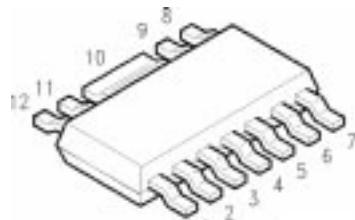


*Datasheet*

- \* Power amplifier for DECT and PCS application
- \* Fully integrated 3 stage amplifier
- \* Operating voltage range: 2.7 to 6 V
- \* Overall power added efficiency 35 %
- \* Input matched to  $50 \Omega$ , simple output match

ESD: **Electrostatic discharge sensitive device,**  
observe handling precautions!



Type	Marking	Ordering code (taped)	Package 1)
CGY 180	CGY 180	Q68000-A8882	MW 12

**Maximum ratings**

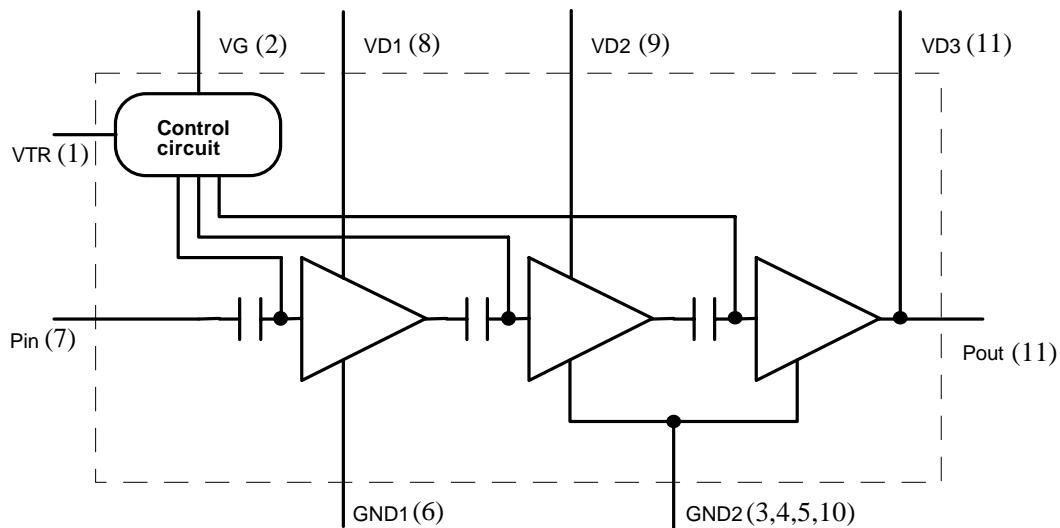
Characteristics	Symbol	max. Value	Unit
Positive supply voltage	$V_D$	8	V
Negative supply voltage <sup>2)</sup>	$V_G$	-8	V
Supply current	$I_D$	1.2	A
Maximum input power	$P_{in,max}$	10	dBm
Channel temperature	$T_{Ch}$	150	°C
Storage temperature	$T_{stg}$	-55...+150	°C
Total power dissipation ( $T_s \leq 81$ °C)	$P_{tot}$	2.3	W
<i>Ts: Temperature at soldering point</i>			
Pulse peak power	$P_{Pulse}$	9.5	W

**Thermal Resistance**

Channel-soldering point	$R_{thChS}$	$\leq 30$	K/W
-------------------------	-------------	-----------	-----

1) Plastic body identical to SOT 223, dimensions see chapter Package Outlines

2)  $V_G = -8V$  only in combination with  $V_{TR} = 0V$ ;  $V_G = -6V$  while  $V_{TR} \neq 0V$

**Functional Block Diagram:**

Pin #	Configuration
1	<b>VTR</b> Control voltage for transmit (0V) / receive (open) mode
2	<b>VG</b> Negative voltage at control circuit (-4V...-8V)
3	<b>GND2</b> RF and DC ground of the 2nd and 3rd stage
4	<b>GND2</b> RF and DC ground of the 2nd and 3rd stage
5	<b>GND2</b> RF and DC ground of the 2nd and 3rd stage
6	<b>GND1</b> RF and DC ground of the 1st stage
7	<b>RFin</b> RF input power
8	<b>VD1</b> Pos. drain voltage of the 1st stage
9	<b>VD2</b> Pos. drain voltage of the 2nd stage
10	<b>GND2</b> RF and DC ground of the 2nd and 3rd stage
11	<b>VD3, Pout</b> Pos. drain voltage of the 3rd stage, RF output power
12	<b>n.c.</b>

**Control circuit:**

**VG** supply: Negative voltage (stabilization is not necessary) in the range of -4V...-8V.

**VTR** supply: During transmit operation: 0V., negative supply current 1mA...2.5mA.

During receive operation: not connected (shut off mode)

The operation current ID of CGY 180 is adjusted by the internal control circuit.

