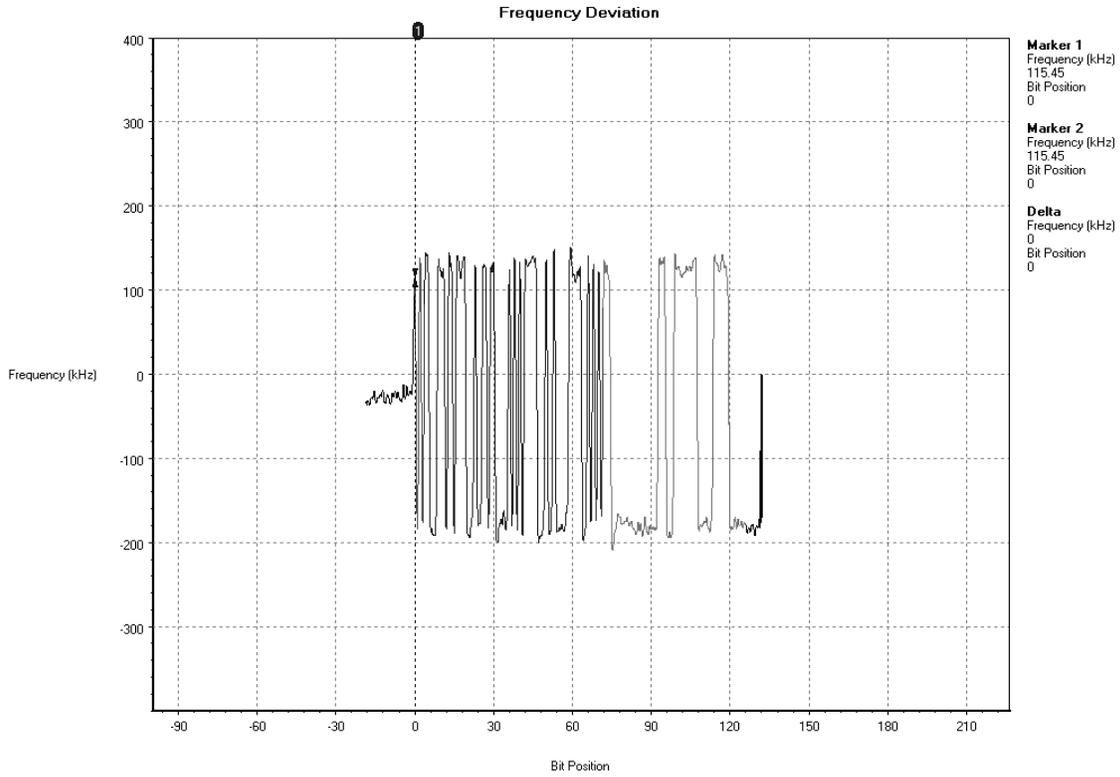


Figure 14. Frequency Offset with 12 pF//12 pF Capacitors

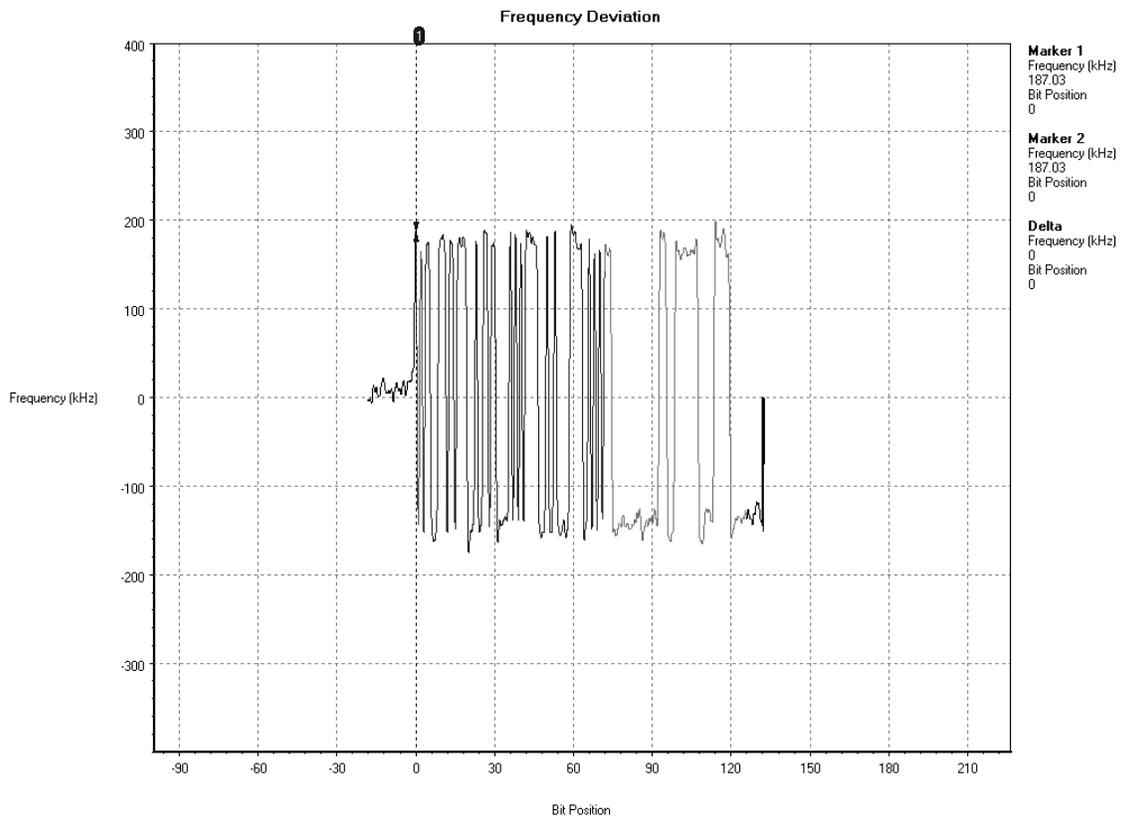


Figure 15. Frequency Offset with 10 pF//10 pF Capacitors

9.0 Digital Smart Radio (Continued)

9.7.2 TCXO (Temperature Compensated Crystal Oscillator)

The LMX9830 also can operate with an external TCXO (Temperature Compensated Crystal Oscillator). The TCXO signal is directly connected to the CLK+.

1. Input Impedance

The LMX9830 CLK+ pin has an input impedance of 2pF capacitance in parallel with >400kΩ resistance

9.7.3 Optional 32 kHz Oscillator

A second oscillator is provided (see Figure 16) that is tuned to provide optimum performance and low-power consumption while operating with a 32.768 kHz crystal. An external crystal clock network is required between the 32kHz_CLKI clock input (pad B13) and the 32kHz_CLKO clock output (pad C13) signals. The oscillator is built in a Pierce configuration

and uses two external capacitors. Table 21 provides the oscillator's specifications.

In case the 32Khz is placed optionally, it is recommended to remove C2 and replace C1 with a zero ohm resistor.

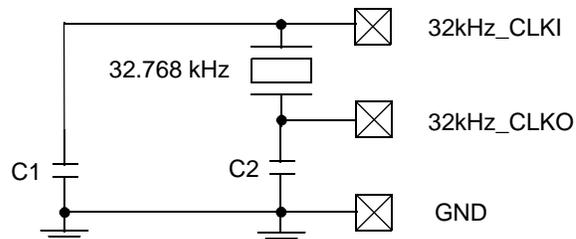


Figure 16. 32.768 kHz Oscillator

Table 21. 32.768 kHz Oscillator Specifications

| Symbol | Parameter | Condition | Min | Typ | Max | Unit |
|--------------------|--------------------------|-----------|------|--------|------|------|
| V _{DD} | Supply Voltage | | 1.62 | 1.8 | 1.98 | V |
| I _{DDACT} | Supply Current (Active) | | | 2 | | μA |
| f | Nominal Output Frequency | | | 32.768 | | kHz |
| V _{PPOSC} | Oscillating Amplitude | | | 1.8 | | V |
| | Duty Cycle | | 40 | | 60 | % |

9.7.4 ESR (Equivalent Series Resistance)

LMX9830 can operate with a wide range of crystals with different ESR ratings. Reference Table 22 on page 24 and Figure 17 on page 25 for more details.

Table 22. System Clock Requirements

| Parameter | Min | Typ | Max | Unit |
|---|-----|-----|-----|--------------|
| External Reference Clock Frequency ¹ | 10 | 13 | 20 | MHz |
| Frequency Tolerance (over full operating temperature and aging) | -20 | ±15 | +20 | ppm |
| Crystal Serial Resistance | | | 230 | Ω |
| External Reference Clock Power Swing, pk to pk | 100 | 200 | 400 | mV |
| Aging | | | ±1 | ppm per year |

1. Supported frequencies **from external oscillator** (in MHz): 10.00, 10.368, 12.00, 12.60, 12.80, 13.00, 13.824, 14.40, 15.36, 16.00, 16.20, 16.80, 19.20, 19.68, 19.80

