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LMX2492, LMX2492-Q1

SNAS624B - MARCH 2014 - REVISED MAY 2015

LMX2492/LMX2492-Q1 14 GHz Low Noise Fractional N PLL with Ramp/Chirp Generation

Technical

Documents

1 Features

- -227 dBc/Hz Normalized PLL Noise
- 500 MHz 14 GHz Wideband PLL
- 3.15 5.25 V Charge Pump PLL Supply
- Versatile Ramp / Chirp Generation
- 200 MHz Max Phase Detector Frequency
- FSK / PSK Modulation Pin
- Digital Lock Detect
- Single 3.3 V Supply Capability
- Automotive 125°C Q100 Grade 1 Qualification
- Non-Automotive (LMX2492) Option

2 Applications

- Automotive FMCW Radar
- Military Radar
- Microwave Backhaul
- Test and Measurement
- Satellite Communications
- Wireless Infrastructure
- Sampling Clock for High Speed ADC/DAC

4 Simplified Schematic

3 Description

Tools &

Software

The LMX2492/92-Q1 is a low noise 14 GHz wideband delta-sigma fractional N PLL with ramp and chirp generation. It consists of a phase frequency detector, programmable charge pump, and high frequency input for the external VCO. The LMX2492/92-Q1 supports a broad and flexible class of ramping capabilities, including FSK, PSK, and configurable piecewise linear FM modulation profiles of up to 8 segments. It supports fine PLL resolution and fast ramp with up to a 200 MHz phase detector rate. The LMX2492/92-Q1 allows any of its registers to be read back. The LMX2492/92-Q1 can operate with a single 3.3 V supply. Moreover, supporting up to 5.25 V charge pump can eliminate the need of external amplifier, leading to a simpler solution with improved phase noise performance.

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Device Information

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LMX2492-Q1RTW	WQFN (24)	4.00 mm x 4.00 mm
LMX2492RTW	WQFN (24)	4.00 mm x 4.00 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.



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5 Revision History

Cł	hanges from Revision A (June 2014) to Revision B	Page
•	Changed Changed CLK, DATA, and LE to right Input/Output Format	3
•	Changed terminal to pin	3
•	Changed Same specs, but new format tables for storage and ESD Ratings	4
•	Changed TYP to NOM	4
•	Added Added comment for lower voltage operation.	5
•	Changed Fixed Diagram. A14 is hightest bit	6
•	Added note in Applications and Implementation section	27

Changes from Original (March 2014) to Revision A

•	Changed from 35 to 10	. 6
	Changed from 10 to 4	
	Changed from 10 to 4	
	Changed from 25 to 10	
•	Changed from 25 t o10	. 6
		_

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6 Pin Configuration and Functions



Pin Functions

P	N	1/0	DECODIDITION	
NUMBER	NAME	I/O	DESCRIPTION	
0	DAP	GND	Die Attach Pad. Connect to PCB ground plane.	
1	GND	GND	Ground for charge pump.	
2,3	GND	GND	Ground for Fin Buffer	
4,5	Fin Fin*	Input	Complimentary high frequency input pins. Should be AC coupled. If driving single-ended, impedance as seen from Fin and Fin* pins looking outwards from the part should be roughly the same.	
6	Vcc	Supply	Power Supply for Fin Buffer	
7	Vcc	Supply	Supply for On-chip LDOs	
8	Vcc	Supply	Supply for OSCin Buffer	
9	OSCin	Input	Reference Frequency Input	
10	GND/ OSCin*	GND/Input	outwards. However, this may also be grounded as well.	
11	GND	GND	Ground for OSCin Buffer	
12	MOD	Input/Output	Multiplexed Pin for Ramp Triggers, FSK/PSK Modulation, FastLock, Readback, and Diagnostics.	
13	CE	Input	Chip Enable	
14	CLK	Input	Serial Programming Clock.	
15	DATA	Input	Serial Programming Data	
16	LE	Input	Serial Programming Latch Enable	
17	MUXout	Input/Output	Multiplexed Pin for Ramp Triggers, FSK/PSK Modulation, FastLock, Readback, and Diagnostics.	
18	Vcc	Supply	Supply for delta sigma engine.	
19	Vcc	Supply	Supply for general circuitry.	
20	TRIG1	Input/Output	Multiplexed Pin for Ramp Triggers, FSK/PSK Modulation, FastLock, Readback, and Diagnostics.	
21	TRIG2	Input/Output	Multiplexed Pin for Ramp Triggers, FSK/PSK Modulation, FastLock, Readback, and Diagnostics.	
22	Vcp	Supply	Power Supply for the charge pump.	
23	Rset	NC	No connect.	
24	CPout	Output	Charge Pump Output	

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

		MIN	MAX	UNIT
Vcp	Supply voltage for charge pump	Vcc	5.5	V
CPout	Charge pump output pin	-0.3	Vcp	V
Vcc	All Vcc pins	-0.3	3.6	V
Others	All other I/O pins	-0.3	Vcc + 0.3	V
T _{Solder}	Lead temperature (solder 4 seconds)		260	°C
T _{Junction}	Junction temperature		150	°C

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 Storage Conditions

applicable before the DMD is installed in the final product

		MIN	MAX	UNIT
T _{stg}	DMD storage temperature	-65	150	°C
MSL	Moisture sensitivity level		3	n/a

7.3 ESD Ratings

			VALUE	UNIT	
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2500		
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±1500	V	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	DEVICE	MIN	NOM	MAX	UNIT
Vcc	PLL supply voltage		3.15	3.3	3.45	V
Vcp	Charge pump supply voltage		Vcc		5.25	V
T _A	Ambient temperature	LMX2492	-40		85	°C
	Ambient temperature	LMX2492-Q1	-40		125	
TJ		LMX2492	-40		125	- °C
	Junction temperature	LMX2492-Q1	-40		135	

7.5 Thermal Information

	THERMAL METRIC ⁽¹⁾	LMX2492 RTW (WQFN) 24 PINS	UNIT
R_{\thetaJA}	Junction-to-ambient thermal resistance	39.4	
$R_{ extsf{ heta}JC}$	Junction-to-case thermal resistance	7.1	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	20	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.



7.6 Electrical Characteristics

(3.15 V ≤ Vcc ≤ 3.45 V. Vcc ≤Vcp ≤5.25 V. Typical values are at Vcc = Vcp = 3.3 V, 25 °C. -40°C \leq T_A \leq 85 °C for the LMX2492 and -40°C \leq T_A \leq 125 °C for the LMX2492-Q1 ; except as specified.)

	PARAMETER CONDITIONS		NDITIONS	MIN	TYP	MAX	UNIT
			Fpd = 10 MHz		45		
		All Vcc Pins	Fpd = 100 MHz		50		
Icc			Fpd = 200 MHz		55		
	Current Consumption		Kpd = 0.1 mA		2		mA
		Vcp Pin	Kpd = 1.6 mA		10		
			Kpd = 3.1 mA		19		
lccPD	Current	POWERDOWN			3		
		OSC_DIFFR=0,	Doubler Disabled	10		600	
4	Frequency for OSCin	OSC_DIFFR=0,	Doubler Enabled	10		300	
f _{OSCin}	terminal	OSC_DIFFR=1, Doubler Disabled		10		1200	MHz
		OSC_DIFFR=1, Doubler Enabled		10		600	
VOSCin	Voltage for OSCin Pin ⁽¹⁾	LMX2492-Q1 Version Only Single Ended XO 30 MHz $\leq f_{OSCin} \leq 100 \text{ MHz}$		0.24		Vcc-0.5	Vpp
		All Other Cases		0.5		Vcc-0.5	
f _{Fin}	Frequency for FinPin ⁽²⁾			500		14000	MHz
P _{Fin}	Power for Fin Pin	Single-Ended Op	peration	-5		5	dBm
f _{PD}	Phase Detector Frequency					200	MHz
PN1Hz	PLL Figure of Merit ⁽³⁾				-227		dBc/Hz
PN10kHz	Normalized PLL 1/f Noise ⁽³⁾	Normalized to 10 kHz offset for a 1 GHz carrier.			-120		dBc/Hz
I _{CPout} TRI	Charge Pump Leakage Tri- state Leakage					10	nA
I _{CPout} MM	Charge Pump Mismatch ⁽⁴⁾	V _{CPout} = Vcp / 2			5 %		
			CPG=1X		0.1		
CPout	Charge Pump Current	V _{CPout} = Vcp / 2					mA
			CPG=31X		3.1		