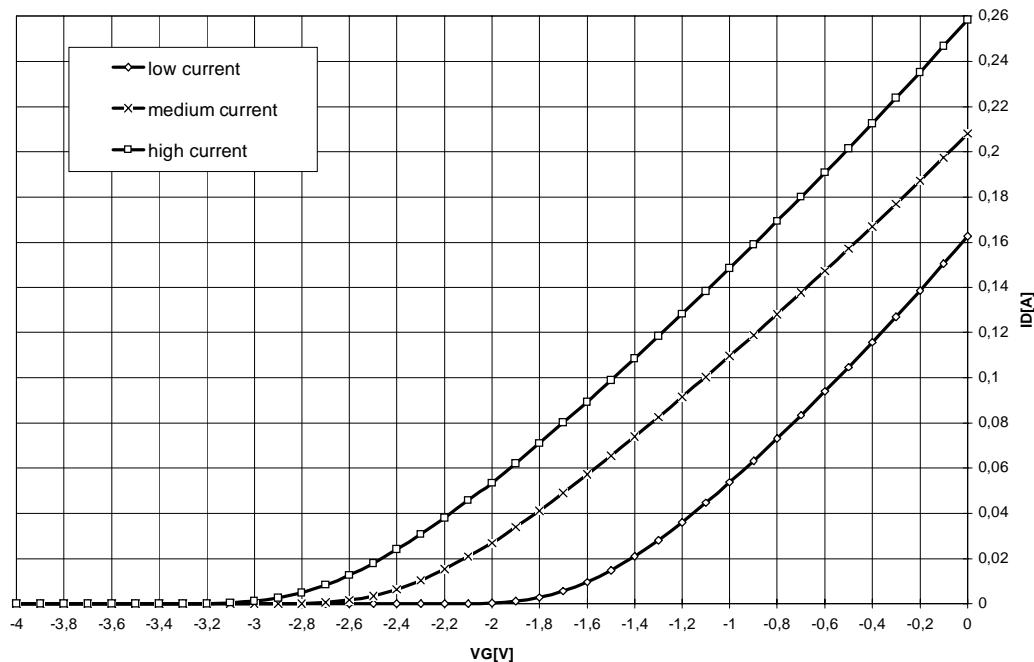
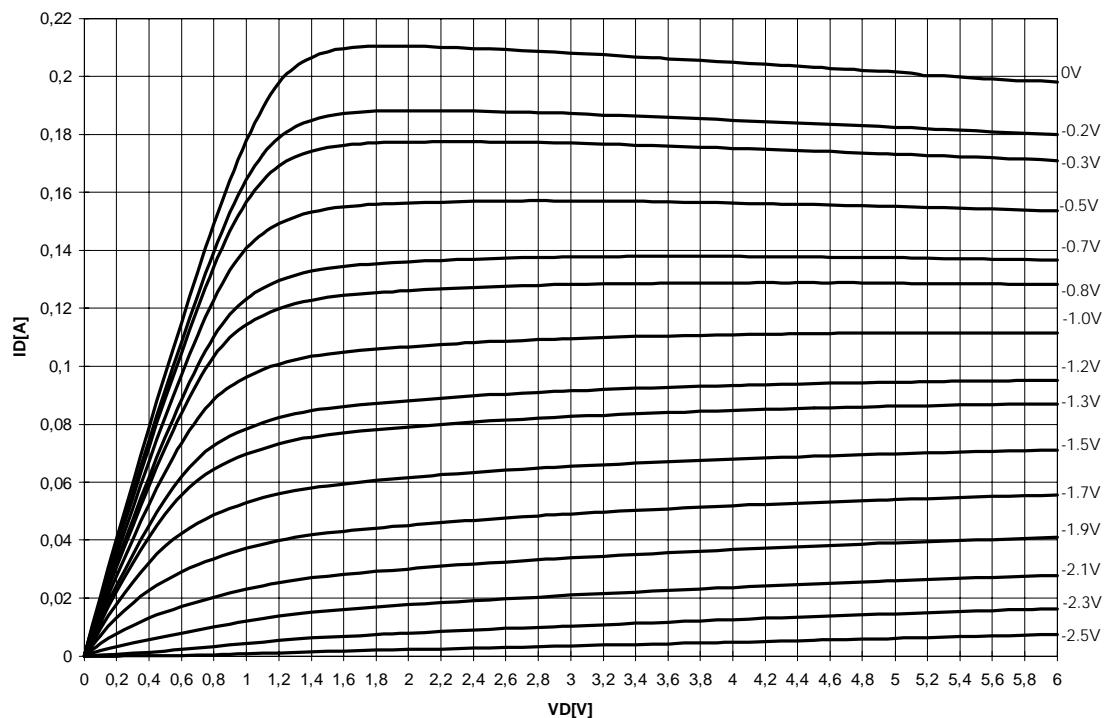
**DC characteristics**

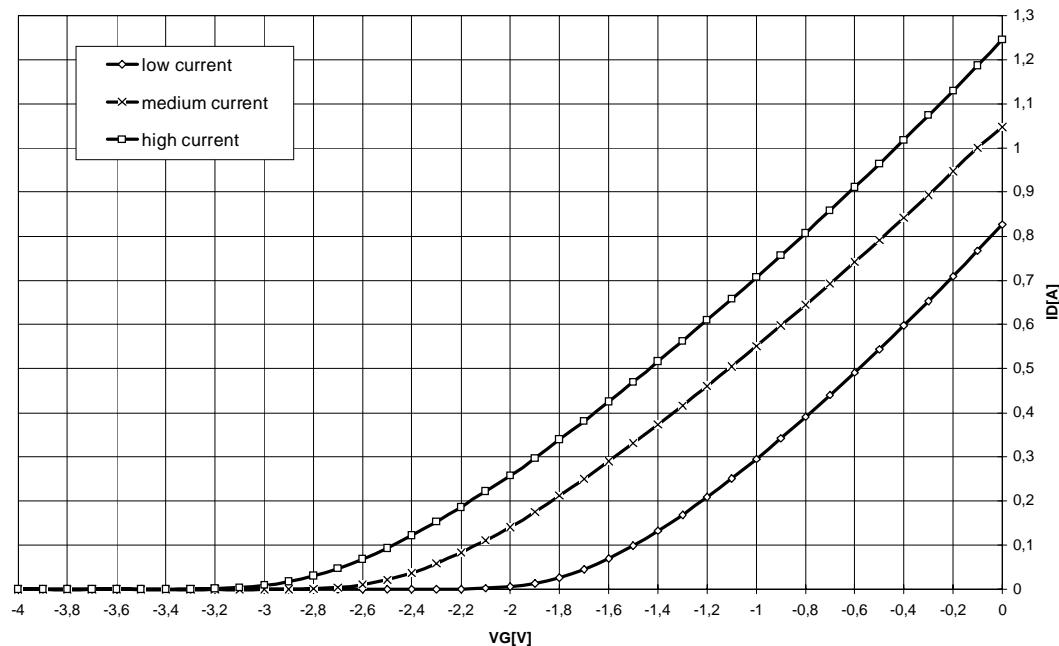
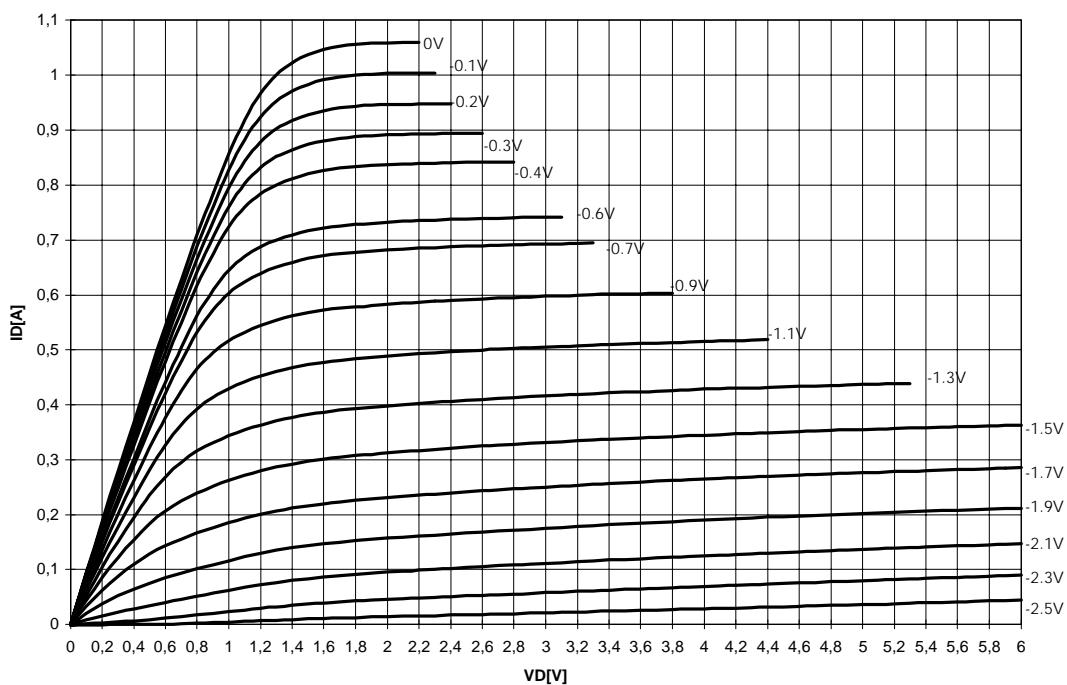
Characteristics	Symbol	Conditions	min	typ	max	Unit
Drain current stage 1	<i>IDSS1</i>	VD=3V, VG=0V, VTR n.c.	150	220	320	mA
	<i>IDSS2</i>		150	220	320	mA
	<i>IDSS3</i>		675	1000	1440	mA
Drain current with active current control	<i>ID</i>	VD=3V, VG=-4V, VTR=0V	290	450	650	mA
Transconductance (stage 1 - 3)	<i>gfs1</i>	VD=3V, ID=90mA	80	100	140	mS
	<i>gfs2</i>	VD=3V, ID=90mA	80	100	140	mS
	<i>gfs3</i>	VD=3V, ID=400mA	360	500	630	mS
Pinch off voltage	<i>Vp</i>	VD=3V, ID<170µA (all stages)	-3.8	-2.8	-1.8	V