9.0 Digital Smart Radio (Continued)

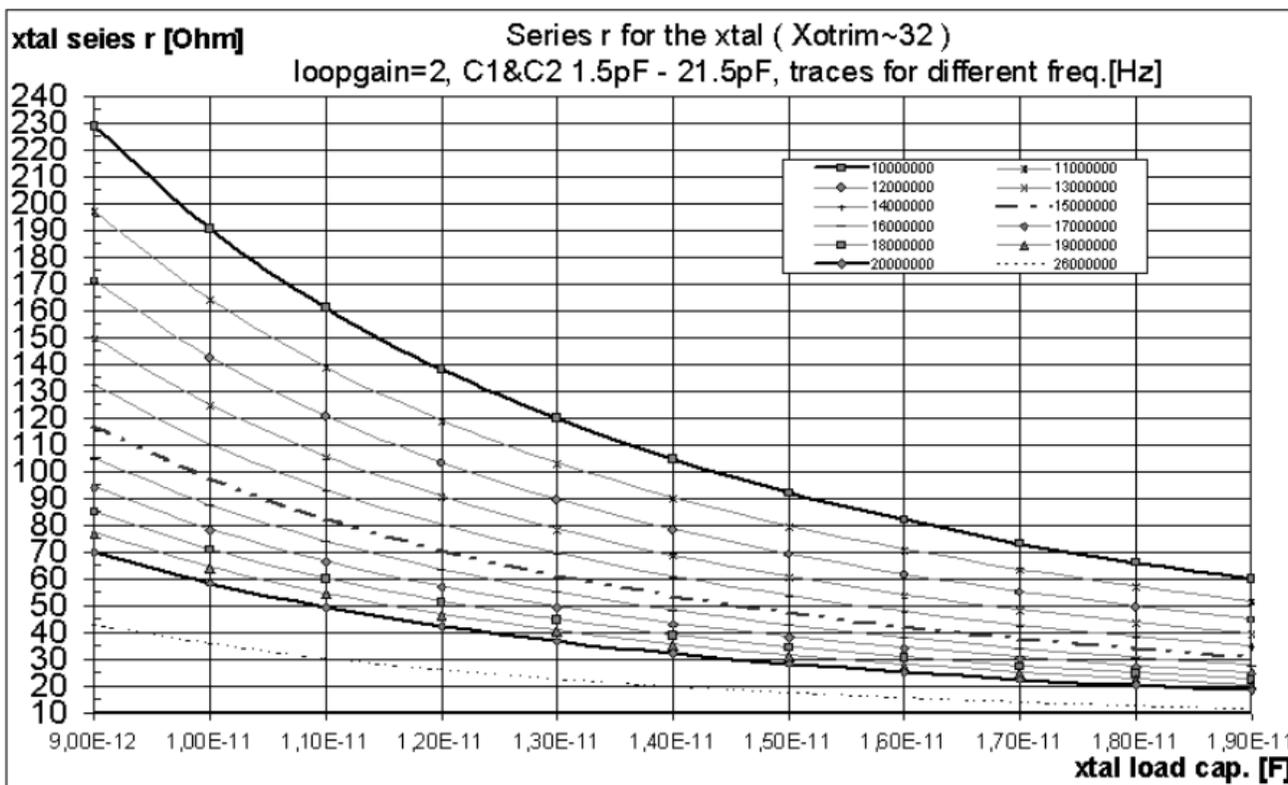


Figure 17. ESR vs. Load capacitance for the crystal circuit

9.8 ANTENNA MATCHING AND FRONT-END FILTERING

Figure 18 shows the recommended component layout to be used between RF output and antenna input. Allows for versatility in the design such that the match to the antenna maybe improved and/or the blocking margin increased by addition of a LC filter. Refer to antenna application note for further details.

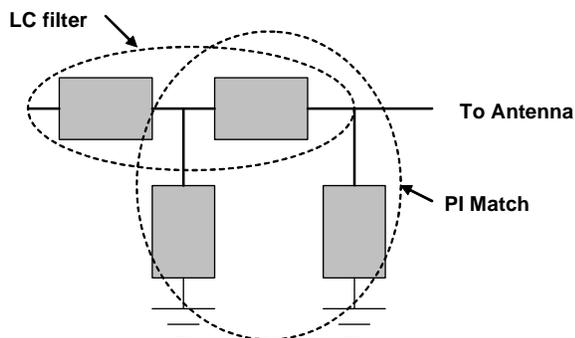


Figure 18. Front end Layout

9.9 LOOP FILTER DESIGN

The LMX9830 has an external loop filter which must be designed for best performance by the end customer. This section therefore gives some foresight into its design. Refer also to Loop Filter application note and National's Webench on-line design tool for more information.

9.9.1 Component Calculations

The following parameters are required for component value calculation of a third order passive loop filter.

| | |
|------------|--|
| ϕ | Phase Margin: Phase of the open loop transfer function |
| F_c | Loop Bandwidth |
| F_{comp} | Comparison Frequency: Phase detector frequency |
| KVOC | VCO gain: Sensitivity of the VCO to control volts |
| $K\Phi$ | Charge Pump gain: Magnitude of the alternating current during lock |
| F_{OUT} | Maximum RF output frequency |
| T31 | Ratio of the poles T3 to T1 in a 3 rd order filter |
| γ | Gamma optimization parameter |

9.0 Digital Smart Radio (Continued)

The third order loop filter being defined has the following topology, shown in Figure 19..

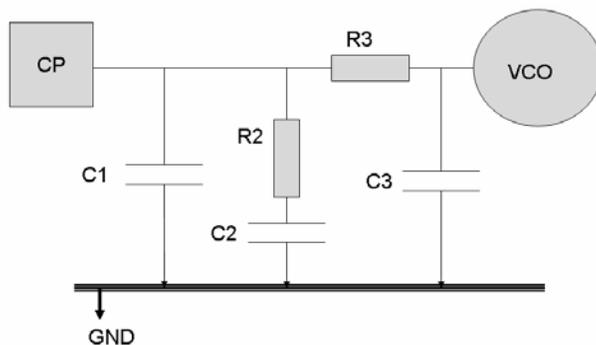


Figure 19. Third Order Loop filter

$$N = \frac{F_{out}}{F_{comp}} \quad \text{and} \quad \omega_C = 2\pi F_C$$

Calculate the poles and zeros. Use exact method to solve for T1 using numerical methods,

$$\phi = \tan^{-1}\left(\frac{\gamma}{\omega_C \cdot T1 \cdot T1 + T31}\right) - \tan^{-1}(\omega_C \cdot T1) - \tan^{-1}(\omega_C \cdot T1 \cdot T31)$$

$$T3 = T31 \times T1 \quad T2 = \frac{\gamma}{\omega_C^2 \cdot (T1 + T3)}$$

Calculate the loop filter coefficients,

$$A0 = \frac{K\phi \cdot K_{vco}}{\omega_C^2 \cdot N} \cdot \sqrt{\frac{1 + \omega_C^2 \cdot T2^2}{(1 + \omega_C^2 \cdot T1^2)(1 + \omega_C^2 \cdot T3^2)}}$$

$$A1 = A0 \cdot (T1 + T3) \quad A2 = A0 \cdot T1 \cdot T3$$

Summary;

| Symbol | Description | Units |
|--------|--------------------------------------|-------|
| n | N counter value | None |
| | Loop Bandwidth | rad/s |
| T1 | Loop filter pole | S |
| T2 | Loop filter zero | S |
| T3 | Loop filter zero | S |
| A0 | Total capacitance | nF |
| A1 | First order loop filter coefficient | nFs |
| A2 | Second order loop filter coefficient | nFs2 |

9.0 Digital Smart Radio (Continued)

Components can then be calculated from loop filter coefficients

$$C1 = \frac{A2}{T2^2} \cdot \left(1 + \sqrt{1 + \left(\frac{T2 \cdot A0 - T2 \cdot A1}{A2}\right)^2}\right)$$

$$C3 = \frac{1 \cdot T2^2 \cdot C1^2 + T2 \cdot A1 \cdot C1 - A2 \cdot A0}{T2^2 \cdot C1 - A2} \quad C2 = A0 - C1 - C3$$

$$R2 = \frac{T2}{C2} \quad R3 = \frac{A2}{C1 \cdot C3 \cdot T2}$$

Some typical values for the LMX9830 are:

| | | |
|----------------------|------|-----------|
| Comparison Frequency | 13 | MHz |
| Phase Margin | 48 | PI rad |
| Loop bandwidth | 100 | kHz |
| T3 over T1 ratio | 40 | % |
| Gamma | 1.0 | |
| VCO gain | 120 | MHz per V |
| Charge pump gain | 0.6 | mA |
| Fout | 2441 | MHz |

$$\text{Phase noise (in-band)} = \text{PN1Hz} + 20\text{Log}[N] + 10\text{Log}[F_{\text{comp}}]$$

Where PH1Hz is the PLL normalized noise floor in 1 Hz resolution bandwidth.

Further out from the carrier, the phase noise will be affected by the loop filter roll-off and hence its bandwidth.

$$\text{As a rule-of-thumb; } \Delta \text{ Phase noise} = 40\text{Log}[\Delta F_c]$$

Where ΔF_c is the relative change in loop BW expressed as a fraction.

Which give the following component values:

| | | |
|----|------|------|
| C1 | 0.17 | nF |
| C2 | 2.38 | nF |
| C3 | 0.04 | nF |
| R2 | 1737 | ohms |
| R3 | 7025 | ohms |

For example if the loop bandwidth is reduced from 100kHz to 50kHz or by one half, then the change in phase noise will be -12dB. Loop BW in reality should be selected to meet the lower limit of the modulation deviation, this will yield the best possible phase noise.

Even further out from the carrier, the phase noise will be mainly dominated by the VCO noise assuming the crystal is relatively clean.

Lock-time is dependent on three factors, the loop bandwidth, the maximum frequency jump that the PLL must make and the final tolerance to which the frequency must settle. As a rule-of-thumb it is given by:

9.9.2 Phase Noise and Lock-Time Calculations

Phase noise has three sources, the VCO, crystal oscillator and the rest of the PLL consisting of the phase detector, dividers, charge pump and loop filter. Assuming the VCO and crystal are very low noise, it is possible to put down approximate equations that govern the phase noise of the PLL.

$$LT = \frac{400}{F_c} (1 - \log_{10} \Delta F) \quad \text{Where } \Delta F = \frac{\text{Frequency} - \text{tolerance}}{\text{Frequency} - \text{jump}}$$

These equations are approximations of the ones used by Webench to calculate phase noise and lock-time.