(1) For optimal phase noise performance, a slew rate of at least 3 V/ns is recommended

Tested to 13.5 GHz, Guaranteed to 14 GHz by characterization (2)

(3) PLL Noise Metrics are measured with a clean OSCin signal with a high slew rate using a wide loop bandwidth. The noise metrics model the PLL noise for an infinite loop bandwidth as: $PLL_Total = 10 \times log(10^{PLL_Flat/10} + 10^{PLL_Flicker(Offset)/10})$

 $\label{eq:pll_fight} \begin{array}{l} \mathsf{PLL}_{\mathsf{Flat}} = \mathsf{PN1Hz} + 20 \times \mathsf{log}(\mathsf{N}) + 10 \times \mathsf{log}(\mathsf{Fpd/1Hz}) \\ \mathsf{PLL}_{\mathsf{Flicker}} = \mathsf{PN10kHz} + 10 \times \mathsf{log}(\mathsf{Offset/10kHz}) + 20 \times \mathsf{log}(\mathsf{Fvco/1GHz}) \end{array}$

Charge pump mismatch varies as a function of charge pump voltage. Consult typical performance characteristics to see this variation. (4)

Electrical Characteristics (continued)

 $(3.15 \text{ V} \le \text{Vcc} \le 3.45 \text{ V}. \text{ Vcc} \le \text{Vcp} \le 5.25 \text{ V}. \text{ Typical values are at Vcc} = \text{Vcp} = 3.3 \text{ V}, 25 ^{\circ}\text{C}.$ -40°C $\le \text{T}_{A} \le 85 ^{\circ}\text{C}$ for the LMX2492 and -40°C $\le \text{T}_{A} \le 125 ^{\circ}\text{C}$ for the LMX2492-Q1 ; except as specified.)

	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
LOGIC OU	TPUT TERMINALS (MUXout,T	RIG1,TRIG2,MOD)				
V _{OH}	Output High Voltage		0.8 x Vcc	Vcc		V
V _{OL}	Output Low Voltage			0	0.2 x Vcc	V
LOGIC INF	UT TERMINALS (CE,CLK,DA	FA,LE,MUXout,TRIG1,TRIG2,MOD)				
VIH	Input High Voltage		1.4		Vcc	V
V _{IL}	Input Low Voltage		0		0.6	V
I _{IH}	Input Leakage		-5	1	5	uA
T _{CE} LOW	Chip enable Low Time		5			us
T _{CE} HIGH	Chip enable High Time		5			us

7.7 Timing Requirements, Programming Interface (CLK, DATA, LE)

		MIN	NOM	MAX	UNIT
T _{CE}	Clock To LE Low Time	10			ns
T _{CS}	Data to Clock Setup Time	4			ns
T _{CH}	Data to Clock Hold Time	4			ns
T _{CWH}	Clock Pulse Width High	10			ns
T _{CWL}	Clock Pulse Width Low	10			ns
T _{CES}	Enable to Clock Setup Time	10			ns
T _{EWH}	Enable Pulse Width High	10			ns



Figure 1. Serial Data Input Timing

There are several other considerations for programming:

- The DATA is clocked into a shift register on each rising edge of the CLK signal. On the rising edge of the LE signal, the data is sent from the shift register to an actual counter.
- If no LE signal is given after the last data bit and the clock is kept toggling, then these bits will be read into the next lower register. This eliminates the need to send the address each time.
- A slew rate of at least 30 V/us is recommended for the CLK, DATA, and LE signals
- Timing specifications also apply to readback. Readback can be done through the MUXout, TRIG1, TRIG2, or MOD terminals.



7.8 Typical Characteristics





Figure 4. Fin Input Sensitivity

8



Theoretical 1/f Theoretical Flat Theoretical Model

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8 Detailed Description

8.1 Overview

The LMX2492/92-Q1 is a microwave PLL, consisting of a reference input and divider, high frequency input and divider, charge pump, ramp generator, and other digital logic. The Vcc power supply pins run at a nominal 3.3 volts, while the charge pump supply pin, Vcp, operates anywhere from Vcc to 5 volts. The device is designed to operate with an external loop filter and VCO. Modulation is achieved by manipulating the MASH engine.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 OSCin Input

The reference can be applied in several ways. If using a differential input, this should be terminated differentially with a 100 ohm resistance and AC coupled to the OSCin and GND/OSCin* terminals. If driving this single-ended, then the GND/OSCin* terminal may be grounded, although better performance is attained by connecting the GND/OSCin* terminal through a series resistance and capacitance to ground to match the OSCin terminal impedance.

8.3.2 OSCin Doubler

The OSCin doubler allows the input signal to the OSCin to be doubled in order to have higher phase detector frequencies. This works by clocking on both the rising and falling edges of the input signal, so it therefore requires a 50% input duty cycle.

8.3.3 R Divider

The R counter is 16 bits divides the OSCin signal from 1 to 65535. If DIFF_R = 0, then any value can be chosen in this range. If DIFF_R=1, then the divide is restricted to 2,4,8, and 16, but allows for higher OSCin frequencies.

8.3.4 PLL N Divider

The 16 bit N divider divides the signal at the Fin terminal down to the phase detector frequency. It contains a 4/5 prescaler that creates minimum divide restrictions, but allows the N value to increment in values of one.

Modulator Order	Minimum N Divide
Integer Mode, 1st Order Modulator	16
2nd Order Modulator	17
3rd Order Modulator	19
4th Order Modulator	25

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8.3.5 Fractional Circuitry

The fractional circuitry controls the N divider with delta sigma modulation that supports a programmable first, second, third, and fourth order modulator. The fractional denominator is a fully programmable 24-bit denominator that can support any value from 1,2,..., 2^{24} , with the exception when the device is running one of the ramps, and in this case it is a fixed size of 2^{24} .

8.3.6 PLL Phase Detector and Charge Pump

The phase detector compares the outputs of the R and N dividers and generates a correction voltage corresponding to the phase error. This voltage is converted to a correction current by the charge pump. The phase detector frequency, f_{PD} , can be calculated as follows: $f_{PD} = f_{OSCin} \times OSC_2X / R$.

The charge pump supply voltage on this device, Vcp, can be either run at the Vcc voltage, or up to 5.25 volts in order to get higher tuning voltages to present to the VCO.

8.3.7 External Loop Filter

The loop filter is external to the device and is application specific. Texas Instruments website has details on this at www.ti.com.

8.3.8 Fastlock and Cycle Slip Reduction

The FastlockTM and Cycle Slip Reduction features can be used to improved lock time. When the frequency is changed, a timeout counter can be used to engage these features for a prescribed number of phase detector cycles. During this time that the timeout counter is counting down, the device can be used to pull a terminal from high impedance to ground switch in an extra resistor (R2pLF), change the charge pump current (FL_CPG), or change the phase detector frequency. TRIG2 is recommended for switching the resistor with a setting of TRIG2_MUX = Fastlock (2) and TRIG2_PIN = Inverted/Open Drain (5).



Parameter	Normal Operation	Fastlock Operation		
Charge Pump Gain	CPG	FL_CPG		
Device Pin (TRIG1, TRIG2, MOD, or MUXout)	High Impedance	Grounded		

The resistor and the charge pump current are changed simultaneously so that the phase margin remains the same while the loop bandwidth is by a factor of K as shown in the following table:

Parameter	Symbol	Calculation
Charge Pump Gain in Fastlock	FL_CPG	Typically use the highest value.
Loop Bandwidth Multiplier	К	K=sqrt(FL_CPG/CPG)
External Resistor	R2pLF	R2 / (K-1)



Cycle slip reduction is another method that can also be used to speed up lock time by reducing cycle slipping. Cycle slipping typically occurs when the phase detector frequency exceeds about 100x the loop bandwidth of the PLL. Cycle slip reduction works in a different way than fastlock. To use this, the phase detector frequency is decreased while the charge pump current is simultaneously increased by the same factor. Although the loop bandwidth is unchanged, the ratio of the phase detector frequency to the loop bandwidth is, and this is helpful for cases when the phase detector frequency is high. Because cycle slip reduction changes the phase detector rate, it also impacts other things that are based on the phase detector rate, such as the fastlock timeout-counter and ramping controls.