**Electrical characteristics**

( $T_A = 25^\circ\text{C}$ ,  $f=1.89 \text{ GHz}$ ,  $Z_S=Z_L=50 \text{ Ohm}$ ,  $VD=3.0V$ ,  $VG=-4V$ , VTR pin connected to ground, unless otherwise specified)

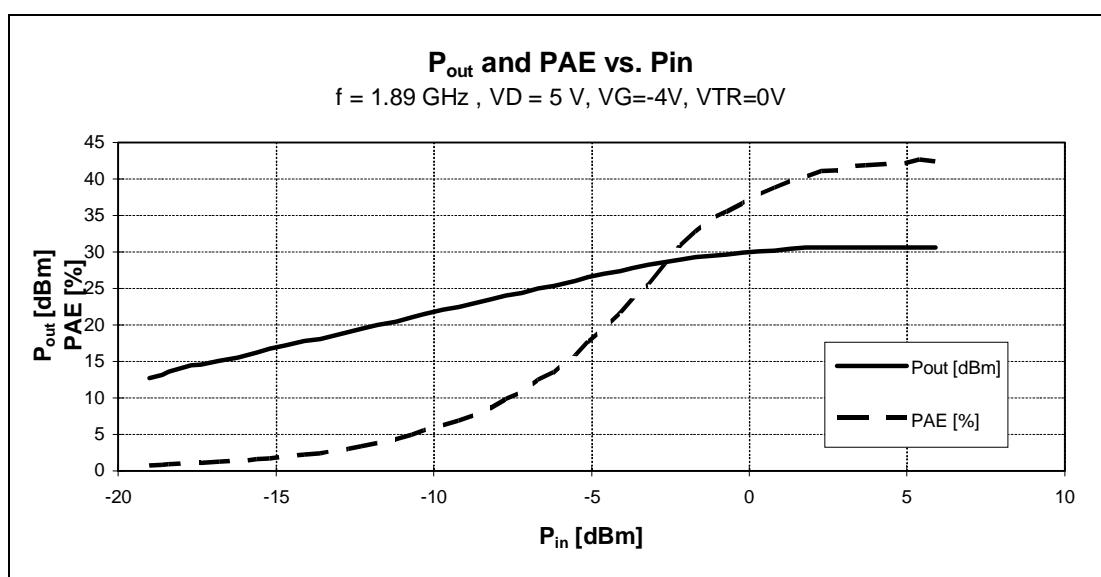
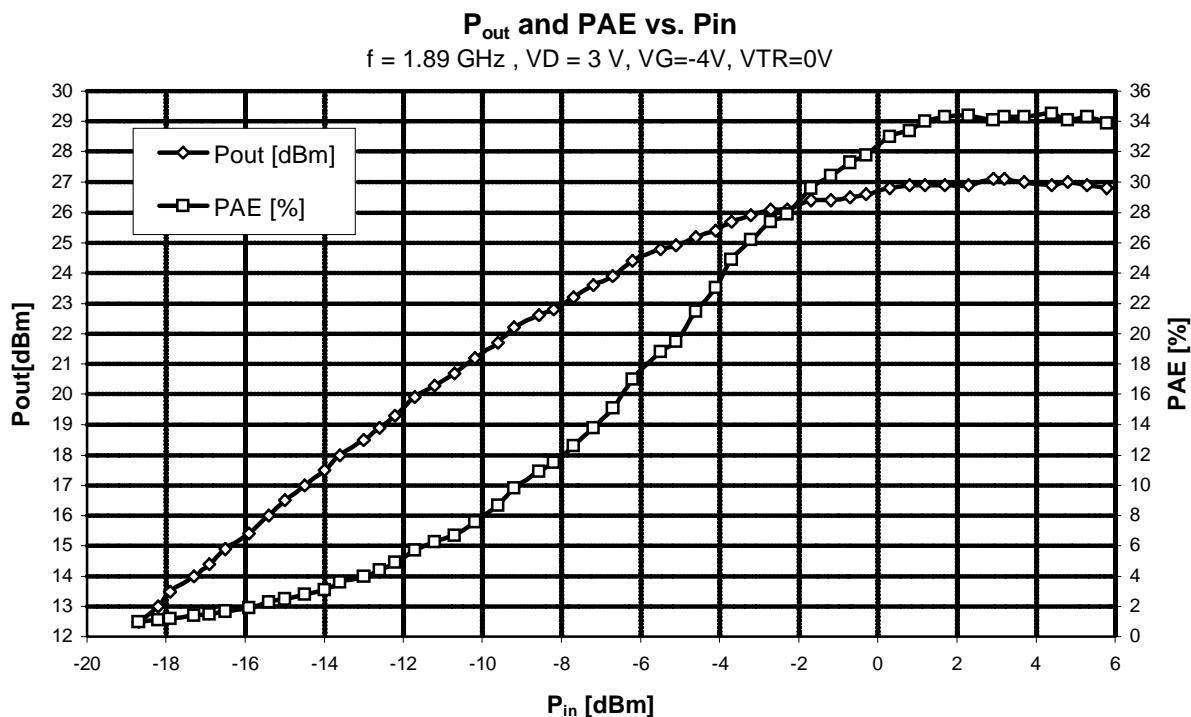
Characteristics	Symbol	min	typ	max	Unit
Supply current <i>Pin = 0 dBm</i>	$I_{DD}$	-	450	-	mA
Negative supply current (transmit operation)	$I_G$	-	1	2.5	mA
Shut-off current VTR n.c.	$I_D$	-	50	180	$\mu\text{A}$
Negative supply current (shut off mode, VTR pin n.c.)	$I_G$	-	10	50	$\mu\text{A}$
Gain <i>P<sub>in</sub> = -20dBm</i>	$G$	28	30	-	dB
Output Power <i>P<sub>in</sub> = 0 dBm</i>	$P_o$	25.5	27	-	dBm
Output Power $VD=5V$ ; $P_{in} = 0 \text{ dBm}$	$P_o$	-	30	-	dBm
Overall Power added Efficiency <i>P<sub>in</sub> = 0 dBm</i>	$\eta$	30	35	-	%
Harmonics ( $P_{in} = 0 \text{ dBm}$ ) $VD=3V$ ; ( $P_{out} = 27 \text{ dBm}$ )	$2f_0$ $3f_0$	- -	- -	-28 -25	dBc
Harmonics ( $P_{in} = 0 \text{ dBm}$ ) $VD=5V$ ; ( $P_{out} = 30 \text{ dBm}$ )	$2f_0$ $3f_0$	- -	- -	-25 -22	dBc
Input VSWR $VD=3V$ ;	-	-	2 : 1	2.5 : 1	-
Third order intercept point $VD=3V$ ; pulsed with a duty cycle of 10%; $f_1=1.8900\text{GHz}$ ; $f_2=1.891728\text{GHz}$ ;	$IP_3$	-	33.5	-	dBm
Third order intercept point $VD=4.8V$ ; pulsed with a duty cycle of 10%; $f_1=1.8900\text{GHz}$ ; $f_2=1.891728\text{GHz}$ ;	$IP_3$	-	38.5	-	dBm
Load mismatch $Pin=0dBm$ , $VD\leq 6V$ , $Z_S=50 \text{ Ohm}$ , Load VSWR = 20:1 for all phase, VTR=0V, VG=-4V	-	No module damage for 10 sec.			-
Stability $Pin=0dBm$ , $VD=2-7V$ , $Z_S=50 \text{ Ohm}$ , Load VSWR = 3:1 for all phase, VTR=0V, VG=-4V	-	All spurious output more than 60 dB below desired signal level			-