9.0 Digital Smart Radio (Continued)

9.9.3 Practical Optimization

In an example where frequency drift and drift rate can be improved through loop filter tweaks, consider the results taken below. The drift rate is 26.1 kHz per 50us and the maximum drift is 25 kHz for DH1 packets, both of which are

exceeding or touching the Bluetooth pass limits. These measurements are taken with component values shown above

TRM/CA09/C (Carrier Drift)

Hopping On - Low Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|------------|------------|------------|----------------|
| Drift Rate / 50us | 26,1 kHz | N/A | -30,5 kHz | +/-20 kHz |
| Max Drift | 25 kHz | N/A | 36 kHz | DH1: +/-25 kHz |
| Average Drift | -1 kHz | N/A | 12 kHz | DH3: +/-40 kHz |
| Packets Tested | 10 | N/A | 10 | DH5: +/-40 kHz |
| Packets Failed | 2 | N/A | 10 | |
| Overall Result | Failed | N/A | Failed | |

Results below were taken on the same board with three loop filter values changed. C2 and R2 have been increased in value and C1 has been reduced. The drift rate

has improved by 13 kHz per 50 μs and the maximum drift has improved by 10 kHz

TRM/CA09/C (Carrier Drift)

Hopping On - Low Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|------------|------------|------------|----------------|
| Drift Rate / 50us | -13,6 kHz | N/A | 15,6 kHz | +/-20 kHz |
| Max Drift | 15 kHz | N/A | 21 kHz | DH1: +/-25 kHz |
| Average Drift | 3 kHz | N/A | 1 kHz | DH3: +/-40 kHz |
| Packets Tested | 10 | N/A | 10 | DH5: +/-40 kHz |
| Packets Failed | 0 | N/A | 0 | |
| Overall Result | Passed | N/A | Passed | |

The effect of changing these three components is to reduce the loop bandwidth which reduces the phase noise. The reduction in this noise level corresponds directly to the reduction of noise in the payload area where drift is measured. This noise reduction comes at the expense of lock-time which can be increased to 120 μs without suffering any ill effects, however if we continue to reduce the loop BW further the lock-time will increase such that the PLL does not have time to lock before data transmission and the drift will again increase. Before the lock-time goes out of spec, the modulation index will start to fall since it is being cut by the reducing loop BW. Therefore a compromise has to be found between lock-time, phase noise and modulation, which yields best performance.

Note: The values shown in the LMX9830 datasheet, are the best case optimized values that have been shown to produce the best overall results and are recommended as a starting point for this design.

Another example of how the loop filter values can affect frequency drift rate, these results below show the DUT with maximum drift on mid and high channels failing. Adjusting the loop bandwidth as shown provides the improvement required to pass qualification.

9.0 Digital Smart Radio (Continued)

Original results:

Hopping OFF - Low Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|------------|------------|------------|----------------|
| Drift Rate / 50uS | -15.00 kHz | -28.10 kHz | -19.10 kHz | +/- 20 kHz |
| Maximum Drift | -19 kHz | -37 kHz | -20 kHz | DH1: +/- 25kHz |
| Average Drift | -11 kHz | -32 kHz | -10 kHz | DH3: +/- 40kHz |
| Packets Tested | 10 | 10 | 10 | DH5: +/- 40kHz |
| Packets Failed | 0 | 1 | 0 | |
| Result | Pass | Fail | Pass | |

Hopping OFF - Med Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|-------------------|------------|------------|----------------|
| Drift Rate / 50uS | -18.60 kHz | 16.30 kHz | -18.00 kHz | +/- 20 kHz |
| Maximum Drift | -29 kHz | -44 kHz | -28 kHz | DH1: +/- 25kHz |
| Average Drift | -19 kHz | -37 kHz | -19 kHz | DH3: +/- 40kHz |
| Packets Tested | 10 | 10 | 10 | DH5: +/- 40kHz |
| Packets Failed | 2 | 2 | 0 | |
| Result | Fail | Fail | Pass | |

Hopping OFF - High Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|-------------------|------------|------------|----------------|
| Drift Rate / 50uS | -16.30 kHz | 16.80 kHz | -17.70 kHz | +/- 20 kHz |
| Maximum Drift | -36 kHz | -61 kHz | -38 kHz | DH1: +/- 25kHz |
| Average Drift | -31 kHz | -48 kHz | -29 kHz | DH3: +/- 40kHz |
| Packets Tested | 10 | 10 | 10 | DH5: +/- 40kHz |
| Packets Failed | 10 | 8 | 0 | |
| Result | Fail | Fail | Pass | |

New results:

Hopping OFF - Low Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|------------|------------|------------|----------------|
| Drift Rate / 50uS | -12.00 kHz | -15.10 kHz | 18.80 kHz | +/- 20 kHz |
| Maximum Drift | -15 kHz | -35 kHz | -19 kHz | DH1: +/- 25kHz |
| Average Drift | -6 kHz | -25 kHz | -9 kHz | DH3: +/- 40kHz |
| Packets Tested | 10 | 10 | 10 | DH5: +/- 40kHz |
| Packets Failed | 0 | 0 | 0 | |
| Result | Pass | Pass | Pass | |

Hopping OFF - Med Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|------------|------------|------------|----------------|
| Drift Rate / 50uS | -14.20 kHz | -16.10 kHz | 17.20 kHz | +/- 20 kHz |
| Maximum Drift | -16 kHz | -34 kHz | -22 kHz | DH1: +/- 25kHz |
| Average Drift | -11 kHz | -27 kHz | -9 kHz | DH3: +/- 40kHz |
| Packets Tested | 10 | 10 | 10 | DH5: +/- 40kHz |
| Packets Failed | 0 | 0 | 0 | |
| Result | Pass | Pass | Pass | |

Hopping OFF - High Channel

| | <u>DH1</u> | <u>DH3</u> | <u>DH5</u> | <u>Limits</u> |
|-------------------|------------|------------|------------|----------------|
| Drift Rate / 50uS | -12.70 kHz | -17.40 kHz | -16.50 kHz | +/- 20 kHz |
| Maximum Drift | -23 kHz | -29 kHz | -25 kHz | DH1: +/- 25kHz |
| Average Drift | -12 kHz | -25 kHz | -16 kHz | DH3: +/- 40kHz |
| Packets Tested | 10 | 10 | 10 | DH5: +/- 40kHz |
| Packets Failed | 0 | 0 | 0 | |
| Result | Pass | Pass | Pass | |

10.0 Integrated Firmware (Continued)

9.9.4 Component Values for NSC Reference Designs

The following is a list of components for the loop filter values used on National reference design, (Serial Dongle) they have been tweaked and optimized in each case to yield optimum performance for each case. The values differ slightly from one platform to another due to board parasitics caused by layout differences.

| Platform | C8 | C7 | C9 | R23 | R14 |
|----------------|-------|--------|------|------|-----|
| LMX9830 Dongle | 220pF | 2200pF | 39pF | 3.3k | 10k |

10.0 Integrated Firmware

The LMX9830 includes the full Bluetooth stack up to RFCOMM to support the following profiles:

- GAP (Generic Access Profile)
- SDAP (Service Discovery Application Profile)
- SPP (Serial Port Profile)

Figure 20 shows the Bluetooth protocol stack with command interpreter interface. The command interpreter offers a number of different commands to support the functionality given by the different profiles. Execution and interface timing is handled by the control application.

The chip has an internal data area in RAM that includes the parameters shown in Table 23 on page 31.

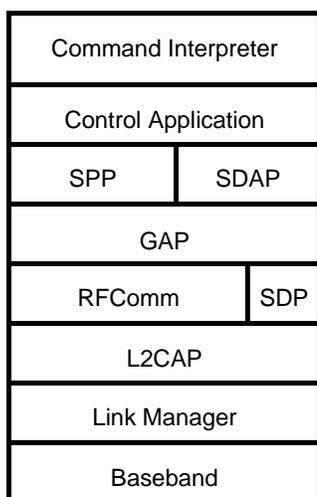


Figure 20. LMX9830 Software Implementation

10.1 FEATURES

10.1.1 Operation Modes

On boot-up, the application configures the module following the parameters in the data area.

Automatic Operation

No Default Connections Stored:

In Automatic Operation the module is connectable and discoverable and automatically answers to service requests.

The command interpreter listens to commands and links can be set up. The full command list is supported.

If connected by another device, the module sends an event back to the host, where the RFCOMM port has been connected, and switches to transparent mode.

Default Connections Stored:

If default connections were stored on a previous session, once the LMX9830 is reset, it will attempt to connect each device stored within the data RAM three times. The host will be notified about the success of the link setup via a link status event.

Non-Automatic Operation

In Non-Automatic Operation, the LMX9830 does not check the default connections section within the Data RAM. If connected by another device, it will NOT switch to transparent mode and continue to interpret data sent on the UART.

Transparent Mode

The LMX9830 supports transparent data communication from the UART interface to a bluetooth link.

If activated, the module does not interpret the commands on the UART which normally are used to configure and control the module. The packages don't need to be formatted as described in Table 25 on page 34. Instead all data are directly passed through the firmware to the active bluetooth link and the remote device.

Transparent mode can only be supported on a point-to-point connection. To leave Transparent mode, the host must send a UART_BREAK signal to the module

Force Master Mode

In Force Master mode tries to act like an Accesspoint for multiple connections. For this it will only accept the link if a Master/slave role switch is accepted by the connecting device. After successful link establishment the LMX9830 will be Master and available for additional incoming links. On the first incoming link the LMX9830 will switch to transparent depending on the setting for automatic or command mode. Additional links will only be possible if the device is not in transparent mode.

10.1.2 Default Connections

The LMX9830 supports the storage of up to 3 devices within its NVS. Those connections can either be connected after reset or on demand using a specific command.

10.1.3 Event Filter

The LMX9830 uses events or indicators to notify the host about successful commands or changes at the bluetooth interface. Depending on the application the LMX9830 can be configured. The following levels are defined:

- No Events:
 - The LMX9830 is not reporting any events. Optimized for passive cable replacement solutions.
- Standard LMX9830 events:
 - only necessary events will be reported
- All events:
 - Additional to the standard all changes at the physical layer will be reported.

10.0 Integrated Firmware (Continued)

10.1.4 Default Link Policy

Each Bluetooth Link can be configured to support M/S role switch, Hold Mode, Sniff Mode and Park Mode. The default link policy defines the standard setting for incoming and outgoing connections.

10.1.5 Audio Support

The LMX9830 offers commands to establish and release synchronous connections (SCO) to support Headset or Handsfree applications. The firmware supports one active link with all available package types (HV1, HV2, HV3), routing the audio data between the bluetooth link and the advanced audio interface. In order to provide the analog data interface, an external audio codec is required. The LMX9830 includes a list of codecs which can be used.

