



Typical Applications

Programmable divider for offset synthesizer and variable divide by N applications:

- Satellite Communication Systems
- Point-to-Point and Point-to-Multi-Point Radios
- LMDS
- SONET

Functional Diagram



HMC394LP4 / 394LP4E

GaAs HBT PROGRAMMABLE 5-BIT COUNTER, DC - 2.2 GHz

Features

SSB Phase Noise: -153 dBc/Hz @ 100 kHz Selectable Division from 2 to 32 Parallel 5-Bit Control Wide Input Power Range: -20 to +10 dBm 24 Lead 4x4mm QFN Package: 9 mm²

General Description

The HMC394LP4 & HMC394LP4E are low noise GaAs HBT programmable 5-bit counters in 4 x 4mm leadless surface mount packages. This device allows continuous division from N= 2 to 32 at input frequencies up to 2.2 GHz. The output voltage swing is 800 mV with a duty cycle inversely proportional to N. The low additive SSB phase noise of -153 dBc/Hz at 100 KHz offset makes this counter an excellent choice for low noise synthesizer applications.

Electrical Specifications, $T_{A} = +25^{\circ}$ C, 50 Ohm System, Vcc= 5V

Parameter	Conditions	Min.	Тур.	Max.	Units
Maximum Input Frequency		2.2			GHz
Minimum Input Frequency	Sine Wave Input [1]			0.1	GHz
Input Power Range	Fin = 0.1 to 2.2 GHz	-15	>-20	+10	dBm
Output Power	Divide-by-2		4		dBm
SSB Phase Noise	Fin = 1 GHz, N = 4		-153		dBc/Hz
Output Transition Time			100		ps
Supply Current (Icc)			194		mA

1. Divider will operate down to DC for square-wave input signal.

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Input Sensitivity Window, 5 major Divide Ratio States, T= 25 °C



Output Power, 5 Major Divide Ratio States, T= 25 °C







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SSB Phase Noise Performance, Fin= 1 GHz, N= 4, T= 25 °C



Typical Supply Current vs. Vcc

Vcc (V)	Icc (mA)
4.75	176
5.0	194
5.25	210

Note: Divider will operate over full voltage range shown above.

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2nd Harmonic, Pin= 0 dBm, T= 25 °C



4th Harmonic, Pin= 0 dBm, T= 25 °C



Output Voltage Wavform, Fin= 1 GHz, N= 2, Pin= 0 dBm, T= 25 °C



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3rd Harmonic, Pin= 0 dBm, T= 25 °C



5th Harmonic, Pin= 0 dBm, T= 25 °C



Output Voltage Wavform, Fin= 1 GHz, N= 17, Pin= 0 dBm, T= 25 °C



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Absolute Maximum Ratings

RF Input (Vcc = +5V)	+13 dBm
Vcc	+5.5V
VLogic	-1.6 to -1.2 Vcc
Maximum Channel Temperature	135 °C
Continuous Pdiss (T = 85 °C) (derate 55 mW/°C above 85 °C)	1.155 W
Thermal Resistance (θ_{j-c}) Junction to Case (Ground Paddle)	21.5 °C/W
Storage Temperature	-65 to +150°C
Operating Temperature	+55 to +85°C

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Outline Drawing



BOTTOM VIEW



NOTES:

- 1. LEADFRAME MATERIAL: COPPER ALLOY
- 2. DIMENSIONS ARE IN INCHES [MILLIMETERS]
- 3. LEAD SPACING TOLERANCE IS NON-CUMULATIVE.
- 4. PAD BURR LENGTH SHALL BE 0.15mm MAXIMUM.
- PAD BURR HEIGHT SHALL BE 0.05mm MAXIMUM.
- 5. PACKAGE WARP SHALL NOT EXCEED 0.05mm.
- 6. ALL GROUND LEADS AND GROUND PADDLE MUST BE SOLDERED TO PCB RF GROUND.
- 7. REFER TO HITTITE APPLICATION NOTE FOR SUGGESTED LAND PATTERN.

Package Information

Part Number	Package Body Material	Lead Finish	MSL Rating	Package Marking ^[3]
HMC394LP4	Low Stress Injection Molded Plastic	Sn/Pb Solder	MSL1 ^[1]	H394 XXXX
HMC394LP4E	RoHS-compliant Low Stress Injection Molded Plastic	100% matte Sn	MSL1 ^[2]	<u>H394</u> XXXX