**DC - characteristics****Input characteristics** - typical measured values of stage 1 and 2 , VD1 or VD2=3V**Output characteristics** - typical measured values of stage 1 and 2

**Input characteristics - typical measured values of stage 3, VD3 = 3V****Output characteristics - typical measured values of stage 3**

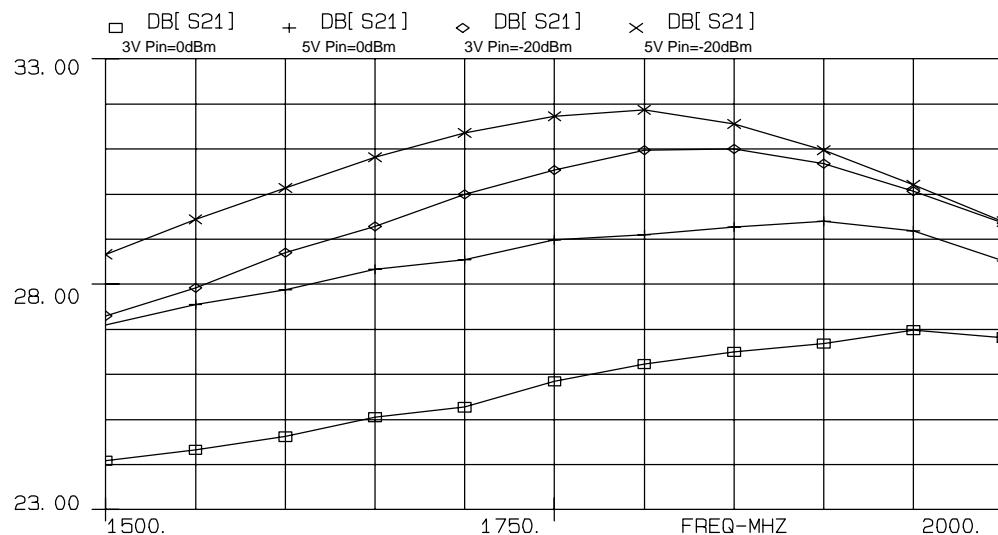
**Output power and power added efficiency**