Table 23. Operation Parameters Stored in LMX9830

| Parameter | Default Value | Description |
|------------------------|---|--|
| BDADDR | (To be requested from IEEE) | Bluetooth device address |
| Local Name | Serial port device | Friendly Name |
| PinCode | 0000 | Bluetooth PinCode |
| Operation Mode | Automatic ON | Automatic mode ON or OFF |
| Default Connections | 0 | Up to seven default devices to connect to |
| SDP Database | 1 SPP entry: Name: COM1 Authentication and encryption enabled | Service discovery database, control for supported profiles |
| UART Speed | 9600 | Sets the speed of the physical UART interface to the host |
| UART Settings | 1 Stop bit, parity disabled | Parity and stop bits on the hardware UART interface |
| Ports to Open | 0000 0001 | Defines the RFCOMM ports to open |
| Link Keys | No link keys | Link keys for paired devices |
| Security Mode | 2 | Security mode |
| Page Scan Mode | Connectable | Connectable/Not connectable for other devices |
| Inquiry Scan Mode | Discoverable | Discoverable/Not Discoverable/Limited Discoverable for other devices |
| Default Link Policy | All modes allowed | Configures modes allowed for incoming or outgoing connections (Role switch, Hold mode, Sniff mode...) |
| Default Link Timeout | 20 seconds | The Default Link Timeout configures the timeout, after which the link is assumed lost, if no packages have been received from the remote device. |
| Event Filter | Standard LMX9830 events reported | Defines the level of reporting on the UART - no events - standard events - standard including ACL link events |
| Default Audio Settings | none | Configures the settings for the external codec and the air format. <ul style="list-style-type: none"> • Codecs: <ul style="list-style-type: none"> — Motorola MC145483 / Winbond W681360 — OKI MSM7717 / Winbond W681310 — PCM Slave • Airformat: <ul style="list-style-type: none"> — CVSD — μ-Law — A-Law |

11.0 Low Power Modes

The LMX9830 supports different Low Power Modes to reduce power in different operating situations. The modular structure of the LMX9830 allows the firmware to power down unused modules.

The Low power modes have influence on:

- UART transport layer
 - enabling or disabling the interface
- Bluetooth Baseband activity
 - firmware disables LLC and Radio if possible

11.1 POWER MODES

The following LMX9830 power modes, which depend on the activity level of the UART transport layer and the radio activity are defined:

The radio activity level mainly depends on application requirements and is defined by standard bluetooth operations like inquiry/page scanning or an active link.

A remote device establishing or disconnecting a link may also indirectly change the radio activity level.

The UART transport layer by default is enabled on device power up. In order to disable the transport layer the com-

mand "Disable Transport Layer" is used. Thus only the Host side command interface can disable the transport layer. Enabling the transport layer is controlled by the HW Wakeup signalling. This can be done from either the Host or the LMX9830. See also "LMX9830 Software User's Guide" for detailed information on timing and implementation requirements.

Table 24. Power Mode activity

| Power Mode | UART activity | Radio activity | Reference Clock |
|------------|---------------|----------------|------------------------|
| PM0 | OFF | OFF | none |
| PM1 | ON | OFF | Main Clock |
| PM2 | OFF | Scanning | Main Clock / 32.768khz |
| PM3 | ON | Scanning | Main Clock |
| PM4 | OFF | SPP Link | Main Clock |
| PM5 | ON | SPP Link | Main Clock |

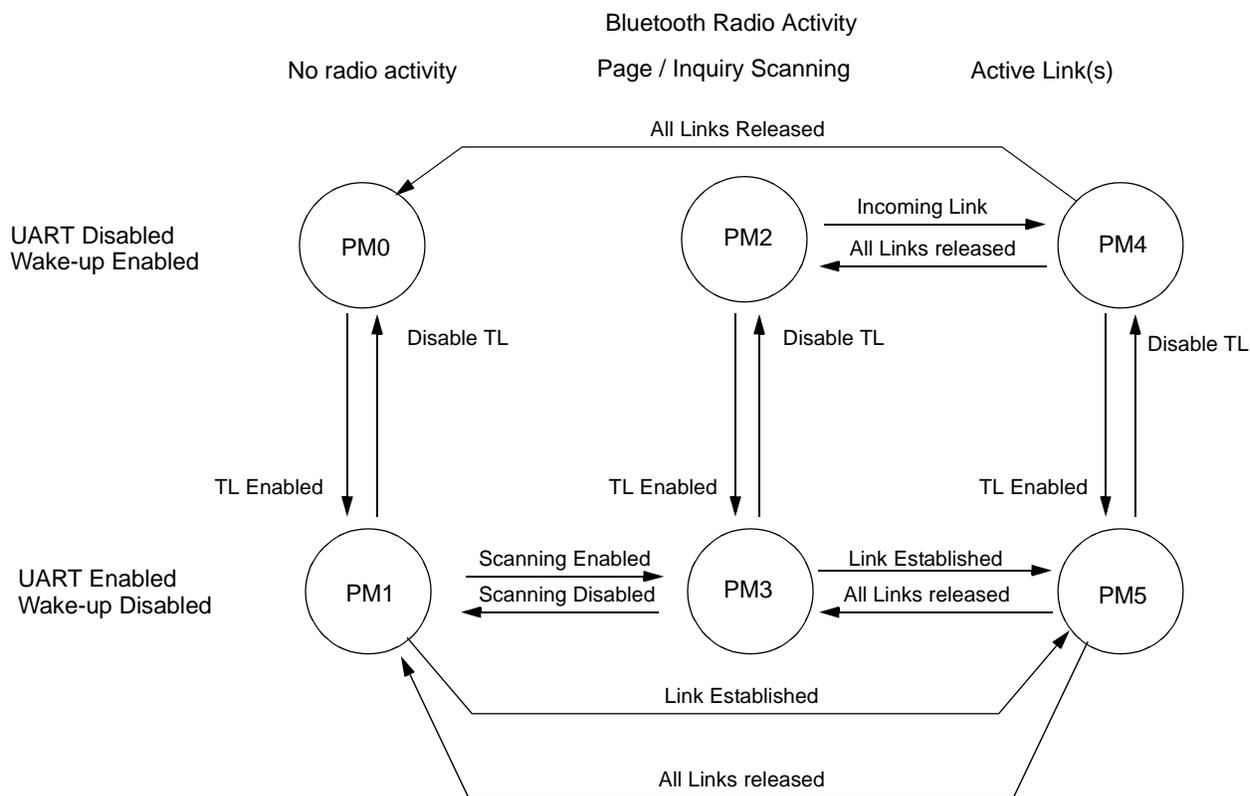


Figure 21. Transition between different Hardware Power Modes

11.2 ENABLING AND DISABLING UART TRANSPORT

11.2.1 Hardware Wake up functionality

In certain usage scenarios the host is able to switch off the transport layer of the LMX9830 in order to reduce power

consumption. Afterwards both devices, host and LMX9830 are able to shut down their UART interfaces.

In order to save system connections the UART interface is reconfigured to hardware wakeup functionality. For a detailed timing and command functionality please see also the "LMX9830 Software User's Guide".

11.0 Low Power Modes (Continued)

The interface between host and LMX9830 is defined as described in Figure 22.

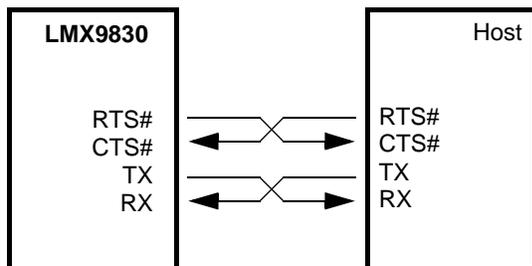


Figure 22. UART NULL modem connection

11.2.2 Disabling the UART transport layer

The Host can disable the UART transport layer by sending the “Disable Transport Layer” Command. The LMX9830 will empty its buffers, send the confirmation event and disable its UART interface. Afterwards the UART interface will be reconfigured to wake up on a falling edge of the CTS pin.

11.2.3 LMX9830 enabling the UART interface

As the Transport Layer can be disabled in any situation the LMX9830 must first make sure the transport layer is enabled before sending data to the host. Possible scenarios can be incoming data or incoming link indicators. If the UART is not enabled the LMX9830 assumes that the Host is sleeping and waking it up by activating RTS. To be able to react on that Wake up, the host has to monitor the CTS pin.

As soon as the host activates its RTS pin, the LMX9830 will first send a confirmation event and then start to transmit the events.

11.2.4 Enabling the UART transport layer from the host

If the host needs to send data or commands to the LMX9830 while the UART Transport Layer is disabled it must first assume that the LMX9830 is sleeping and wake it up using its RTS signal.

When the LMX9830 detects the Wake-Up signal it activates the UART HW and acknowledges the Wake-Up signal by settings its RTS. Additionally the Wake up will be confirmed by a confirmation event. When the Host has received this “Transport Layer Enabled” event, the LMX9830 is ready to receive commands.

12.0 Command Interface

The LMX9830 offers Bluetooth functionality in either a self contained slave functionality or over a simple command interface. The interface is listening on the UART interface.

The following sections describe the protocol transported on the UART interface between the LMX9830 and the host in command mode (see Figure 23). In Transparent mode, no data framing is necessary and the device does not listen for commands.

12.1 FRAMING

The connection is considered "Error free". But for packet recognition and synchronization, some framing is used.

All packets sent in both directions are constructed per the model shown in Table 25.

12.1.1 Start and End Delimiter

The "STX" char is used as start delimiter: STX = 0x02. ETX = 0x03 is used as end delimiter.

12.1.2 Packet Type ID

This byte identifies the type of packet. See Table 26 for details.

12.1.3 Opcode

The opcode identifies the command to execute. The opcode values can be found within the "LMX9830 Software User's Guide" included within the LMX9830 Evaluation Board.

12.1.4 Data Length

Number of bytes in the Packet Data field. The maximum size is defined with 333 data bytes per packet.

12.1.5 Checksum:

This is a simple Block Check Character (BCC) checksum of the bytes "Packet type", "Opcode" and "Data Length". The BCC checksum is calculated as low byte of the sum of all bytes (e.g., if the sum of all bytes is 0x3724, the checksum is 0x24).