8.3.9 Lock Detect and Charge Pump Voltage Monitor

The LMX2492/92-Q1 offers two methods to determine if the PLL is in lock, charge pump voltage monitoring and digital lock detect. These features can be used individually or in conjunction to give a reliable indication of when the PLL is in lock. The output of this detection can be routed to the TRIG1, TRIG2, MOD, or MUXout terminals.

8.3.9.1 Charge Pump Voltage Monitor

The charge pump voltage monitor allows the user to set low (CMP_THR_LOW) and high (CMP_THR_HIGH) thresholds for a comparator that monitors the charge pump output voltage.

Vcp	Threshold	Suggested Level
2.2.1/	CPM_THR_LOW = (Vthresh + 0.08) / 0.085	6 for 0.5V limit
3.3 V	CPM_THR_HIGH = (Vthresh - 0.96) / 0.044	42 for 2.8V limit
5.0 V	CPM_THR_LOW = (Vthresh + 0.056) / 0.137	4 for 0.5V limit
5.0 V	CPM_THR_HIGH = (Vthresh -1.23) / 0.071	46 for 4.5V limit

8.3.9.2 Digital Lock Detect

Digital lock detect works by comparing the phase error as presented to the phase detector. If the phase error plus the delay as specified by the PFD_DLY bit is outside the tolerance as specified by DLD_TOL, then this comparison would be considered to be an error, otherwise passing. The DLD_ERR_CNT specifies how may errors are necessary to cause the circuit to consider the PLL to be unlocked. The DLD_PASS_CNT specifies how many passing comparisons are necessary to cause the PLL to be considered to be locked and also resets the count for the errors. The DLD_TOL value should be set to no more than half of a phase detector period plus the PFD_DLY value. The DLD_ERR_CNT and DLD_PASS_CNT values can be decreased to make the circuit more sensitive. If the circuit is too sensitive, then chattering can occur and the DLD_ERR_CNT, DLD_PASS_CNT, or DLD_TOL values should be increased.

Note that if the OSCin signal goes away and there is no noise or self-oscillation at the OSCin pin, then it is possible for the digital lock detect to indicate a locked state when the PLL really is not in lock. If this is a concern, then digital lock detect can be combined with charge pump voltage monitor to detect this situation.

8.3.10 FSK/PSK Modulation

Two level FSK or PSK modulation can be created whenever a trigger event, as defined by the FSK_TRIG field is detected. This trigger can be defined as a transition on a terminal (TRIG1, TRIG2, MOD, or MUXout) or done purely in software. The RAMP_PM_EN bit defines the modulation to be either FSK or PSK and the FSK_DEV register determines the amount of the deviation. Remember that the FSK_DEV[32:0] field is programmed as the 2's complement of the actual desired FSK_DEV value. This modulation can be added to the modulation created from the ramping functions as well.

RAMP_PM_EN	Modulation Type	Deviation
0	2 Level FSK	Fpd × FSK_DEV / 2 ²⁴
1	2 Level PSK	360° × FSK_DEV / 2 ²⁴

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8.3.11 Ramping Functions

The LMX2492/92-Q1 supports a broad and flexible class of FMCW modulation formed by up to 8 linear ramps. When the ramping function is running, the denominator is fixed to a forced value of $2^{24} = 16777216$. The waveform always starts at RAMP0 when the LSB of the PLL_N (R16) is written to. After it is set up, it will start at the initial frequency and have piecewise linear frequency modulation that deviates from this initial frequency as specified by the modulation. Each of the eight ramps can be individually programmed. Various settings are as follows

Ramp Characteristic	Programming Field Name	Description
Ramp Length	RAMPx_LEN RAMPx_DLY	The user programs the length of the ramp in phase detector cycles. If RAMPx_DLY=1, then each count of RAMPx_LEN is actually two phase detector cycles.
Ramp Slope	RAMPx_LEN RAMPx_DLY RAMPx_INC	The user does not directly program slope of the line, but rather this is done by defining how long the ramp is and how much the fractional numerator is increased per phase detector cycle. The value for RAMPx_INC is calculated by taking the total expected increase in the frequency, expressed in terms of how much the fractional numerator increases, and dividing it by RAMPx_LEN. The value programmed into RAMPx_INC is actually the two's complement of the desired mathematical value.
Trigger for Next Ramp	RAMPx_NEXT_TRIG	The event that triggers the next ramp can be defined to be the ramp finishing or can wait for a trigger as defined by TRIG A, TRIG B, or TRIG C.
Next Ramp	RAMPx_NEXT	This sets the ramp that follows. Waveforms are constructed by defining a chain ramp segments. To make the waveform repeat, make RAMPx_NEXT point to the first ramp in the pattern.
Ramp Fastlock	RAMPx_FL	This allows the ramp to use a different charge pump current or use Fastlock
Ramp Flags	RAMPx_FLAG	This allows the ramp to set a flag that can be routed to external terminals to trigger other devices.

8.3.11.1 Ramp Count

If it is desired that the ramping waveform keep repeating, then all that is needed is to make the RAMPx_NEXT of the final ramp equal to the first ramp. This will run until the RAMP_EN bit is set to zero. If this is not desired, then one can use the RAMP_COUNT to specify how may times the specified pattern is to repeat.

8.3.11.2 Ramp Comparators and Ramp Limits

The ramp comparators and ramp limits use programable thresholds to allow the device to detect whenever the modulated waveform frequency crosses a limit as set by the user. The difference between these is that comparators set a flag to alert the user while a ramp limits prevent the frequency from going beyond the prescribed threshold. In either case, these thresholds are expressed by programming the Extended_Fractional_Numerator.

Extended_Fractional_Numerator = Fractional_Numerator + $(N-N^*)$ × 2^{24} In the above, N is the PLL feedback value without ramping and N* is the instantaneous value during ramping. The actual value programmed is the 2's complement of Extended_Fractional_Numerator.

Туре	Programming Bit	Threshold
Domo Limito	RAMP_LIMIT_LOW	Lower Limit
Ramp Limits	RAMP_LIMIT_HIGH	Upper Limit
Ramp Comparators	RAMP_CMP0 RAMP_CMP1	For the ramp comparators, if the ramp is increasing and exceeds the value as specified by RAMP_CMPx, then the flag will go high, otherwise it is low. If the ramp is decreasing and goes below the value as specified by RAMP_CMPx, then the flag will go high, otherwise it will be low.

8.3.12 Power on Reset (POR)

The power on reset circuitry sets all the registers to a default state when the device is powered up. This same reset can be done by programming SWRST=1. In the programming section, the power on reset state is given for all the programmable fields.

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8.4 Device Functional Modes

The two primary ways to use the LMX2492/92-Q1 are to run it to generate a set of frequencies

8.4.1 Continuous Frequency Generator

In this mode, the LMX2492/92-Q1 generates a single frequency that only changes when the N divider is programmed to a new value. In this mode, the RAMP_EN bit is set to 0 and the ramping controls are not used. The fractional denominator can be programmed to any value from 1 to 16777216. In this kind of application, the PLL is tuned to different channels, but at each channel, the goal is to generate a stable fixed frequency.

8.4.1.1 Integer Mode Operation

In integer mode operation, the VCO frequency needs to be an integer multiple of the phase detector frequency. This can be the case when the output frequency or frequencies are nicely related to the input frequency. As a rule of thumb, if this an be done with a phase detector of as high as the lesser of 10 MHz or the OSCin frequency, then this makes sense. To operate the device in integer mode, disable the fractional circuitry by programming the fractional order (FRAC_ORDER), dithering (FRAC_DITH), and numerator (FRAC_NUM) to zero.