pulsed mode: ton=1ms, duty cycle 10%

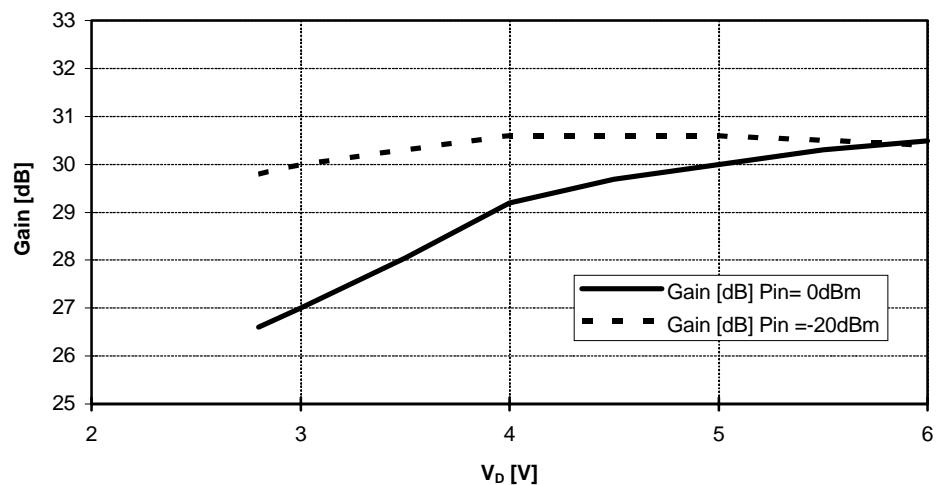


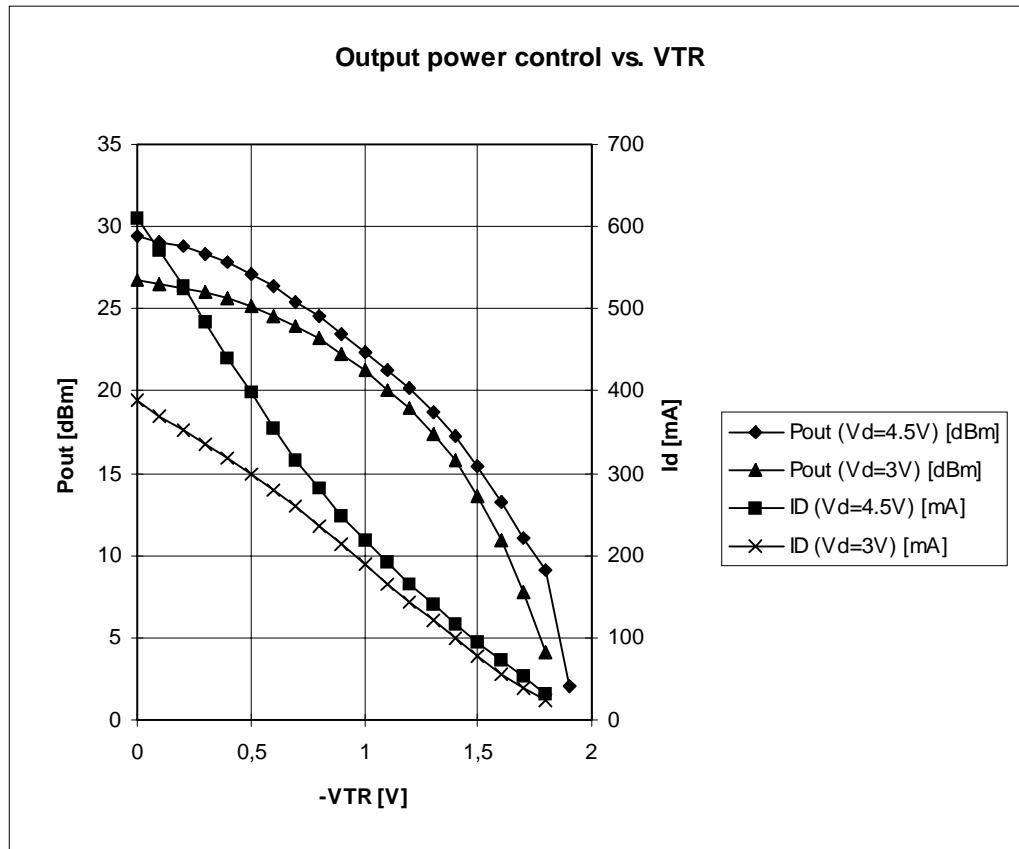
**Gain vs. frequency**

VG=-4V, VTR=0V

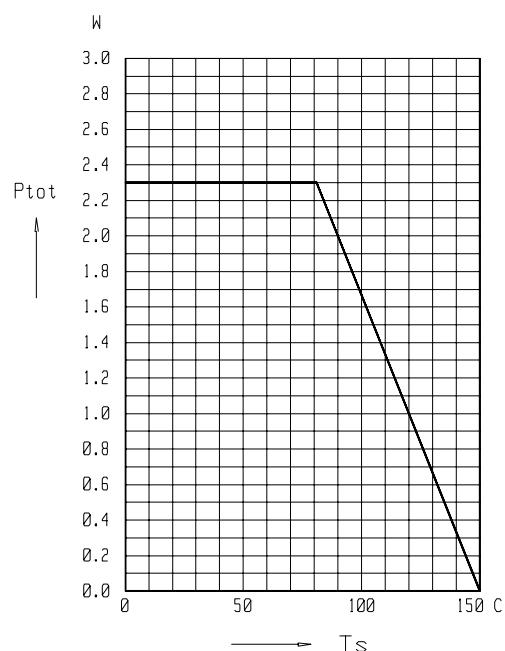
**GAIN vs. DRAIN VOLTAGE**

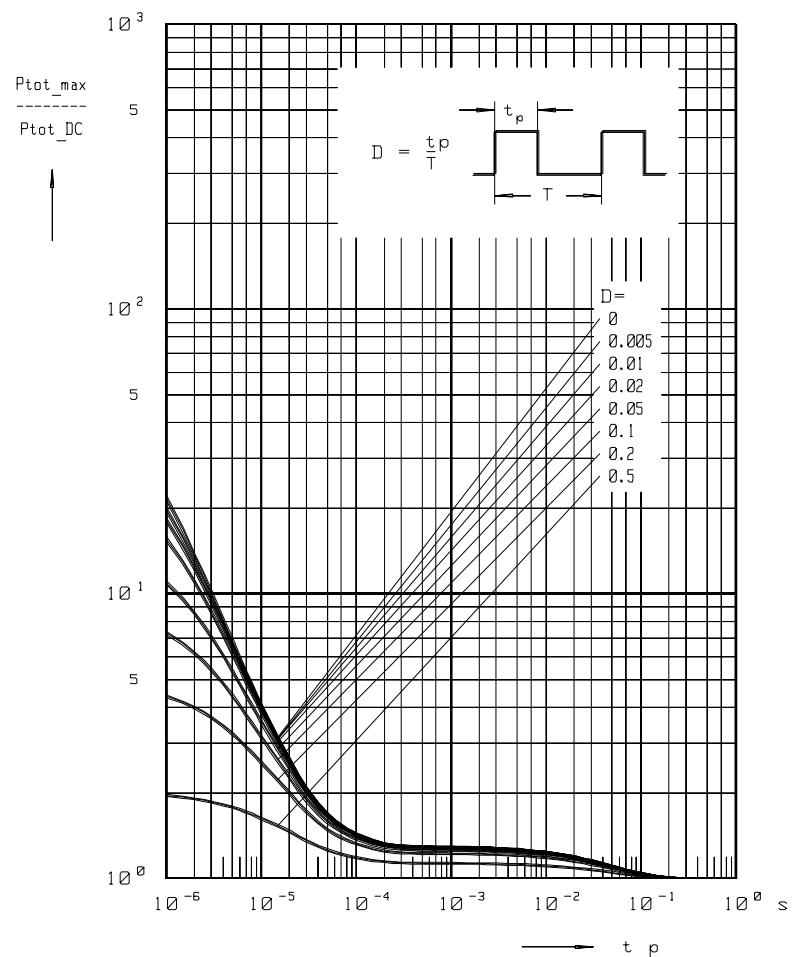
f=1.89 GHz, VD=3V, VG=-4V, VTR=0V

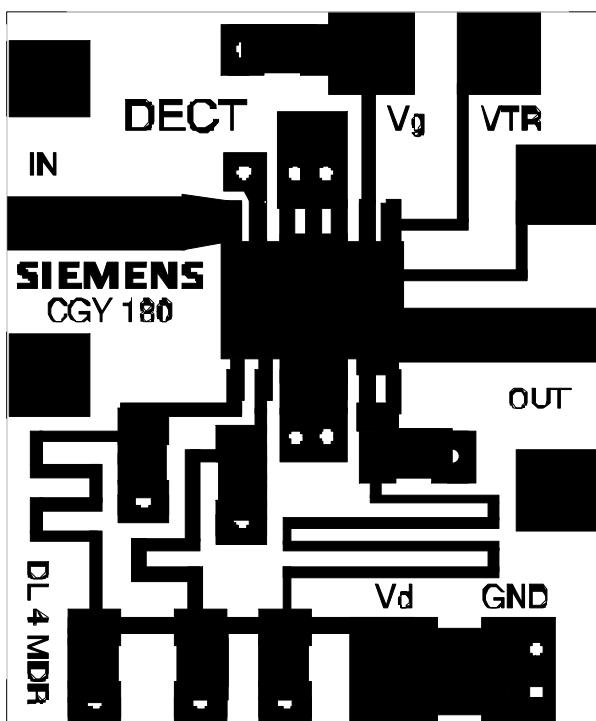




**Total Power Dissipation  $P_{tot}=f(T_s)$**



**Permissible pulse load  $P_{\text{tot\_max}}/P_{\text{tot\_DC}} = f(t_p)$** 

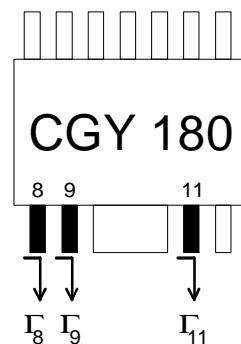
**Test circuit board:**

The following impedances of the bias circuit should be seen from the CGY180 ports:

$$\Gamma_8 = 0.97 / 96^\circ$$

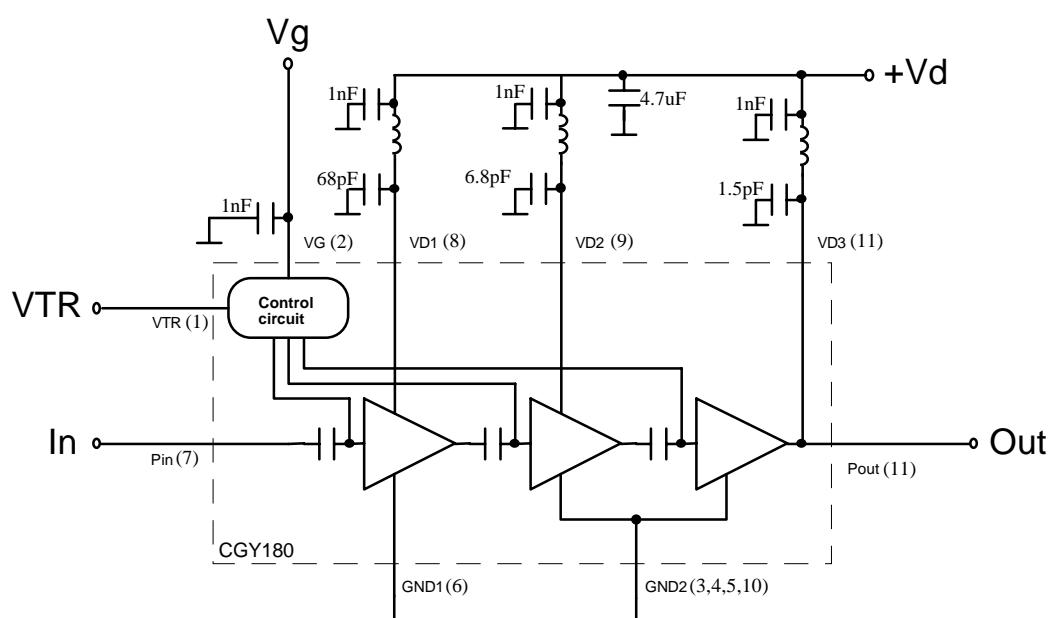
$$\Gamma_9 = 0.96 / 142^\circ$$

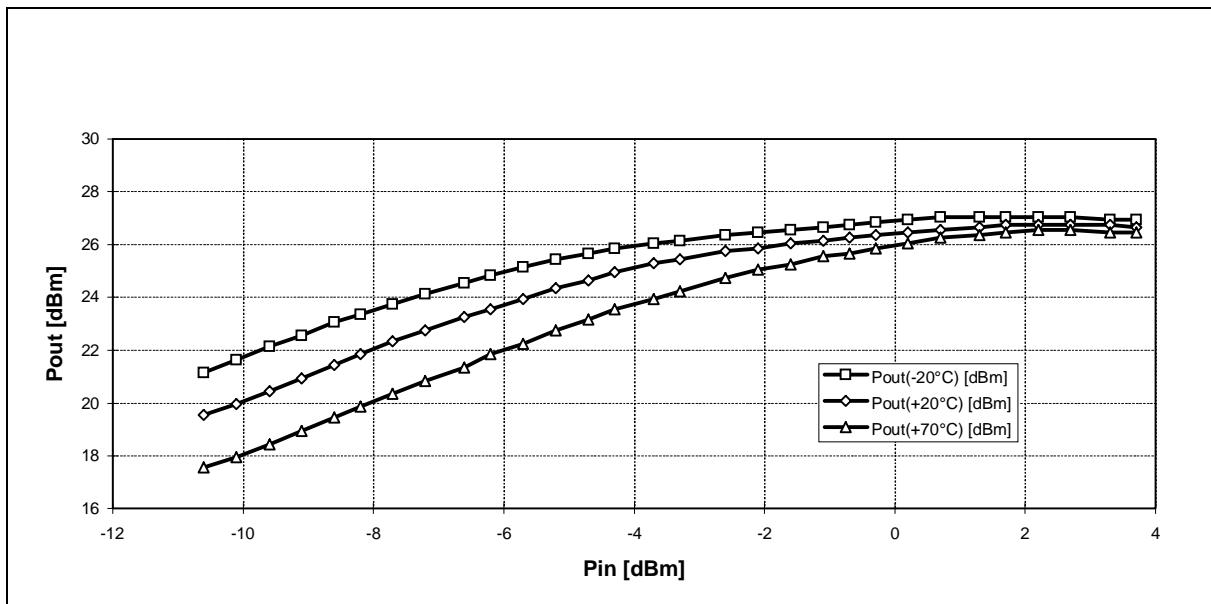