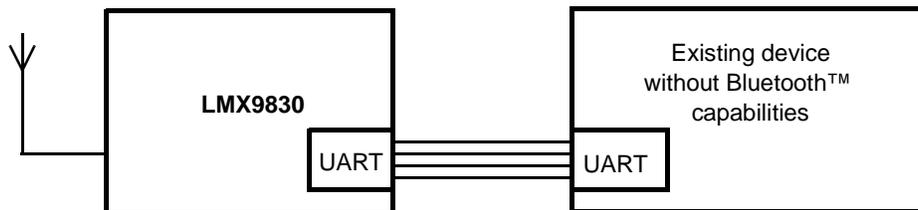


Figure 23. Bluetooth Functionality

Table 25. Package Framing

| Start Delimiter | Packet Type ID | Opcode | Data Length | Checksum | Packet Data | End Delimiter |
|----------------------|----------------|--------|-------------|----------|---------------------|---------------|
| 1 Byte | 1 Byte | 1 Byte | 2 Bytes | 1 Byte | <Data Length> Bytes | 1 Byte |
| ----- Checksum ----- | | | | | | |

Table 26. Packet Type Identification

| ID | Direction | Description |
|-------------|---------------------|--|
| 0x52 'R' | REQUEST (REQ) | A request sent to the Bluetooth module. All requests are answered by exactly one confirm. |
| 0x43 'C' | Confirm (CFM) | The Bluetooth modules confirm to a request. All requests are answered by exactly one confirm. |
| 0x69 'i' | Indication (IND) | Information sent from the Bluetooth module that is not a direct confirm to a request. Indicating status changes, incoming links, or unrequested events. |
| 0x72 'r' | Response (RES) | An optional response to an indication. This is used to respond to some type of indication message. |

12.0 Command Interface (Continued)

12.2 COMMAND SET OVERVIEW

The LMX9830 has a well defined command set to:

- Configure the device:
 - Hardware settings
 - Local Bluetooth parameters
 - Service database

- Set up and handle links

Tables 27 through 37 show the actual command set and the events coming back from the device. A full documented description of the commands can be found in the “LMX9830 Software User’s Guide”.

NOTE: For standard Bluetooth operation only commands from Table 27 through Table 29 will be used. Most of the remaining commands are for configuration purposes only..

Table 27. Device Discovery

| Command | Event | Description |
|--------------------|----------------------------|----------------------------------|
| Inquiry | Inquiry Complete | Search for devices |
| | Device Found | Lists BDADDR and class of device |
| Remote Device Name | Remote Device Name Confirm | Get name of remote device |

Table 28. SDAP Client Commands

| Command | Event | Description |
|------------------------|--------------------------------|--|
| SDAP Connect | SDAP Connect Confirm | Create an SDP connection to remote device |
| SDAP Disconnect | SDAP Disconnect Confirm | Disconnect an active SDAP link |
| | Connection Lost | Notification for lost SDAP link |
| SDAP Service Browse | Service Browse Confirm | Get the services of the remote device |
| SDAP Service Search | SDAP Service Search Confirm | Search a specific service on a remote device |
| SDAP Attribute Request | SDAP Attribute Request Confirm | Searches for services with specific attributes |

Table 29. SPP Link Establishment

| Command | Event | Description |
|--------------------|-------------------------------|--|
| Establish SPP Link | Establishing SPP Link Confirm | Initiates link establishment to a remote device |
| | Link Established | Link successfully established |
| | Incoming Link | A remote device established a link to the local device |
| Set Link Timeout | Set Link Timeout Confirm | Confirms the Supervision Timeout for the existing Link |
| Get Link Timeout | Get Link Timeout Confirm | Get the Supervision Timeout for the existing Link |
| Release SPP Link | Release SPP Link Confirm | Initiate release of SPP link |
| SPP Send Data | SPP Send Data Confirm | Send data to specific SPP port |
| | Incoming Data | Incoming data from remote device |
| Transparent Mode | Transparent Mode Confirm | Switch to Transparent mode on the UART |

Table 30. Storing Default Connections

| Command | Event | Description |
|---------------------------------|------------------------------------|--|
| Connect Default Connection | Connect Default Connection Confirm | Connects to either one or all stored default connections |
| Store Default Connection | Store Default Connection Confirm | Store device as default connection |
| Get list of Default Connections | List of Default Devices | |
| Delete Default Connections | Delete Default Connections Confirm | |

12.0 Command Interface (Continued)

Table 31. Bluetooth Low Power Modes

| Command | Event | Description |
|-------------------------|---------------------------------|---|
| Set Default Link Policy | Set Default Link Policy Confirm | Defines the link policy used for any incoming or outgoing link. |
| Get Default Link Policy | Get Default Link Policy Confirm | Returns the stored default link policy |
| Set Link Policy | Set Link Policy Confirm | Defines the modes allowed for a specific link |
| Get Link Policy | Get Link Policy Confirm | Returns the actual link policy for the link |
| Enter Sniff Mode | Enter Sniff Mode Confirm | |
| Exit Sniff Mode | Exit Sniff Mode Confirm | |
| Enter Hold Mode | Enter Hold Mode Confirm | |
| | Power Save Mode Changed | Remote device changed power save mode on the link |

Table 32. Audio Control Commands

| Command | Event | Description |
|------------------------|-----------------------------------|--|
| Establish SCO Link | Establish SCO Link Confirm | Establish SCO Link on existing RfComm Link |
| | SCO Link Established Indicator | A remote device has established a SCO link to the local device |
| Release SCO Link | Release SCO Link Confirm | Release SCO Link |
| | SCO Link Released Indicator | SCO Link has been released |
| Change SCO Packet Type | Change SCO Packet Type Confirm | Changes Packet Type for existing SCO link |
| | SCO Packet Type changed indicator | SCO Packet Type has been changed |
| Set Audio Settings | Set Audio Settings Confirm | Set Audio Settings for existing Link |
| Get Audio Settings | Get Audio Settings Confirm | Get Audio Settings for existing Link |
| Set Volume | Set Volume Confirm | Configure the volume |
| Get Volume | Get Volume Confirm | Get current volume setting |
| Mute | Mute Confirm | Mutes the microphone input |

Table 33. Wake Up Functionality

| Command | Event | Description |
|-------------------------|-------------------------|---|
| Disable Transport Layer | Transport Layer Enabled | Disabling the UART Transport Layer and activates the Hardware Wakeup function |

Table 34. SPP Port Configuration and Status

| Command | Event | Description |
|---------------------|-----------------------------|--|
| Set Port Config | Set Port Config Confirm | Set port setting for the "virtual" serial port link over the air |
| Get Port Config | Get Port Config Confirm | Read the actual port settings for a "virtual" serial port |
| | Port Config Changed | Notification if port settings were changed from remote device |
| SPP Get Port Status | SPP Get Port Status Confirm | Returns status of DTR, RTS (for the active RF-Comm link) |
| SPP Port Set DTR | SPP Port Set DTR Confirm | Sets the DTR bit on the specified link |

12.0 Command Interface (Continued)

Table 34. SPP Port Configuration and Status (Continued)

| Command | Event | Description |
|------------------------|--------------------------------|--|
| SPP Port Set RTS | SPP Port Set RTS Confirm | Sets the RTS bit on the specified link |
| SPP Port BREAK | SPP Port BREAK | Indicates that the host has detected a break |
| SPP Port Overrun Error | SPP Port Overrun Error Confirm | Used to indicate that the host has detected an overrun error |
| SPP Port Parity Error | SPP Port Parity Error Confirm | Host has detected a parity error |
| SPP Port Framing Error | SPP Port Framing Error Confirm | Host has detected a framing error |
| | SPP Port Status Changed | Indicates that remote device has changed one of the port status bits |

Table 35. Local Bluetooth Settings

| Command | Event | Description |
|--------------------------|----------------------------------|--|
| Read Local Name | Read Local Name Confirm | Read actual friendly name of the device |
| Write Local Name | Write Local Name Confirm | Set the friendly name of the device |
| Read Local BDADDR | Read Local BDADDR Confirm | |
| Change Local BDADDR | Change Local BDADDR Confirm | Note: The BDADDR has to be obtained from the IEEE organization. See http://standards.ieee.org/regauth/oui/ |
| Store Class of Device | Store Class of Device Confirm | |
| Set Scan Mode | Set Scan Mode Confirm | Change mode for discoverability and connectability |
| | Set Scan Mode Indication | Reports end of Automatic limited discoverable mode |
| Get Fixed Pin | Get Fixed Pin Confirm | Reads current PinCode stored within the device |
| Set Fixed Pin | Set Fixed Pin Confirm | Set the local PinCode |
| | PIN request | a PIN code is requested during authentication of an ACL link |
| Get Security Mode | Get Security Mode Confirm | Get actual Security mode |
| Set Security Mode | Set Security Mode Confirm | Configure Security mode for local device (default 2) |
| Remove Pairing | Remove Pairing Confirm | Remove pairing with a remote device |
| List Paired Devices | List of Paired Devices | Get list of paired devices stored in the LMX9830 data memory |
| Set Default Link Timeout | Set Default Link Timeout Confirm | Store default link supervision timeout |
| Get Default Link Timeout | Get Default Link Timeout Confirm | Get stored default link supervision timeout |
| Force Master Role | Force Master Role Confirm | Enables/Disables the request for master role at incoming connections |

Table 36. Local Service Database Configuration

| Command | Event | Description |
|--------------------------|--------------------------------|---|
| Store generic SDP Record | Store SDP Record Confirm | Create a new service record within the service database |
| Enable SDP Record | Enable SDP Record Confirm | Enable or disable SDP records |
| Delete All SDP Records | Delete All SDP Records Confirm | |
| Ports to Open | Ports to Open Confirmed | Specify the RFCOMM Ports to open on startup |