8.4.1.2 Fractional Mode Operation

In fractional mode, the output frequency does not need to be an integer multiple of the phase detector frequency. This makes sense when the channel spacing is more narrow or the input and output frequencies are not nicely related. There are several programmable controls for this such as the modulator order, fractional dithering, fractional numerator, and fractional denominator. There are many trade-offs with choosing these, but here are some guidelines

Parameter	Field Name	How to Choose
Fractional Numerator and Denominator	FRAC_NUM FRAC_DEN	The first step is to find the fractional denominator. To do this, find the frequency that divides the phase detector frequency by the channel spacing. For instance, if the output ranges from 5000 to 5050 in 5 MHz steps and the phase detector is 100 MHz, then the fractional denominator is 100 MHz/5 = 20. So for a an output of 5015 MHz, the N divider would be 50 + 3/20. In this case, the fractional numerator is 3 and the fractional denominator is 20. Sometimes when dithering is used, it makes sense to express this as a larger equivalent fraction. Note that if ramping is active, the fractional denominator is forced to 2^{24} .
Fractional Order	FRAC_ORDER	There are many trade-offs, but in general try either the 2nd or 3rd order modulator as starting points. The 3rd order modulator may give lower main spurs, but may generate others. Also if dithering is involved, it can generate phase noise.
Dithering	FRAC_DITH	Dithering can reduce some fractional spurs, but add noise. Consult application note AN-1879 for more details on this.

8.4.2 Modulated Waveform Generator

In this mode, the device can generate a broad class of frequency sweeping waveforms. The user can specify up to 8 linear segments in order to generate these waveforms. When the ramping function is running, the denominator is fixed to a forced value of $2^{24} = 16777216$

In addition to the ramping functions, there is also the capability to use a terminal to add phase or frequency modulation that can be done by itself or added on top of the waveforms created by the ramp generation functions.



8.5 Programming

8.5.1 Loading Registers

The device is programmed using several 24 bit registers. The first 16 bits of the register are the address, followed by the next 8 bits of data. The user has the option to pull the LE terminal high after this data, or keep sending data and it will apply this data to the next lower register. So instead of sending three registers of 24 bits each, one could send a single 40 bit register with the 16 bits of address and 24 bits of data. For that matter, the entire device could be programmed as a single register if desired.

8.6 Register Map

Registers are programmed in REVERSE order from highest to lowest. Registers NOT shown in this table or marked as reserved can be written as all 0's unless otherwise stated. The POR value is the power on reset value that is assigned when the device is powered up or the SWRST bit is asserted.

Reg	gister	D7	D6	D5	D4	D3	D2	D1	D0	POR
0	0	0	0	0	1	1	0	0	0	0x18
1	0x1				Rese	erved				0x00
2	0x2	0	0	0	0	0	SWRST	POWERD	OWN[1:0]	0x00
3-15	0x3 - 0xF				Rese	erved				-
16	0x10				PLL_	N[7:0]				0x64
17	0x11				PLL_N	N[15:8]				0x00
18	0x12	0	FR	AC_ORDER[2:0]	FRAC_DI	THER[1:0]	PLL_N	l[17:16]	0x00
19	0x13				FRAC_N	NUM[7:0]				0x00
20	0x14		FRAC_NUM[15:8]						0x00	
21	0x15		FRAC_NUM[23:16]						0x00	
22	0x16		FRAC_DEN[7:0]						0x00	
23	0x17				FRAC_D	EN[15:8]				0x00
24	0x18		FRAC_DEN[23:16]						0x00	
25	0x19				PLL_	R[7:0]				0x04
26	0x1A				PLL_F	R[15:8]				0x00
27	0x1B	0	FL_CS	SR[1:0]	PFD_D	LY[1:0]	PLL_R_ DIFF	0	OSC_2X	0x08
28	0x1C	0	0	CPPOL			CPG[4:0]			0x00
29	0x1D	I	FL_TOC[10:8	8]			FL_CPG[4:0]			0x00
30	0x1E	0	CPM_ FLAGL			CPM_THR	_LOW[5:0]			0x0a
31	0x1F	0	CPM_ FLAGH			CPM_THR	_HIGH[5:0]			0x32
32	0x20				FL_TC	DC[7:0]				0x00
33	0x21				DLD_PASS	S_CNT[7:0]				0x0f
34	0x22	I	DLD_TOL[2:0)]		DLD	_ERR_CNTR	R[4:0]		0x00
35	0x23	MOD_ MUX[5]	1	MUXout _MUX[5]	TRIG2 _MUX[5]	TRIG1 _MUX[5]	0	0	1	0x41
36	0x24		TRIG1_MUX[4:0] TRIG1_PIN[2:0]						0]	0x08
37	0x25		TRIG2_MUX[4:0] TRIG2_PIN[2:0]						0]	0x10
38	0x26		Ν	IOD_MUX[4:	0]		Ν	MOD_PIN[2:0	0]	0x18
39	0x27		MU	JXout_MUX[4	4:0]		М	UXout_PIN[2	::0]	0x38
40-57	0x28-0x39				Rese	erved				-

 Table 1. Register Map



Register Map (continued)

Re	gister	D7	D6	D5	D4	D3	D2	D1	D0	POR
58	0x3A	F	RAMP_TF	RIG_A[3:0]		0	RAMP_ PM_EN	RAMP_ CLK	RAMP_EN	0x00
59	0x3B	F	RAMP_TF	RIG_C[3:0]			RAMP_TI	RIG_B[3:0]		0x00
60	0x3C				RAMP_C	MP0[7:0]				0x00
61	0x3D				RAMP_C	MP0[15:8]				0x00
62	0x3E				RAMP_CM	/IP0[23:16]				0x00
63	0x3F				RAMP_CM	/IP0[31:24]				0x00
64	0x40				RAMP_CM	P0_EN[7:0]				0x00
65	0x41				RAMP_C	MP1[7:0]				0x00
66	0x42				RAMP_C	MP1[15:8]				0x00
67	0x43				RAMP_CM	/IP1[23:16]				0x00
68	0x44				RAMP_CM	/IP1[31:24]				0x00
69	0x45		RAMP_CMP1_EN[7:0]						0x00	
70	0x46	0	0 FSK_TRIG[1:0] RAMP LIMH[3			RAMP_ LIML[32]	FSK_ DEV[32]	RAMP_ CMP1[32]	RAMP_ CMP0[32]	0x08
71	0x47				FSK_D	EV[7:0]				0x00
72	0x48				FSK_DI	EV[15:8]				0x00
73	0x49				FSK_DE	V[23:16]				0x00
74	0x4A				FSK_DE	V[31:24]				0x00
75	0x4B				RAMP_LIM	T_LOW[7:0]				0x00
76	0x4C				RAMP_LIMI	F_LOW[15:8]				0x00
77	0x4D				RAMP_LIMIT	_LOW[23:16	;]			0x00
78	0x4E				RAMP_LIMIT	_LOW[31:24	·]			0x00
79	0x4F				RAMP_LIMI	T_HIGH[7:0]				Oxff
80	0x50		RAMP_LIMIT_HIGH[15:8]						Oxff	
81	0x51		RAMP_LIMIT_HIGH[23:16]						0xff	
82	0x52		RAMP_LIMIT_HIGH[31:24]					Oxff		
83	0x53				RAMP_C	DUNT[7:0]				0x00
84	0x54	RAMP_TRIG_I	NC[1:0]	RAMP_ AUTO		RAM	/IP_COUNT[12:8]		0x00
85	0x55				Rese	erved				0x00

Table 1. Register Map (continued)



Register Map (continued)

