

Circuit Note CN-0017



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| Devices Connected/Referenced | |
|------------------------------|-----------------------------------|
| AD9779A | Dual 16-Bit, 1 GSPS DAC |
| ADL5371 | 500 MHz to 1500 MHz I/Q Modulator |
| | |

Interfacing the ADL5371 I/Q Modulator to the AD9779A Dual-Channel, 1 GSPS High Speed DAC

CIRCUIT FUNCTION AND BENEFITS

This circuit provides a simple, elegant interface between the ADL5371 I/Q modulator and the AD9779A high speed DAC. The ADL5371 and the AD9779A are well-matched devices because they have the same bias levels and similarly high signal-to-noise ratios (SNR). The matched bias levels of 500 mV allow for a "glueless" interface—there is no requirement for a level shifting network that would add noise and insertion loss along with extra components. The addition of the swing-limiting resistors (RSLI, RSLQ) allows the DAC swing to be scaled appropriately without loss of resolution or of the 0.5 V bias level. The high SNR of each device preserves a high SNR through the circuit.

CIRCUIT DESCRIPTION

The ADL5371 is designed to interface with minimal components to members of Analog Devices family of TxDAC* converters (AD97xx). The baseband inputs of the ADL5371 require a dc common-mode bias voltage of 500 mV. With each AD9779A output swinging from 0 mA to 20 mA,



Figure 1. Interface Between the AD9779A and ADL5371 with 50 Ω Resistors to Ground to Establish the 500 mV DC Bias for the ADL5371 Baseband Inputs (Simplified Schematic)

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By adding resistors RSLI and RSLQ to the interface, the output swing of the DAC can be reduced without any loss of DAC resolution. The resistor is placed as a shunt between each side of the differential pair, as shown in Figure 1. It has the effect of reducing the ac swing without changing the dc bias already established by the 50 Ω resistors and the DAC output current.

The value of this ac swing-limiting resistor is chosen based on the desired ac voltage swing. Figure 2 shows the relationship between the swing-limiting resistor and the peak-to-peak ac swing that it produces when 50 Ω bias-setting resistors are used. Note that all Analog Devices I/Q modulators present a relatively high input impedance on their baseband inputs (typically >1 k Ω). As a result, the input impedance of the I/Q modulator will have no effect on the scaling of the DAC output signal.



Figure 2. Relationship Between the AC Swing-Limiting Resistor and the Peak-to-Peak Voltage Swing with 50 Ω Bias-Setting Resistors

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It is generally necessary to low-pass filter the DAC outputs to remove image frequencies when driving a modulator. The above interface lends itself well to the introduction of such a filter. The filter can be inserted between the dc bias setting resistors and the ac swing-limiting resistor. Doing so establishes the input and output impedances for the filter.

A simulated filter example is shown in Figure 3 with a third-order elliptical filter with a 3 dB frequency of 3 MHz. Matching input and output impedances makes the filter design easier, so the shunt resistor chosen is 100 Ω , producing an ac swing of 1 V p-p differential for a 0 mA to 20 mA DAC full-scale output current. In a practical application, the use of standard value components, along with the input impedance of the I/Q modulator (2900 k Ω in parallel with a few picofarads of input capacitance), will slightly change the frequency response of this circuit.



Figure 3. DAC Modulator Interface with 3 MHz Third-Order, Low-Pass Filter (Calculated Component Values)

All the power supply pins of the ADL5371 must be connected to the same 5 V source. Adjacent pins of the same name can be tied together and decoupled to a large area ground plane with a 0.1 μ F capacitor. These capacitors should be located as close as possible to the device. The power supply can range between 4.75 V and 5.25 V.

The COM1 pin, COM2 pin, COM3 pin, and COM4 pin should be tied to the same ground plane through low impedance paths. The exposed paddle on the underside of the package should also be soldered to a low thermal and electrical impedance ground plane. If the ground plane spans multiple layers on the circuit board, they should be stitched together with nine vias under the exposed paddle. The AN-772 application note discusses the thermal and electrical grounding of the LFCSP_VQ in greater detail.

COMMON VARIATIONS

The interface described here can be used to interface any TxDAC converter with ground referenced 0 mA to 20 mA output currents to any I/Q modulator with a 0.5 V input bias level. For zero-IF applications, the AD9783 dual DAC provides an LVDS interface, while the CMOS-driven AD9788 dual DAC can generate a fine resolution complex IF input to the I/Q modulator. The ADL5370/ADL5371/ADL5372/ADL5373/ ADL5374 family of I/Q modulators provides narrow-band operation with high output 1 dB compression point and OIP3, whereas the ADL5375 provides broadband high performance operation from 400 MHz to 6 GHz. The ADL5385 I/Q modulator uses a 2 × LO and operates from 50 MHz to 2.2 GHz.

LEARN MORE

- AN-772 Application Note, A Design and Manufacturing Guide for the Lead Frame Chip Scale Package (LFCSP). Analog Devices.
- MT-016 Tutorial, *Basic DAC Architectures III: Segmented DACs.* Analog Devices.
- MT-017 Tutorial, *Oversampling Interpolating DACs*. Analog Devices.
- MT-031 Tutorial, *Grounding Data Converters and Solving the Mystery of 'AGND' and 'DGND'*. Analog Devices.
- MT-080 Tutorial, Mixers and Modulators. Analog Devices.
- MT-101 Tutorial, Decoupling Techniques. Analog Devices.
- Zumbahlen, Hank. 2006. *Basic Linear Design*. Analog Devices. ISBN 0915550281. Chapters 4 and 11. Also available as *Linear Circuit Design Handbook*. Elsevier-Newnes, 2008, ISBN 0750687037, Chapters 4 and 11.

Data Sheets

AD9779A Data Sheet.

ADL5371 Data Sheet.

REVISION HISTORY

| 5/09—Rev. 0 to Rev. A | |
|-----------------------------------|-----------|
| Updated Format | Universal |
| 10/08—Revision 0: Initial Version | |

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