12.0 Command Interface (Continued)

Table 37. Local Hardware Commands

| Command | Event | Description |
|-----------------------------|-------------------------------------|--|
| Set Default Audio Settings | Set Default Audio Settings Confirm | Configure Default Settings for Audio Codec and Air Format, stored in NVS |
| Get Default Audio Settings | Get Default Audio Settings Confirm | Get stored Default Audio Settings |
| Set Event Filter | Set Event Filter Confirm | Configures the reporting level of the command interface |
| Get Event Filter | Get Event Filter Confirm | Get the status of the reporting level |
| Read RSSI | Read RSSI Confirm | Returns an indicator for the incoming signal strength |
| Change UART Speed | Change UART Speed Confirm | Set specific UART speed; needs proper ISEL pin setting |
| Change UART Settings | Change UART Settings Confirm | Change configuration for parity and stop bits |
| Test Mode | Test Mode Confirm | Enable Bluetooth, EMI test, or local loopback |
| Restore Factory Settings | Restore Factory Settings Confirm | |
| Reset | Dongle Ready | Soft reset |
| Firmware Upgrade | | Stops the bluetooth firmware and executes the In-system-programming code |
| Set Clock Frequency | Set Clock Frequency Confirm | Write Clock Frequency setting in the NVS |
| Get Clock Frequency | Get Clock Frequency Confirm | Read Clock Frequency setting from the NVS |
| Set PCM Slave Configuration | Set PCM Slave Configuration Confirm | Write the PCM Slave Configuration in the NVS |
| Write ROM Patch | Write ROM Patch Confirm | Store ROM Patch in the SimplyBlue module |
| Read Memory | Read Memory Confirm | Read from the internal RAM |
| Write Memory | Write Memory Confirm | Write to the internal RAM |
| Read NVS | Read NVS Confirm | Read from the NVS (EEPROM) |
| Write NVS | Write NVS Confirm | Write to the NVS (EEPROM) |

Table 38. Initialization Commands

| Command | Event | Description |
|------------------------|--------------------------------|--|
| Set Clock and Baudrate | Set Clock and Baudrate Confirm | Write Baseband frequency and Baudrate used |
| Enter Bluetooth Mode | Enter Bluetooth Mode Confirm | Request SimplyBlue module to enter BT mode |
| Set Clock and Baudrate | Set Clock and Baudrate Confirm | Write Baseband frequency and Baudrate used |

Table 39. GPIO Control commands

| Command | Event | Description |
|----------------------|-------------------------------|---|
| Set GPIO WPU | Set GPIO WPU Confirm | Enable/Disable weak pull up resistor on GPIOs |
| Get GPIO Input State | Get GPIO Input States Confirm | Read the status of the GPIOs |
| Set GPIO Direction | Set GPIO Direction Confirm | Set the GPIOs direction (Input, Output) |
| Set GPIO Output High | Set GPIO Output High Confirm | Set GPIOs Output to logical High |
| Set GPIO Output Low | Set GPIO Output Low Confirm | Set GPIOs Output to logical Low |

13.0 Usage Scenarios

13.1 SCENARIO 1: POINT-TO-POINT CONNECTION

LMX9830 acts only as slave, no further configuration is required.

Example: Sensor with LMX9830; hand-held device with standard Bluetooth option.

The SPP conformance of the LMX9830 allows any device using the SPP to connect to the LMX9830.

Because of switching to Transparent automatically, the controller has no need for an additional protocol layer; data is sent raw to the other Bluetooth device.

On default, a PinCode is requested to block unallowed targeting.

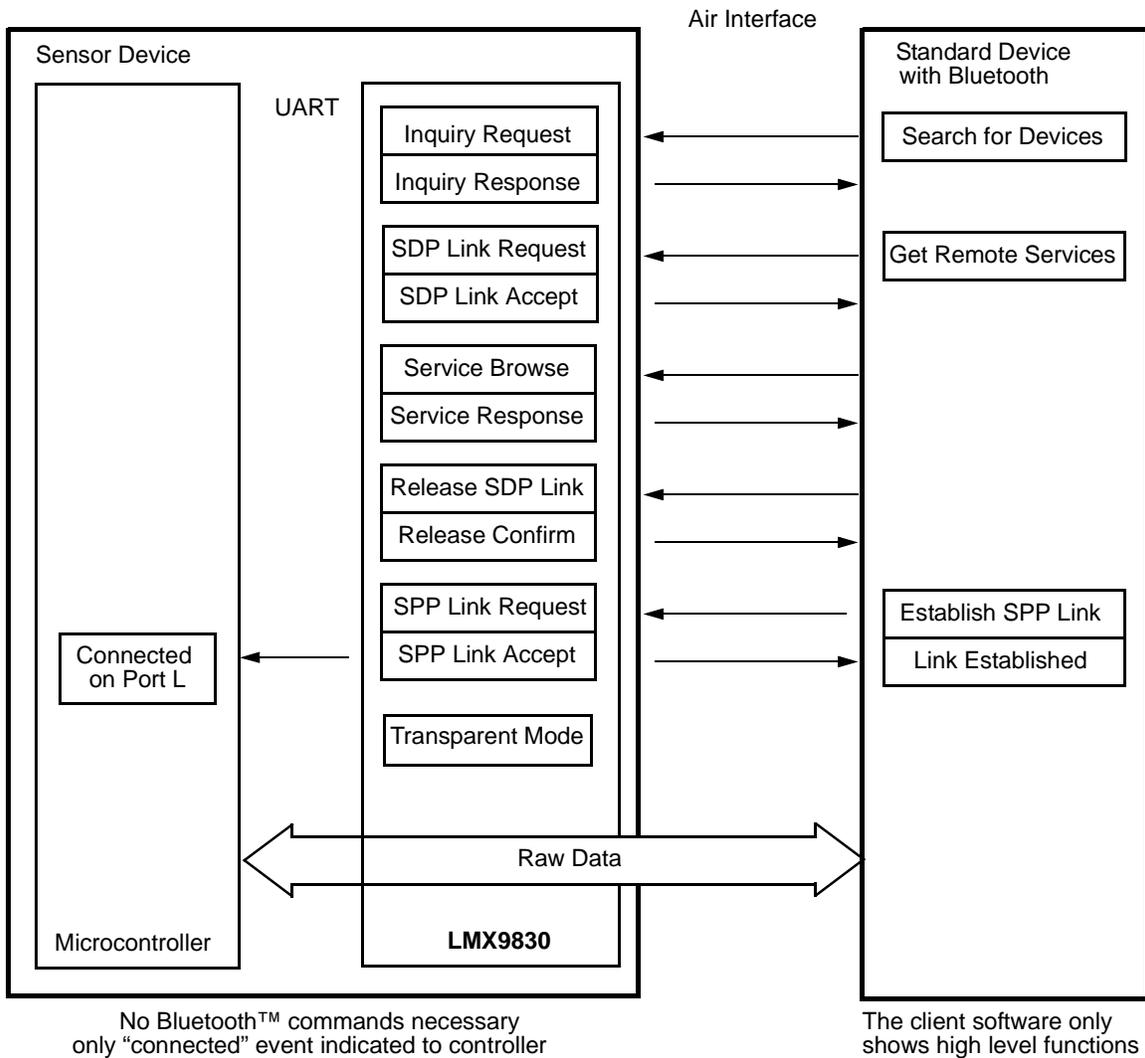


Figure 24. Point-to-Point Connection

13.0 Usage Scenarios (Continued)

13.2 SCENARIO 2: AUTOMATIC POINT-TO-POINT CONNECTION

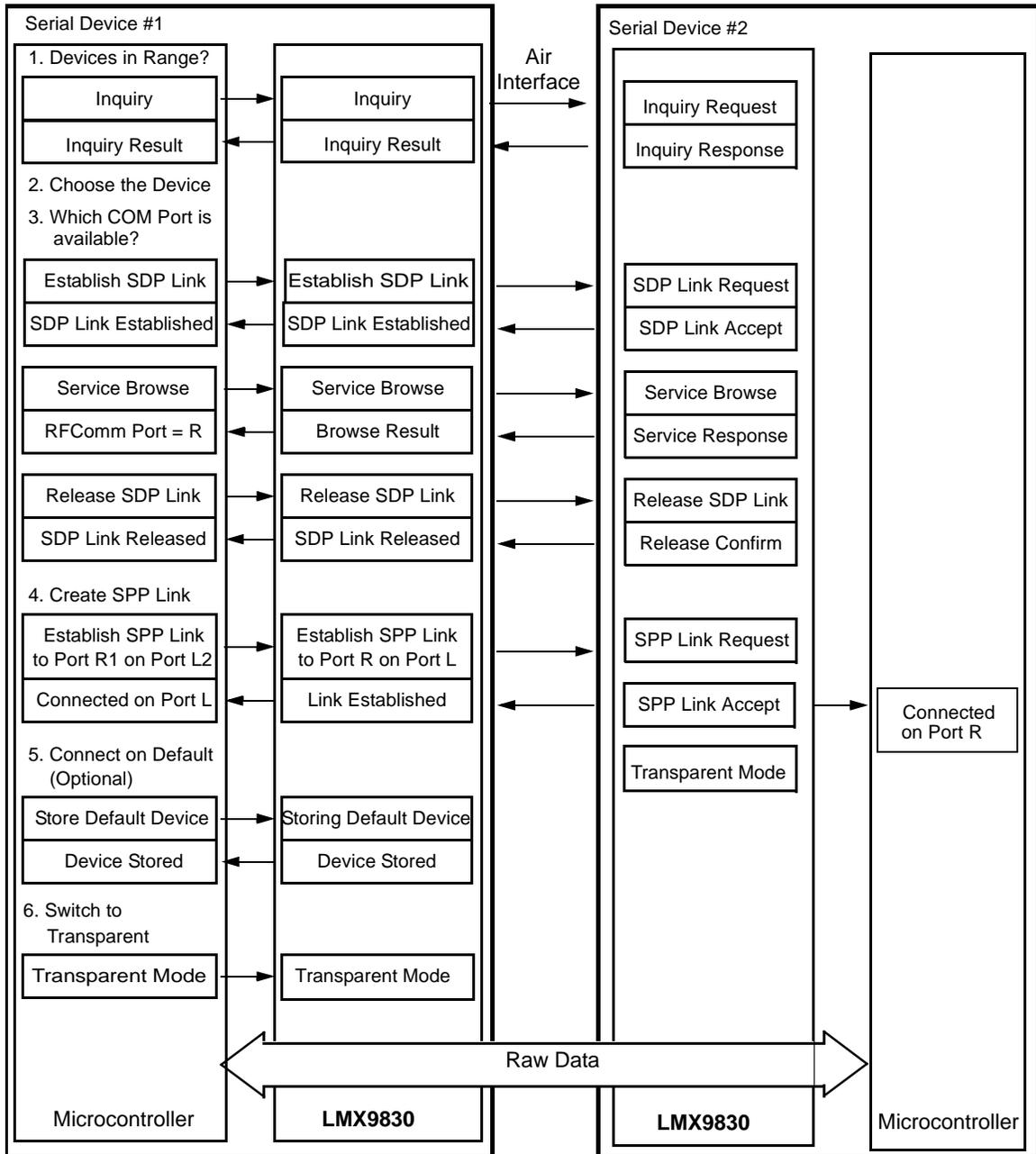
LMX9830 at both sides.