Reg	gister	D7	D6	D5	D4	D3	D2	D1	D0	POR
86	0x56				RAMP0_I	NC[7:0]				0x00
87	0x57		RAMP0_INC[15:8]							
88	0x58		RAMP0_INC[23:16]							0x00
89	0x59	RAMP0_ DLY	RAMP0_ FL			RAMP0_	INC[29:24]			0x00
90	0x5A				RAMP0_L	EN[7:0]				0x00
91	0x5B				RAMP0_LI	EN[15:8]				0x00
92	0x5C	RA	MP0_NEXT[2:0]	RAMF NEXT_TF		RAMP0_ RST	RAMP0_	_FLAG[1:0]	0x00
93	0x5D				RAMP1_I	NC[7:0]				0x00
94	0x5E				RAMP1_IN	IC[15:8]				0x00
95	0x5F				RAMP1_IN	C[23:16]				0x00
96	0x60	RAMP1_ DLY	RAMP1_ FL			RAMP1_	INC[29:24]			0x00
97	0x61		RAMP1_LEN[7:0]							0x00
98	0x62		RAMP1_LEN[15:8]							0x00
99	0x63	RA	MP1_NEXT[2:0]	RAMF NEXT_TF		RAMP1_ RST	RAMP1_	_FLAG[1:0]	0x00
100	0x64				RAMP2_I	NC[7:0]				0x00
101	0x65				RAMP2_IN	IC[15:8]				0x00
102	0x66				RAMP2_IN	C[23:16]				0x00
103	0x67	RAMP2 DLY	RAMP2_ FL		RAMP2_INC[29:24]					0x00
104	0x68				RAMP2_L	EN[7:0]				0x00
105	0x69				RAMP2_LI	EN[15:8]				0x00
106	0x6A	RA	MP2_NEXT[2:0]	RAMF NEXT_TF		RAMP2_ RST	RAMP2_	_FLAG[1:0]	0x00
107	0x6B				RAMP3_I	NC[7:0]				0x00
108	0x6C				RAMP3_IN	IC[15:8]				0x00
109	0x6D		RAMP3_INC[23:16]							0x00
110	0x6E	RAMP3_ DLY	RAMP3_ FL			RAMP3_INC[29:24]				0x00
111	0x6F		RAMP3_LEN[7:0]							0x00
112	0x70				RAMP3_LI	EN[15:8]				0x00
113	0x71	RA	MP3_NEXT[2:0]	RAMF NEXT_TR		RAMP3_ RST	RAMP3_	_FLAG[1:0]	0x00

Table 1. Register Map (continued)



Register Map (continued)

Regi	ster	D7	D6	D5	D4	D3	D2	D1	D0	POR
114	0x72				RAMP4_	INC[7:0]				0x00
115	0x73		RAMP4_INC[15:8]							
116	0x74				RAMP4_II	NC[23:16]				0x00
117	0x75	RAMP4_ DLY								0x00
118	0x76				RAMP4_	LEN[7:0]				0x00
119	0x77				RAMP4_L	.EN[15:8]				0x00
120	0x78	RA	RAMP4_NEXT[2:0]			P4_ RIG[1:0]	RAMP4_ RST	RAMP4_	_FLAG[1:0]	0x00
121	0x79				RAMP5_	INC[7:0]				0x00
122	0x7A				RAMP5_I	NC[15:8]				0x00
123	0x7B				RAMP5_II	NC[23:16]				0x00
124	0x7C	RAMP5_ DLY	RAMP5_ FL			RAMP5_	INC[29:24]			0x00
125	0x7D				RAMP5_	LEN[7:0]				0x00
126	0x7E		RAMP5_LEN[15:8]							0x00
127	0x7F	RA	MP5_NEXT[2:0]	RAM NEXT_T		RAMP5_ RST	RAMP5_	_FLAG[1:0]	0x00
128	0x80		RAMP6_INC[7:0]							0x00
129	0x81				RAMP6_I	NC[15:8]				0x00
130	0x82				RAMP6_II	NC[23:16]				0x00
131	0x83	RAMP6_ DLY	RAMP6_ FL		RAMP6_INC[29:24]					0x00
132	0x84				RAMP6_	LEN[7:0]				0x00
133	0x85				RAMP6_L	.EN[15:8]				0x00
134	0x86	RA	MP6_NEXT[2:0]	RAM NEXT_T		RAMP6_ RST	RAMP6_	_FLAG[1:0]	0x00
135	0x87				RAMP7_	INC[7:0]	·•			0x00
136	0x88				RAMP7_I	NC[15:8]				0x00
137	0x89				RAMP7_II	NC[23:16]				0x00
138	0x8A	RAMP7_ DLY	RAMP7_ FL			RAMP7_	INC[29:24]			0x00
139	0x8B		RAMP7_LEN[7:0]							0x00
140	0x8C				RAMP7_L	.EN[15:8]				0x00
141	0x8D	RA	MP7_NEXT[2:0]	RAM NEXT_T		RAMP7_ RST	RAMP7_	_FLAG[1:0]	0x00
142-32767	0x8E- 0x7fff				Rese	rved				0x00

Table 1. Register Map (continued)

8.7 Register Field Descriptions

The following sections go through all the programmable fields and their states. Additional information is also available in the applications and feature descriptions sections as well. The POR column is the power on reset state that this field assumes if not programmed.

8.7.1 POWERDOWN and Reset Fields

Field	Location	POR	Description and States		
				Value	POWERDOWN State
				0	POWERDOWN, ignore CE
POWERDOWN	R2[1:0]	0	POWERDOWN Control	1	Power Up, ignore CE
[1:0]		Ū		2	Power State Defined by CE terminal state
				3	Reserved
				Value	Reset State
SWRST	R2[2]	0	Software Reset. Setting this bit sets all registers to their POR default values.	0	Normal Operation
	_	L		1	Register Reset

Table 2. POWERDOWN and Reset Fields



8.7.2 Dividers and Fractional Controls

Field	Location	POR	Description	n and States	
PLL_N [17:0]	R18[1] to R16[0]	16	Feedback N counter Divide value. Minimum the register R16 begins any ramp execution		
				Value	Dither
				0	Weak
FRAC_ DITHER [1:0]	R18[3:2]	0	Dither used by the fractional modulator	1	Medium
[1.0]				2	Strong
				3	Disabled
				Value	Modulator Order
				0	Integer Mode
				1	1st Order Modulator
FRAC_ORDER [2:0]	R18[6:4]	0	Fractional Modulator order	2	2nd Order Modulator
[2.0]				3	3rd Order Modulator
				4	4th Order Modulator
				5-7	Reserved
FRAC_NUM [23:0]	R21[7] to R19[0]	0	Fractional Numerator. This value should denominator.	be less that	an or equal to the fractional
FRAC_DEN [23:0]	R24[7] to R22[0]	0	Fractional Denominator. If the RAMP_EN=1, this field is ignored and the denominator i fixed to 2^{24} .		
PLL_R [15:0]	R26[7] to R25[0]	1	Reference Divider value. Selecting 1 will bypass counter.		
	R27[0] 0		0 Enables the Doubler before the Reference divider	Value	Doubler
OSC_2X		0		0	Disabled
				1	Enabled
			0 Enables the Differential R counter. This allows for higher OSCin frequencies, but restricts PLL_R to divides of 2,4,8 or	Value	R Divider
PLL_R _DIFF	R27[2]	0		0	Single-Ended
			16.	1	Differential
				Value	Pulse Width
			Sets the charge pump minimum pulse	0	Reserved
PFD_DLY [1:0]	R27[4:3]	1	width. This could potentially be a trade-off between fractional spurs and phase noise.	1	860 ps
[1.0]			Setting 1 is recommended for general use.	2	1200 ps
				3	1500 ps
				Value	Charge Pump State
				0	Tri-State
CPG	Dool 4 ol			1	100 uA
[4:0]	R28[4:0]	0	Charge pump gain	2	200 uA
				31	3100 uA
			Charge Pump Polarity	Value	Charge Pump Polarity
CPPOL	POL R28[5] 0	0	Positive is for a positive slope VCO	0	Negative
			characteristic, negative otherwise.	1	Positive

Table 3. Dividers and Fractional Controls

STRUMENTS

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8.7.2.1 Speed Up Controls (Cycle Slip Reduction and Fastlock)