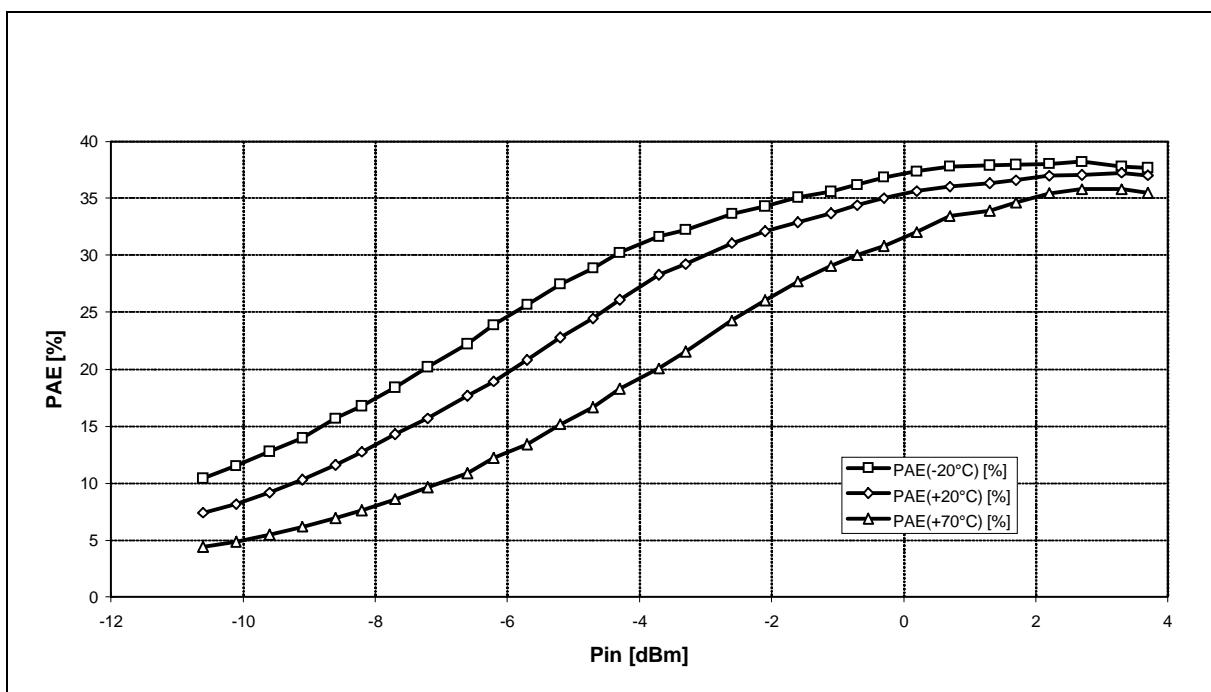
$$\Gamma_{11} = 0.94 / -134^\circ$$



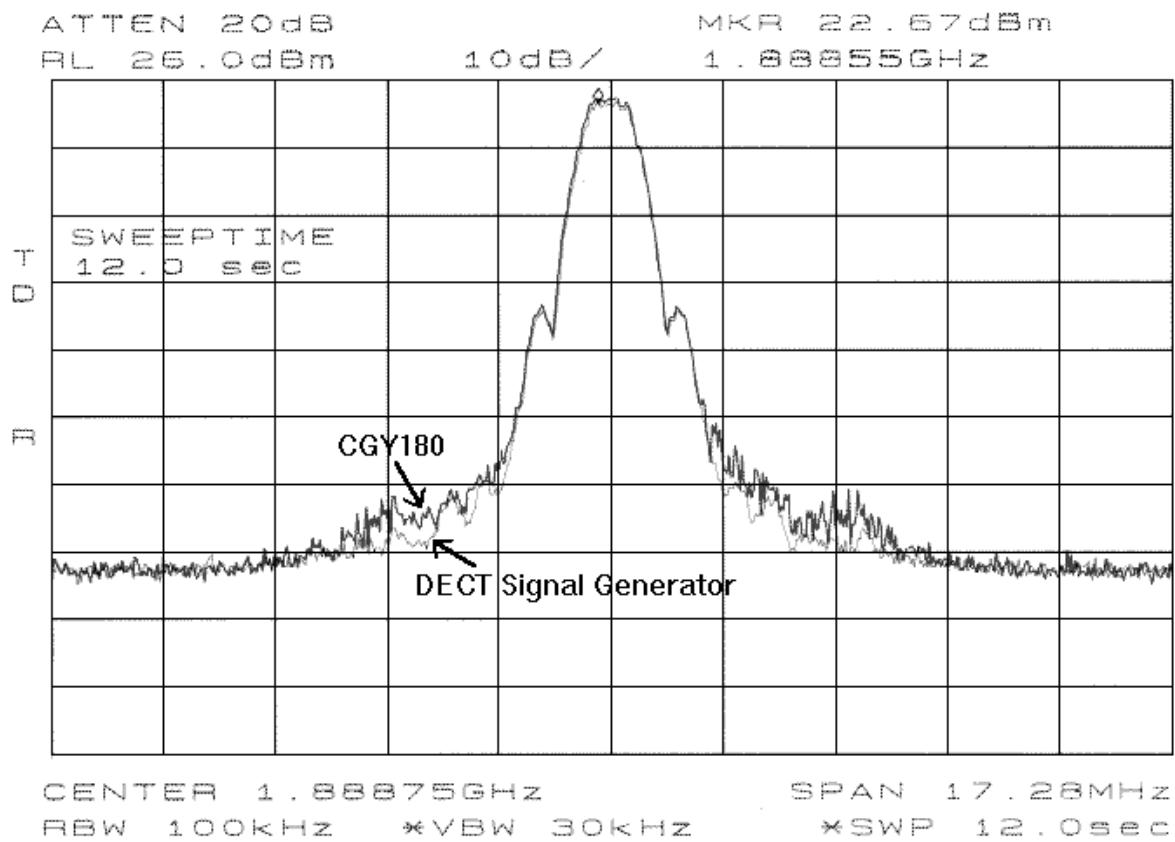
(values measured at f=1.89 GHz)