Example: Serial Cable Replacement.

Device #1 controls the link setup with a few commands as described.

If step 5 is executed, the stored default device is connected (step 4) after reset (in Automatic mode only) or by sending the command "Connect to Default Device". The command can be sent to the device at any time.

If step 6 is left out, the microcontroller has to use the command "Send Data" instead of sending data directly to the module.



1. Port R indicates the remote RFCOMM channel to connect to. Usually the result of the SDP request.
2. Port L indicates the Local RFCOMM channel used for that connection.

Figure 25. Automatic Point-to-Point Connection

13.0 Usage Scenarios (Continued)

13.3 SCENARIO 3: POINT-TO-MULTIPOINT CONNECTION

LMX9830 acts as master for several slaves.

Example: Two sensors with LMX9830; one hand-held device with implemented LMX9830.

Serial Devices #2 and #3 establish the link automatically as soon as they are contacted by another device. No controller interaction is necessary for setting up the Bluetooth link. Both switch automatically into Transparent mode. The host sends raw data over the UART.

Serial Device #1 is acting as master for both devices. As the host has to decide to or from which device data is coming from, data must be sent using the "Send data command". If the device receives data from the other devices, it is packaged into an event called "Incoming data event". The event includes the device related port number.

If necessary, a link configuration can be stored as default in the master Serial Device #1 to enable the automatic reconnect after reset, power-up, or by sending the "connect default connection" command.

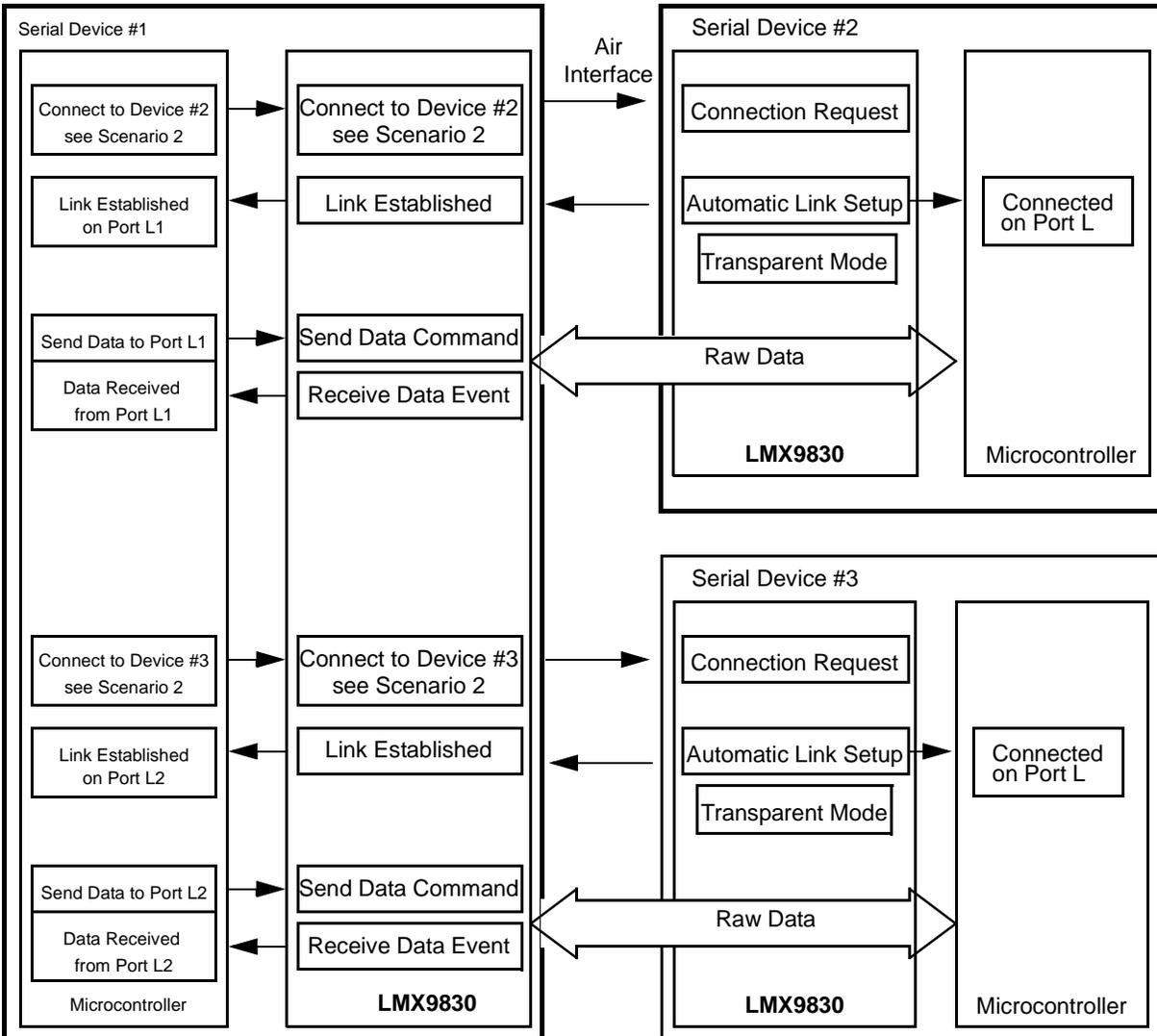
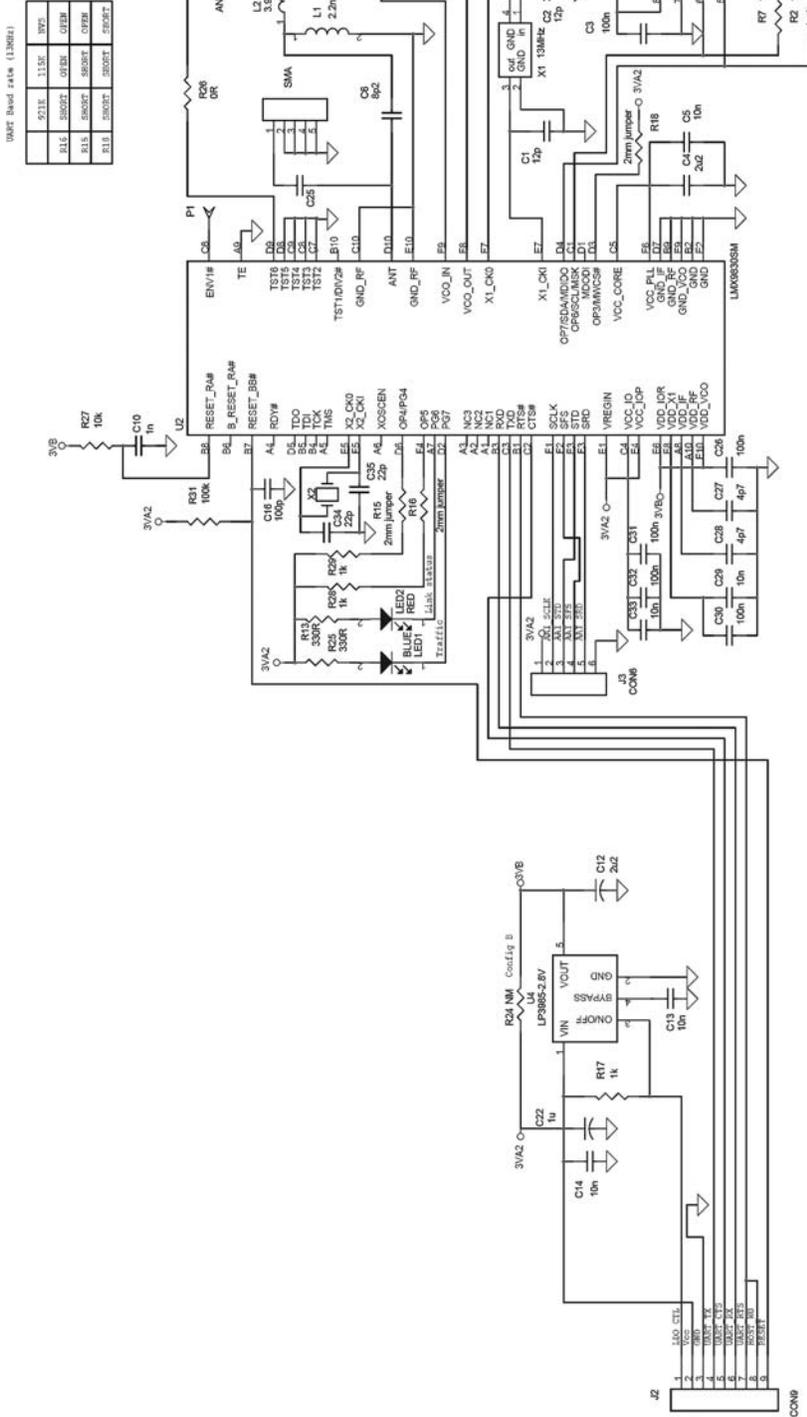


Figure 26. Point-to-Multipoint Connection

14.0 Reference design



Note: For a schematic including an RS232 communication with the host, please refer to the “LMX9830DONGLE designer guide”.