Field	Location	POR	Description	and States	
			Cycle Slip Reduction (CSR) reduces the phase detector frequency by multiplying both the R and N counters by the CSR	Value	CSR Value
				0	Disabled
			value while either the FastLock Timer is	1	x 2
FL_CSR	R27[6:5]	0	counting or the RAMPx_FL=1 and the part is ramping. Care must be taken that the R	2	x 4.
[1:0]			and N divides remain inside the range of the counters. Cycle slip reduction is generally not recommended during ramping.	3	Reserved
	R29[4:0]	0		Value	Fastlock Charge Pump Gain
			Charge pump gain only when Fast Lock Timer is counting down or a ramp is running with RAMPx_FL=1	0	Tri-State
FL_ CPG				1	100 uA
[4:0]				2	200 uA
				31	3100 uA
			Fast Lock Timer. This counter starts	Value	Fastlock Timer Value
			counting when the user writes the PLL_N(Register R16). During this time the	0	Disabled
FL_ TOC	R29[7:5]		FL_CPG gain is sent to the charge pump,	1	1 x 32 = 32
[10:0]	and R32[7:0]	0	and the FL_CSR shifts the R and N counters if enabled. When the counter		
	R32[7:0]		terminates, the normal CPG is presented and the CSR undo's the shifts to give a normal PFD frequency.	2047	2047 x 32 = 65504

Table 4. FastLock and Cycle Slip Reduction



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8.8 Lock Detect and Charge Pump Monitoring

Field	Location	POR	Description	n and States	
				Value	Threshold
CPM_THR _LOW		0.04	Charge pump voltage low threshold value.	0	Lowest
_ [5:0] [_]	R30[5:0]	0x0A	When the charge pump voltage is below this threshold, the LD goes low.		
				63	Highest
				Value	Flag Indication
CPM_FLAGL	R30[6]	-	This is a read only bit. Low indicates the charge pump voltage is	0	Charge pump is below CPM_THR_LOW threshold
			below the minimum threshold.	1	Charge pump is above CPM_THR_LOW threshold
				Value	Threshold
CPM_THR _HIGH	D04[5:0]	000	Charge pump voltage high threshold value. When the charge pump voltage is above this threshold, the LD goes low.	0	Lowest
[5:0]	R31[5:0]	0x32			
				63	Highest
	R31[6]	[6] -	This is a read only bit. Charge pump voltage high comparator reading. High indicates the charge pump voltage is above the maximum threshold.	Value	Threshold
CPM_FLAGH				0	Charge pump is below CPM_THR_HIGH threshold
				1	Charge pump is above CPM_THR_HIGH threshold
DLD_ PASS_CNT [7:0]	R33[7:0]	0xff	Digital Lock Detect Filter amount. There mu and less than DLD_ERR edges before the D smaller will speed the detection of lock, but a	LD is conside	ered in lock. Making this number
DLD_ ERR_CNT [4:0]	R34[4:0]	0	Digital Lock Detect error count. This is th DLD_TOL that are allowed before DLD is recommended value is 4.		
				Value	Window and Fpd Frequency
				0	1 ns (Fpd > 130 MHz)
			Digital Lock detect edge window. If both N and R edges are within this window, it is considered a "good" edge. Edges that are	1	1.7 ns (80 MHz , Fpd ≤ 130 MHz)
DLD _TOL	D24[7:5]	0	farther apart in time are considered "error"	2	3 ns (60 MHz , Fpd ≤ 80 MHz)
[2:0]	R34[7:5]	U	edges. Window choice depends on phase	3	6 ns (45 MHz , Fpd ≤ 60 MHz)
			detector frequency, charge pump minimum pulse width, fractional modulator order and the users desired margin.	4	10 ns (30 MHz < Fpd ≤ 45 MHz)
				5	18 ns (Fpd ≤ 30 MHz)
				6 and 7	Reserved

 Table 5. Lock Detect and Charge Pump Monitor

8.9 TRIG1,TRIG2,MOD, and MUXout Pins

Field	Location	POR	Description and States			
				Value	Pin Drive State	
TRIG1_PIN	R36[2:0]	0		0	TRISTATE (default)	
[2:0]	R30[2.0]	0		1	Open Drain Output	
				2	Pullup / Pulldown Output	
TRIG2 _PIN [2:0]	R37[2:0]	0	This is the terminal drive state for the	3	Reserved	
MODPIN [2:0]	R38[2:0]	0	TRIG1, TRIG2, MOD, and MUXout Pins	4	GND	
				5	Inverted Open Drain Output	
MUXoutPIN [2:0]	R39[2:0]	2:0] 0		6	Inverted Pullup / Pulldown Output	
				7	Input	

Table 6. TRIG1, TRIG2, MOD, and MUXout Terminal States



Field	Location	POR	Description	n and States	
				Value	MUX State
				0	GND
				1	Input TRIG1
				2	Input TRIG2
				3	Input MOD
				4	Output TRIG1 after synchronizer
				5	Output TRIG2 after synchronizer
				6	Output MOD after synchronizer
				7	Output Read back
				8	Output CMP0
				9	Output CMP1
				10	Output LD (DLD good AND CPM good)
				11	Output DLD
				12	Output CPMON good
				13	Output CPMON too High
				14	Output CPMON too low
TRIG1_MUX	R36[7:3],		These fields control what signal is muxed to or from the TRIG1,TRIG2, MOD, and	15	Output RAMP LIMIT EXCEEDED
[5:0] TRIG2_MUX	R37.3 R36[7:3],	1	MUXout pins. Some of the abbreviations used are:	16	Output R Divide/2
[5:0]	R35.3	2	COMP0, COMP1: Comparators 0 and 1	17	Output R Divide/4
MOD_MUX [5:0]	R37[7:3], R35.4	3	LD, DLD: Lock Detect, Digital Lock Detect CPM: Charge Pump Monitor	18	Output N Divide/2
MUXout_MUX	R38[7:3],		CPG: Charge Pump Gain	19	Output N Divide/4
[5:0]	R35.7		CPUP: Charge Pump Up Pulse CPDN: Charge Pump Down Pulse	20	Reserved
				21	Reserved
				22	Output CMP0RAMP
				23	Output CMP1RAMP
				24	Reserved
				25	Reserved
				26	Reserved
				27	Reserved
				28	Output Faslock
				29	Output CPG from RAMP
				30	Output Flag0 from RAMP
				31	Output Flag1 from RAMP
				32	Output TRIGA
				33	Output TRIGB
				34	Output TRIGC
				35	Output R Divide
				36	Output CPUP
				37	Output CPDN
				38	Output RAMP_CNT Finished
				39 to 63	Reserved

Table 7. TRIG1, TRIG2, MOD, and MUXout Selections

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8.10 Ramping Functions

Field	Location	POR	Description	and States		
			Enables the RAMP functions. When this bit	Value	Ramp	
RAMP_EN	R58[0]	0	is set, the Fractional Denominator is fixed to 2 ²⁴ . RAMP execution begins at RAMP0	0	Disabled	
			upon the PLL_N[7:0] write. The Ramp should be set up before RAMP_EN is set.	1	Enabled	
			RAMP clock input source. The ramp can	Value	Source	
RAMP_CLK	R58[1]	0	be clocked by either the phase detector clock or the MOD terminal based on this	0	Phase Detector	
			selection.	1	MOD Terminal	
				Value	Modulation Type	
RAMP_PM_EN	R58[2]	0	Phase modulation enable.	0	Frequency Modulation	
				1	Phase Modulation	
				Value	Source	
				0	Never Triggers (default)	
				1	TRIG1 terminal rising edge	
				2	TRIG2 terminal rising edge	
				3	MOD terminal rising edge	
				4	DLD Rising Edge	
RAMP_TRIGA [3:0] RAMP_TRIGB [3:0] RAMP_TRIGC [3:0]	R58[7:4] R59[3:0] R59[7:4]		Trigger A,B, and C Sources	5	CMP0 detected (level)	
				6	RAMPx_CPG Rising edge	
		0		7	RAMPx_FLAG0 Rising edge	
				8	Always Triggered (level)	
				9	TRIG1 terminal falling edge	
				10	TRIG2 terminal falling edge	
				11	MOD terminal falling edge	
				12	DLD Falling Edge	
				13	CMP1 detected (level)	
				14	RAMPx_CPG Falling edge	
				15	RAMPx_FLAG0 Falling edge	
RAMP_CMP0 [32:0]	R70[0], R63[7] to R60[0]	0	Twos compliment of Ramp Comparator 0 va R70.	lue. Be aware	e of that the MSB is in Register	
RAMP_CMP0_EN [7:0]	R64[7:0]	0	Comparator 0 is active during each RAMP or is active in and 0 for ramps it should be igno corresponds to R64[7]			
RAMP_CMP1 [32:0]	R70[1], R68[7] to R65[0]	0	Twos compliment of Ramp Comparator 1 va R70.	lue. Be aware	e of that the MSB is in Register	
RAMP_CMP1_EN [7:0]	R69[7:0]	0	Comparator 1 is active during each RAMP co is active in and 0 for ramps it should be igno corresponds to R64[7].			
				Value	Trigger	
			Deviation trigger source. When this trigger	0	Always Triggered	
FSK_TRIG [1:0]	R76[4] to R75[3]	0	source specified is active, the FSK_DEV	1	Trigger A	
· · · · ·	-[-]		value is applied.	2	Trigger B	
				3	Trigger C	
FSK_DEV [32:0]	R70[2], R74[7] to R71[0]	0	Twos compliment of the deviation value for frequency modulation and phase modulation. This value should be written with 0 when not used. Be aware that the MSB is in Register R70.			