[1] Max peak reflow temperature of 235 °C

[2] Max peak reflow temperature of 260 °C

[3] 4-Digit lot number XXXX

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Pin Description

Pin Number	Function	Description	Interface Schematic	
1 - 5	AO - A4	CMOS compatible control input bit 0 (LSB) - 4.		
6, 9, 10, 11, 12, 15, 18, 19, 20, 21, 22	GND	Ground: Backside of package has exposed metal ground slug which must be connected to ground.	GND =	
7, 8, 23, 24	VCC	Supply voltage 5V \pm 0.25V must be applied to all four pins.	Vcc 0	
13	ĪN	RF input 180° out of phase with pin 14 must be AC ground.	Vcc IN	
14	IN	RF input must be DC blocked.	Bias =	
16	OUT	Divided output pulse.	OUT	
17	OUT	Divided output pulse 180° out of phase with pin 16.	OUT	

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HMC394LP4 Programming Truth Table

Function	(LSB) A0	A1	A2	A3	A4
Output Low	0	0	0	0	0
/2	1	0	0	0	0
/ 3	0	1	0	0	0
/ 4	1	1	0	0	0
-	-	-	-	-	-
/ 32	1	1	1	1	1
Note: A0 through A4 are CMOS compatible logic control inputs.					

Application Circuit, (2 GHz - Divide-by-20)



* Note: Mount bypass capacitors as close to pins as possible.

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Evaluation PCB



The circuit board used in the application should use RF circuit design techniques. Signal lines should have 50 Ohm impedance while the package ground leads and backside ground slug should be connected directly to the ground plane similar to that shown. A sufficient number of via holes should be used to connect the top and bottom ground planes. The evaluation circuit board shown is available from Hittite upon request.

List of Materials for Evaluation PCB 104898 ^[1]

Item	Description		
J1 - J3	PCB Mount SMA RF Connector		
C1 - C2 ^[2]	1.0 µF Tantalum Capacitor		
C3 - C4 ^[2]	.01 µF Capacitor, 0603 Pkg.		
C5 - C8	1000 pF Capacitor, 0603 Pkg.		
R1*	Resistor SIP 10 K ohm		
S1 - S6	Jumper (shunt) 2mm		
U1	HMC394LP4 / HMC394LP4E 5-Bit Counter		
PCB	104435 Eval Board		

[1] Reference this number when ordering complete evaluation PCB

[2] Optional components.

HMC394LP4 Evaluation PCB Truth Table

Function	S1	S2	S3	S4	S5
Output Low	0	0	0	0	0
/2	1	0	0	0	0
/ 3	0	1	0	0	0
/ 4	1	1	0	0	0
-	-	-	-	-	-
/ 32	1	1	1	1	1
Note: 0 = Jumper Installed. 1 = Jumper Not Installed.					

Note: The evaluation PCB for the HMC394LP4 contains 10K Ohm pull up resistors for each of the five control inputs A0 through A4. Programming the 31 distinct division ratios consists of installing or removing jumpers S1 through S5, as shown above. Jumper S6 must always be installed to provide ground to pin 20.

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Applications Information

Simplified Block Diagram



Asynchronous Programming

The 5-Bit programmable counter counts-down from the programmed value of the data bits to zero and issues an output pulse at the end of each cycle. Settling time of the programmable 5-Bit counter is defined as the maximum time required for the counter to change the division ratio N to a new value after the data bits have settled. The worst case settling time occurs if the data bits A0 thru A4 are changing during the load cycle. Under this condition, the data bits may potentially be erroneous when they are clocked in and in the worst case could be all 1's, requiring 32 clock cycles until the correct data is re-loaded into the flip flops. The worst case asynchronous settling time can be calculated as follows:

T_{SETTLING MAX} = 32/f_{IN} (For Asynchronous Programming)

As an example, if the input frequency is 1 GHz, the maximum settling time is 32 nS



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Synchronous Programming

For applications which can not tolerate a momentary undefined division ratio, which normally occurs while changing the data bits (A0-A4) at random, synchronous programming can be used. Data is loaded into the counter on every rising edge of the clock which occurs while the output (OUT) is "HIGH". The typical minimum setup and hold times are shown in the table below as a function of frequency. For precision applications, the rising edge of the complementary output may be used to latch the new data bits (A0-A4), so that all bits are settled before the next load cycle.

T_{SETTLING MAX} = N/f_{IN} (For Synchronous Programming)

Where N is the desired division ratio, and f_{IN} = Input Frequency (Hz)

Parameter	0.5 GHz	1 GHz	2 GHz
t _{setup}	200 ps	200 ps	200 ps
t _{HOLD}	700 ps	300 ps	120 ps

Programmable Divider Timing Requirements for Synchronous Programming



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CMOS/TTL Input Characteristics

Maximum Input Logic "0" Voltage ($V_{IL MAXIMUM}$) = 1.1V @ 10 uA. Minimum Input Logic "1" Voltage ($V_{IH MINIMUM}$) = 1.8V @ 500 uA. Input IV characteristics for the logic inputs (A0-A4) are shown below:



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