Recommended that a 4 component T-PI pad be used between RF out and antenna input. Allows for versatility in the design such that the match to the antenna maybe improved and/or blocking margin increased by adding a LC filter.

Conf: Bypass 04

15.0 Soldering

The LMX9830 bumps are designed to melt as part of the Surface Mount Assembly (SMA) process. In order to ensure reflow of all solder bumps and maximum solder joint reliability while minimizing damage to the package, recommended reflow profiles should be used.

Table 40, Table 41 and Figure 27 on page 44 provide the soldering details required to properly solder the LMX9830 to standard PCBs. The illustration serves only as a guide and National is not liable if a selected profile does not work.

See IPC/JEDEC J-STD-020C, July 2004 for more information.

Table 40. Soldering Details

| Parameter | Value |
|---------------------------|---|
| PCB Land Pad Diameter | 13 mil |
| PCB Solder Mask Opening | 19 mil |
| PCB Finish (HASL details) | Defined by customer or manufacturing facility |
| Stencil Aperture | 17 mil |
| Stencil Thickness | 5 mil |
| Solder Paste Used | Defined by customer or manufacturing facility |
| Flux Cleaning Process | Defined by customer or manufacturing facility |
| Reflow Profiles | See Figure 27 on page 44 |

Table 41. Classification Reflow Profiles ^{1,2}

| Profile Feature | NOPB Assembly |
|--|----------------------------------|
| Average Ramp-Up Rate ($T_{S_{MAX}}$ to T_p) | 3°C/second maximum |
| Preheat: Temperature Min ($T_{S_{MIN}}$) Temperature Max ($T_{S_{MAX}}$) Time ($t_{S_{MIN}}$ to $t_{S_{MAX}}$) | 150°C 200°C 60–180 seconds |
| Time maintained above: Temperature (T_L) Time (t_L) | 217°C 60–150 seconds |
| Peak/Classification Temperature (T_p) | 260 + 0°C |
| Time within 5°C of actual Peak Temperature (tp) | 20–40 seconds |
| Ramp-Down Rate | 6°C/second maximum |
| Time 25 °C to Peak Temperature | 8 minutes maximum |
| Reflow Profiles | See Figure 27 |

1. See IPC/JEDEC J-STD-020C, July 2004.
2. All temperatures refer to the top side of the package, measured on the package body surface.

15.0 Soldering (Continued)

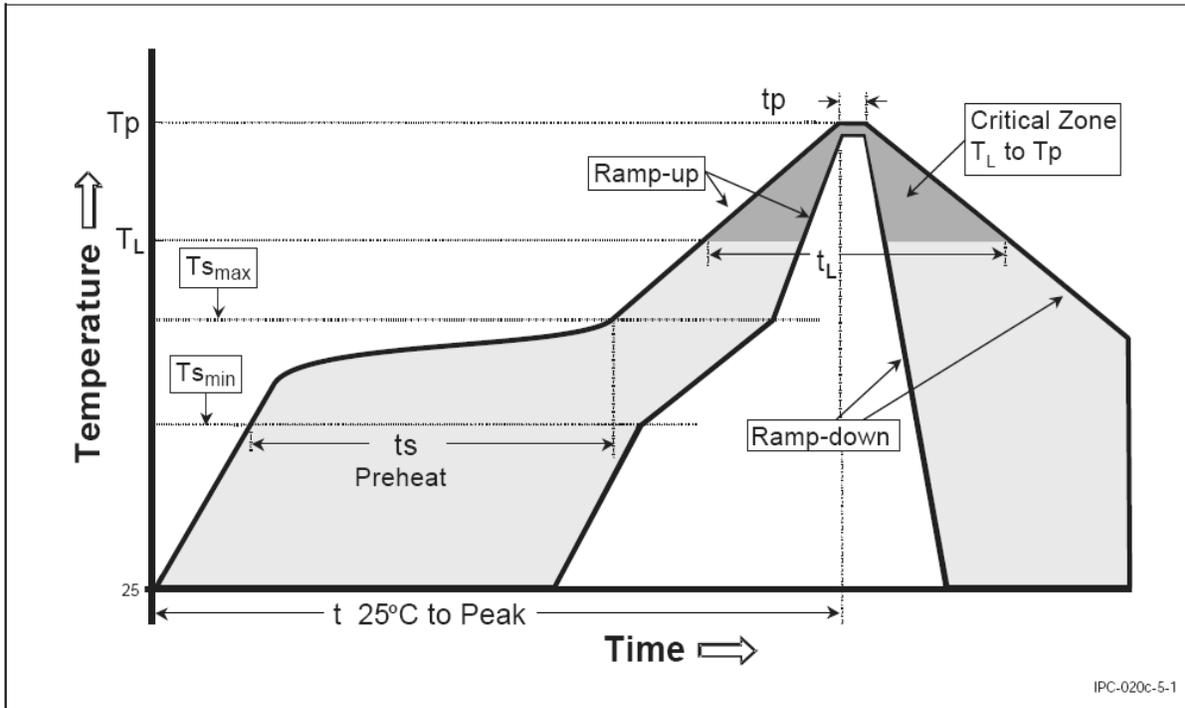


Figure 27. Typical Reflow Profiles

16.0 Datasheet Revision History

This section is a report of the revision/creation process of the datasheet for the LMX9830. Table 42 provides the stages/definitions of the datasheet. Table 43 lists the revision history and Table 44 lists the specific edits to create the current revision.

tion history and Table 44 lists the specific edits to create the current revision.

Table 42. Documentation Status Definitions

| Datasheet Status | Product Status | Definition |
|-------------------------|------------------------|--|
| Advance Information | Formative or in Design | This datasheet contains the design specifications for product development. Specifications may change in any manner without notice. |
| Preliminary | First Production | This datasheet contains preliminary data. Supplementary data will be published at a later date. National Semiconductor Corporation reserves the right to make changes at any time without notice in order to improve design and supply the best possible product.S |
| No Identification Noted | Full production | This datasheet contains final specifications. National Semiconductor Corporation reserves the right to make changes at any time without notice in order to improve design and supply the best possible product. |
| Obsolete | Not in Production | This datasheet contains specifications on a product that has been discontinued by National Semiconductor Corporation. The datasheet is printed for reference information only. |

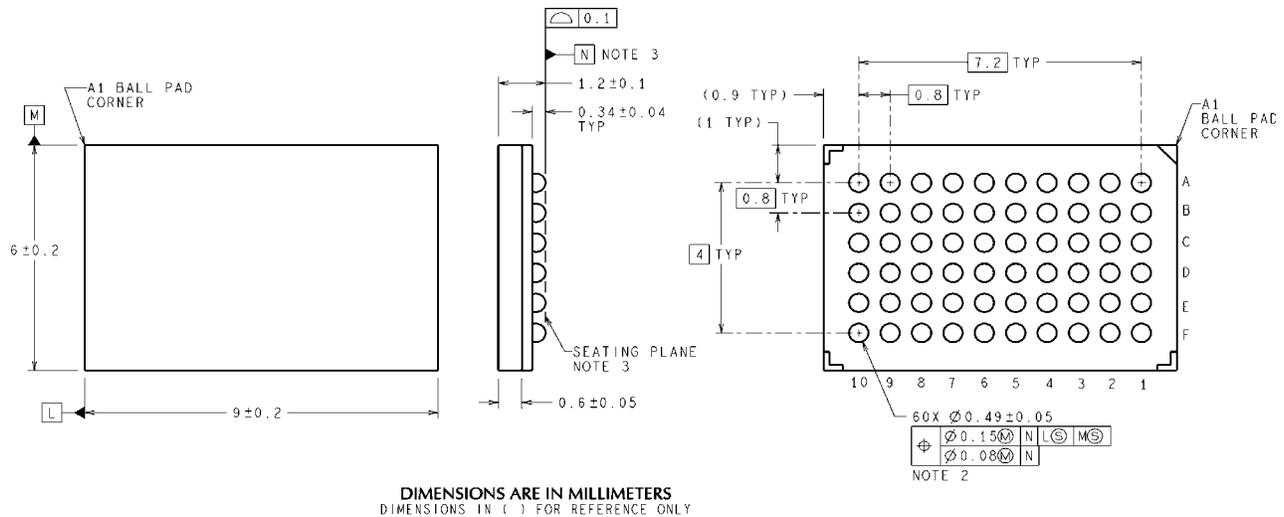
Table 43. Revision History

| Revision # (PDF Date) | Revisions / Comments |
|-----------------------|--|
| 0.1(November 2004) | Advanced datasheet revised to include new radio and additional functionality. Several edits have been made to functional, performance, and electrical details. |
| 0.2(July 2005) | Preliminary version. |
| 0.3(August 2005) | Preliminary version. Updated reference schematic. |
| 1.0 (October 2005) | Final revision. Minor changes to match the final version. |
| 1.0 December 2005 | Section added to digital smart radio and reference design, out of band rejection using front-end components. Also to digital smart radio, section on loop filter design added. |
| 1.1 January 2006 | Updated Sections for front end components and Loop Filter sections and formatted for final release |

Table 44. Edits to Current Revision

| Section | Revisions / Comments |
|---|---|
| All | <ul style="list-style-type: none"> • Misc corrections and typos • Electrical characteristics update. • Put back 32 KHz osc supported. Updated 32 khz schematic in Digital Smart Radio section. • Added quality standard compliance. • Updated BT specification to latest revision. |
| General description | <ul style="list-style-type: none"> • Added RoHS reference. |
| Connection diagram and Pad description | <ul style="list-style-type: none"> • Changed VREGIN to VCC in every definition. |
| General specifications | <ul style="list-style-type: none"> • Updated current consumption to match latest measurements on Mesa 1.1 |
| Autobaudrate description | <ul style="list-style-type: none"> • Corrected RTS to CTS. |
| Power on reset | <ul style="list-style-type: none"> • Timing "Power to Reset" Parameter names changed |
| Schematics | <ul style="list-style-type: none"> • Updated LMX9830 DONGLE reference design schematic. |

17.0 Physical Dimensions inches (millimeters) unless otherwise noted



SLF60A (Rev A)

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