Table 8. Ramping Functions



Ramping Functions (continued)

Field	Location	POR	Description	and States			
RAMP_LIMIT_LOW [32:0]	R70[3], R78[7] to 75[0]	0x000 00000	Twos compliment of the ramp lower limit that the ramp can not go below . The ramp limit occurs before any deviation values are included. Care must be taken if the deviation is used and the ramp limit must be set appropriately. Be aware that the MSB is in Register R70.				
RAMP_LIMIT_HIGH [32:0]	R70[4], R82[7] to 79.0[0]	0xfffffff f	Twos compliment of the ramp higher limit that the ramp can not go above. The ramp limit occurs before any deviation values are included. Care must be taken if the deviation is used and the ramp limit must be set appropriately. Be aware that the MSB is in Register R70.				
RAMP_COUNT [12:0]	R84[4] to R83[0]	0	Number of RAMPs that will be executed before a trigger or ramp enable is brought down. Load zero if this feature is not used. Counter is automatically reset when RAMP_EN goes from 0 to 1.				
		R84[5] 0		Value	Ramp		
RAMP_AUTO	R84[5]		Automatically clear RAMP_EN when	0	RAMP_EN unaffected by ramp counter (default)		
	104[0]		RAMP Count hits terminal count.	1	RAMP_EN automatically brought low when ramp counter terminal counts		
				Value	Source		
RAMP_TRIG_INC			Increment Trigger source for RAMP	0	Increments occur on each ramp transition		
[1:0]	R84[7:6]	0	Counter. To disable ramp counter, load a count value of 0.	1	Increment occurs on trigA		
				2	Increment occurs on trigB		
				3	Increment occurs on trigC		

Table 8. Ramping Functions (continued)

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8.11 Individual Ramp Controls

These bits apply for all eight ramps. For the field names, x can be 0,1,2,3,4,5,6, or 7.

Field	Location	POR	Description and Sta	Description and States						
RAMPx _INC[29:0]	Varies	0	Signed ramp increme	ent.						
				Value	CPG					
RAMPx _FL	Varies	0	This enables fastlock and cycle slip reduction for ramp x.	0	Disabled					
				1	Enabled					
			During this ramp, each increment takes 2 PFD cycles per 0 LEN clock instead of the normal 1 PFD cycle. Slows the	Value	Clocks					
RAMPx _DLY	Varies	0		0	1 PFD clock per RAMP tick.(default)					
_021			ramp by a factor of 2.	1	2 PFD clocks per RAMP tick.					
RAMPx _LEN	Varies	0	Number of PFD clocks (if DLY is 0) to continue to increment 65536 cycles.	RAMP. 1=>1	cycle, 2=>2 etc. Maximum of					
		Varies 0				Value	Flag			
				0	Both FLAG1 and FLAG0 are zero. (default)					
RAMPx _FLAG[1:0]	Varies		General purpose FLAGS sent out of RAMP.	1	FLAG0 is set, FLAG1 is clear					
				2	FLAG0 is clear, FLAG1 is set					
				3	Both FLAG0 and FLAG1 are set.					
			Forces a clear of the ramp accumulator. This is used to	Value	Reset					
RAMP0 RST	Varies	0	erase any accumulator creep that can occur depending on how the ramps are defined. Should be done at the start of a	0	Disabled					
			ramp pattern.	1	Enabled					
				Value	Operation					
RAMPx_			Determines what event is necessary to cause the state	0	RAMPx_LEN					
NEXT _TRIG	Varies	0	RAMPx_LEN counter reaches zero or one of the events for	1	TRIG_A					
[1:0]			Triggers A,B, or C.	2	TRIG_B					
				3	TRIG_C					
RAMP0 _NEXT[2:0]	Varies	0	The next RAMP to execute when the le	The next RAMP to execute when the length counter times out						

Table 9. Individual Ramp Controls

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9 Applications and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The LMX2492/92-Q1 can be used in a broad class of applications such as generating a single frequency for a high frequency clock, generating a tunable range of frequencies, or generating swept waveforms that can be used in applications such as radar.

9.2 Typical Applications

The following schematic is an example of hat could be used in a typical application.



9.2.1 Design Requirements

For these examples, it will be assumed that there is a 100 MHz input signal and the output frequency is between 9400 and 9800 MHz with various modulated waveforms.

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Typical Applications (continued)

Parameter	Symbol	Value	Comments		
Input Frequency	OSCin	100 MHz			
Phase Detector Frequency	Fpd	100 MHz	There are many possibilities, but this choice gives good performance and saves a little current (as shown in the electrical specifications).		
		9400 - 9800 MHz (Simple Chirp)			
VCO Frequency	Fvco	9400 - 9800 (Flattened Ramp)	In the different examples, the VCO frequency is actually changing. However, the same loop filter		
	1,400	9500 - 9625 MHz (Complex Triggered Ramp	design can be used for all three.		
VCO Gain	Кусо	200 MHz/V	This parameter has nothing to do with the LMX2492/92-Q1, but is rather set by the external VCO choice.		

9.2.2 Detailed Design Procedure

The first step is to calculate the reference divider (PLL_R) and feedback divider (PLL_N) values as shown in the table that follows.

Parameter	Symbol and Calculations	Value	Comments
Average VCO Frequency	Fvco _{Avg} = (Fvco _{Max} + Fvco _{Min})/2	9600 MHz	To design a loop filter, one designs for a fixed VCO value, although it is understood that the VCO will tune around. This typical value is usually chosen as the average VCO frequency.
VCO Gain	Кусо	200 MHz/V	This parameter has nothing to do with the LMX2492/92-Q1, but is rather set by the external VCO choice. In this case, it was the RFMD1843 VCO.
PLL Loop Bandwidth	BW	380 kHz	This bandwidth is very wide to allow the VCO frequency to be modulated.
Charge Pump Gain	CPG	3.1 mA	Using the larger gain allows a wider loop bandwidth and gives good phase performance.
R Divider	PLL_R = OSCin / Fpd	1	This value is calculated from previous values.
N Divider	PLL_N = Fvco / Fpd	96	This value is calculated from previous values.
	C1_LF	68 pF	
	C2_LF	3.9 nF	
Loop Filter Components	C3_LF	150 pF	These were calculated by TI design tools.
Componento	R2_LF	390 ohm	
	R3_LF	390 ohm	

Once a loop filter bandwidth is chosen, the external loop filter components of C1_LF, C2_LF, C3_LF, R2_LF, and R3_LF can be calculated with a tool such as the Clock Architect tool available at www.ti.com. It is also highly recommended to look at the EVM instructions. The CodeLoader software is an excellent starting point and example to see how to program this device.



9.2.3 Application Performance Plot - Sawtooth Waveform Example

Using the above design, it can be programmed to generate a sawtooth waveform with the following paramters.