Size: 20 x 25 mm; In, Out: 50 Ohm

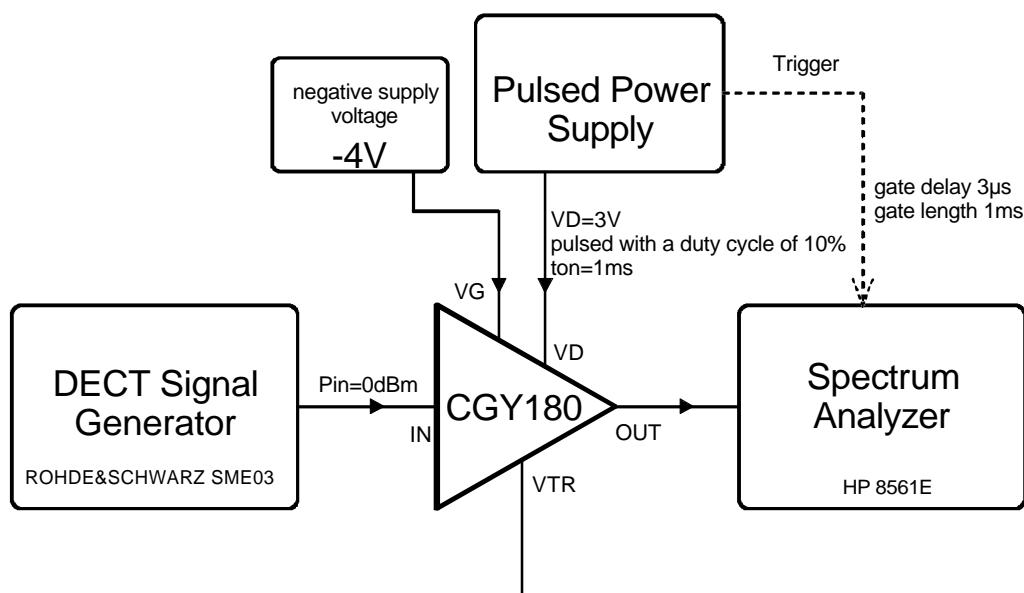
**Principal circuit:**

**Output power at different temperatures\*****Power added efficiency at different temperatures\***

\*)measured with a CGY180 test circuit board (see page 11) VD=3V, VG=-4V, VTR=0V

**Emissions due to modulation:\***

Spectrum of amplified DECT signal  
 Measurement was done with the following equipment:



\*)measured with a CGY180 test circuit board (see page 11) VD=3V, VG=-4V, VTR=0V

## APPLICATION - HINTS

### 1. CW - capability of the CGY180

1.1  $V_D = 3 \text{ V}$

Proving the possibility of CW - operations there must be known the total power dissipation of the device. This value can be found as a function of the temperature in the datasheet (page 8/14). The CGY180 has a maximum total power dissipation of  $P_{\text{tot}} = 2.3 \text{ W}$ .

As an example we take the operating point with a drain voltage  $V_D = 3 \text{ V}$ . The possible ratings of the drain current adjusted by the internal current control of the CGY180 ( $V_G = -4 \text{ V}$ ,  $V_{TR} = 0 \text{ V}$ ) are shown in the following table.

	Min.	Typ.	Max.
$I_D / \text{mA}$	325	450	650

At worst case you see a current of  $I_D = 650 \text{ mA}$ . So the maximum DC - power can be calculated to

$$P_{DC} = V_D \cdot I_D = 1.95W$$

This value is smaller than 2.3W and CW - operation is possible.

1.2  $V_D = 4 \text{ V}$

If you want to use the whole capability of the CGY180, you must consider the power added efficiency PAE. You want to take an operation point of  $V_D = 4 \text{ V}$ . Now there will be a higher current than at  $V_D = 3 \text{ V}$ . We assume a current of  $I_D = 650 \text{ mA}$  and a PAE = 35 %. With these values the DC - power is  $P_{DC} = 2.6 \text{ W}$ . That exceeds the  $P_{\text{totDC}}$  of 2.3 W. Decoupling RF-Power from the CGY180 results in less power dissipation of the device. This is directly correlated with the achieved PAE. To calculate total power dissipation use the formula:

$$P_{\text{totDC}} = P_{DC} (1 - PAE)$$

$P_{\text{tot}}$  for the used operating point shown above will be

$$P_{\text{tot}} = 2.6W(1 - 0.35) = 1.69W$$

It is possible to use the CGY180 for CW - operations up to a drain voltage of  $V_D = 4 \text{ V}$ , if at the same time a PAE of 35% is achieved.

The calculation can be done for any operating point to prove the capability of CW - operation.

## **2. Not using the internal current control**

If you don't want to use the internal current control, it is recommended to connect the negative supply voltage at pin 1 (  $V_{TR}$  ) instead of pin 2 (  $V_G$  ).

## **3. Biasing and use considerations**

In all cases, RF input power should not be applied until the bias voltages have been applied, and RF input power should be turned off prior to removing the bias voltages. Bias application should be timed such that gate voltage (  $V_{GG}$  ) is always applied before the drain voltages (  $V_{DD}$  ), and when returning to the standby mode, gate voltage should only be removed once the drain voltages have been removed.