Parameter	Symbol	Value
Ramp Duration	ΔΤ	100 uS
VCO Frequency	Fvco	9400 - 9800 MHz
Range	ΔF	9400 - 9800 MHz = 400 MHz Change



Because we want the ramp length to be 100 us, this works out to 10,000 phase detector cycles which means that RAMP0_LEN=10000. To change 400 MHz, we know that each one of the 10000 steps is 40 kHz. Given the fractional denominator is $2^{24} = 16777216$ and the phase detector frequency is 100 MHz, this implies that the fractional numerator at the end of the ramp will be 6711. However, since this 6711 number is not exact (closer to 6718.8864), the ramp will creep if we do not reset it. Therefore, we set reset the ramp. After the ramp finishes, we want to start with the same ramp, so RAMP0_NEXT is RAMP0. The results of this analysis are in the table below:

RAMP	RAMP0_LEN	RAMP0_INC	RAMP0_NEXT	RAMP0_RST
RAMP0	ΔT × Fpd = 100 us / 100 MHz = 10000	(ΔF / Fpd) /RAMP0_LEN × 2 ²⁴ = (400/100)/10000 ×16777216 = 6711	0	1

The actual measured waveform for this is shown in the following figure. Note that the frequency that was actually measured was from the divide by two output of the VCO and therefore the measured frequency was half of the actual frequency presented to the PLL. This ramping waveform does show some undershoot as the frequency rapidly returns from 9800 MHz (4900 MHz on the plot) to 9400 MHz (4700 MHz on the plot). This undershoot can be mitigated by adding additional ramps.



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RAMP	RAMPx_LEN	RAMPx_INC	RAMPx_NEXT	RAMPx_RST
RAMP0	10 us/ 100 MHz =1000	0	1	1
RAMP1	90 us / 100 MHz =9000	(ΔF / Fpd) /RAMP1_LEN × 2 ²⁴ = (400/100)/9000 ×16777216 = 7457	2	0
RAMP2	10 us/ 100 MHz =1000	0	3	0
RAMP3	90 us / 100 MHz =9000	$\begin{array}{l} (\Delta F \ / \ Fpd) \ /RAMP1_LEN \ \times \ 2^{24} \\ = \ (-400 \ / \ 100) \ / \ 9000 \ \times \ 16777216 = \ -7457 \\ Program in \ 2's \ complement \ of \ - \ 7457 \\ = \ 2^{30} \ - \ 7457 = \ 1073734367 \end{array}$	0	0

Product Folder Links: LMX2492 LMX2492-Q1

9.2.4 Application Performance Plot - Flat Top Triangle Waveform

Now consider pattern as shown below. The ramp is sometimes used because it can better account for Doppler Shift. The purpose for making the top and bottom portions flat is to help reduce the impact of the PLL overshooting and undershooting in order to make the sloped ramped portions more linear.

Parameter	Symbol	Value
	ΔΤΟ	10 uS
Bomp Durotion	ΔΤ1	90 uS
Ramp Duration	ΔΤ2	10 uS
	ΔT3	90 uS
	ΔF0	0
Denne	ΔF1	400 MHz
Range	ΔF2	0
	ΔF3	-400 MHz





The actual measured waveform for this is shown in the following figure. Note that the frequency that was actually measured was from the divide by two output of the VCO and therefore the measured frequency was half of the actual frequency presented to the PLL. The flattened top and bottom of this triangle wave help mitigate the overshoot and undersoot in the frequency.



The actual measured waveform for this is shown in the following figure. Note that the frequency that was actually measured was from the divide by two output of the VCO and therefore the measured frequency was half of the actual frequency presented to the PLL. The flattened top and bottom of this triangle wave help mitigate the overshoot and undersoot in the frequency.

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9.2.5 Applications Performance Plot -- Complex Triggered Ramp

In this example, the modulation is not started until a trigger pulse from the MOD terminal goes high. Assume a phase detector frequency of 100 MHz and we RAMP1 to be 60 us and ramps 2,3,and 4 to be 12 us each. We set the next trigger for RAMP0 to be trigger A and define trigger A to be the MOD terminal. Then we configure as follows:



Figure 6. Complex Triggered Ramp Example

RAMP	RAMPx _LEN	RAMPx_ INC	RAMPx_FL	RAMPx _NEXT	RAMPx_FLAG	RAMPx_ NEXT_TRIG	RAMPx_RS T
RAMP0	1	0	0	1	FLAG0 and FLAG1	TRIG A	1
RAMP1	6000	1073730639	0	2	FLAG0 and FLAG1	TOC Timeout	1
RAMP2	1200	27963	1	3	Disabled	TOC Timeout	0
RAMP3	1200	17476	0	4	FLAG1	TOC Timeout	0
RAMP4	1200	10486	0	1	FLAG0 and FLAG1	TOC Timeout	0

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The actual measured waveform for this is shown in the following figure. Note that the frequency that was actually measured was from the divide by two output of the VCO and therefore the measured frequency was half of the actual frequency presented to the PLL. The flattened top and bottom of this triangle wave help mitigate the overshoot and undersoot in the frequency.



Figure 7. Actual Measurement for Complex Triggered Ramp

10 Power Supply Recommendations

For power supplies, it is recommended to place 100 nF close to each of the power supply pins. If fractional spurs are a large concern, using a ferrite bead to each of these power supply pins can reduce spurs to a small degree.

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11 Layout

11.1 Layout Guidelines

For layout examples, the EVM instructions are the most comprehensive document. In general, the layout guidelines are similar to most other PLL devices. For the high frequency Fin pin, it is recommended to use 0402 components and match the trace width to these pad sizes. Also the same needs to be done on the Fin* pin. If layout is easier to route the signal to Fin* instead of Fin, then this is acceptable as well.

11.2 Layout Example





12 Device and Documentation Support

12.1 Device Support

12.1.1 Development Support

Texas Instruments has several software tools to aid in the development process including CodeLoder for programming, Clock Design Tool for Loop filter design and phase noise/spur simulation, and the Clock Architect. All these tools are available at www.ti.com.

12.2 Documentation Support

12.2.1 Related Documentation

For the avid reader, the following resources are available at www.ti.com.

Application Note 1879 -- Fractional N Frequency Synthesis

PLL Performance, Simulation, and Design -- by Dean Banerjee

12.3 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 10.	Related	Links

PARTS	PRODUCT FOLDER	SAMPLE and BUY	TECHNICAL DOCUMENTS	TOOLS and SOFTWARE	SUPPORT and COMMUNITY
LMX2492	Click here	Click here	Click here	Click here	Click here
LMX2492-Q1	Click here	Click here	Click here	Click here	Click here

12.4 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E[™] Online Community *TI's Engineer-to-Engineer (E2E) Community.* Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support TI's Design Support Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.5 Trademarks

E2E is a trademark of Texas Instruments. All other trademarks are the property of their respective owners.

12.6 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.



13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
LMX2492QRTWRQ1	ACTIVE	WQFN	RTW	24	1000	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 125	X2492Q	Samples
LMX2492QRTWTQ1	ACTIVE	WQFN	RTW	24	250	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 125	X2492Q	Samples
LMX2492RTWR	ACTIVE	WQFN	RTW	24	1000	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	X2492	Samples
LMX2492RTWT	ACTIVE	WQFN	RTW	24	250	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	X2492	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.



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OTHER QUALIFIED VERSIONS OF LMX2492, LMX2492-Q1 :

- Catalog: LMX2492
- Automotive: LMX2492-Q1

NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Automotive Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMX2492QRTWRQ1	WQFN	RTW	24	1000	178.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1
LMX2492QRTWTQ1	WQFN	RTW	24	250	178.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1
LMX2492RTWR	WQFN	RTW	24	1000	178.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1
LMX2492RTWT	WQFN	RTW	24	250	178.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1

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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMX2492QRTWRQ1	WQFN	RTW	24	1000	213.0	191.0	55.0
LMX2492QRTWTQ1	WQFN	RTW	24	250	213.0	191.0	55.0
LMX2492RTWR	WQFN	RTW	24	1000	213.0	191.0	55.0
LMX2492RTWT	WQFN	RTW	24	250	213.0	191.0	55.0

MECHANICAL DATA

RTW0024A





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